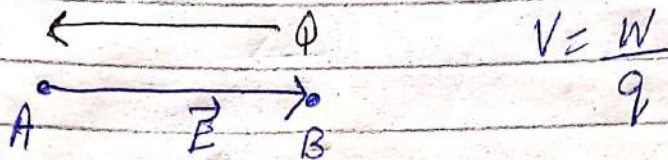


# BASIC ELECTRICAL ENGINEERING

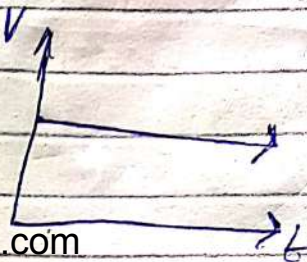
## Voltage or Potential Difference



Energy required to move a charge of  $1C$  from the initial position  $B$  to the final position  $A$  against the electric field.

Constant voltage.

DC voltage



Time Varying voltage

(i) Instantaneous voltage.

Time varying electrical quantities are represented by small letters eg  $v, i$  etc and constant with capital eg  $V, I$  etc.

eg AC voltage

↳ mathematical equation  $v = V_m \sin \omega t$

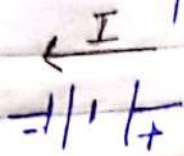
The mathematical equation in terms of time is known as instantaneous equation.

Voltage Rise

Current flows from negative to positive.

voltage drop

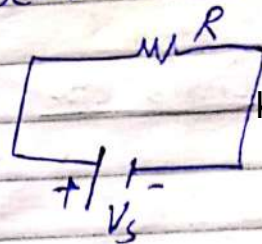
current flows from positive to negative.



The circuit element that delivers energy is known as active circuit element. eg the voltage source  $V_s$  in this example.

The circuit element which takes energy from the source is known as passive circuit element.

eg resistor, capacitor, inductor etc.

Ohm's Law

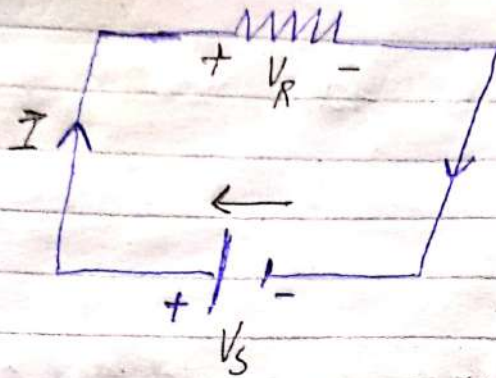
When we apply a voltage through a passive circuit element, there is a current flow through the passive circuit element.

Whenever there is a current in any passive circuit element; it usually results in a voltage drop across that passive circuit element.

In this figure;

$V_s \rightarrow$  voltage rise,  $V_R \rightarrow$  voltage drop

$R \rightarrow$  linear circuit element.



### Kirchoff's voltage Law.

The sum of voltage rises in a loop will be equal to the sum of all the voltage drops in that loop.

As here

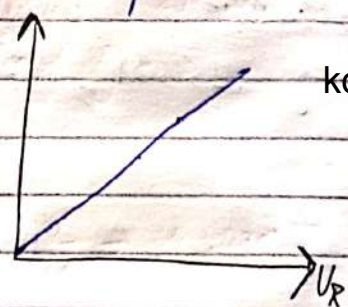
$$V_S = V_R$$

$I$

Ohm's law

$$I \propto V_R$$

$$\textcircled{+} V_R \propto I$$



Any circuit element which has a linear relationship b/w  $I$  and  $V$  is known as linear circuit element. (LCE)  
eg  $R, C, L$ .

Capacitor  $\rightarrow$  L.C.E  $\rightarrow$  for AC voltage

Resistor  $\rightarrow$  L.C.E  $\rightarrow$  for any (AC or DC) voltage.

Inductor  $\rightarrow$  L.C.E  $\rightarrow$  for AC voltage.

A circuit element which allows current to flow through it many directions are called bilateral circuit elements  
eg resistor, Inductor (AC  $V$ ),  
capacitor (DC  $V$ ).

Diode is an example of unilateral circuit element, as it allows current in one direction and blocks in the other.

$$V_s = V_R = IR$$

The property due to which any passive ckt element opposes the flow of current through it is called resistance.

$$I = \frac{V_R}{R} = \frac{V_s}{R}$$

### Electrical Energy:

The amount of energy required to move a charge of  $q$  coulombs against the field.

### Electrical Power:

The electrical energy per unit time.

$$P = \frac{W}{t} = \frac{QV}{t} = IV$$

According to law of conservation of energy

$$\text{Power}_{\text{applied}} = \text{Power}_{\text{dissipated}}$$

Resistor converts the electrical energy into heat energy and dissipates it into the surroundings.

$$P_s = P_R$$

$$V_s I = V_R I = P_R$$

$$P_R = I^2 R$$

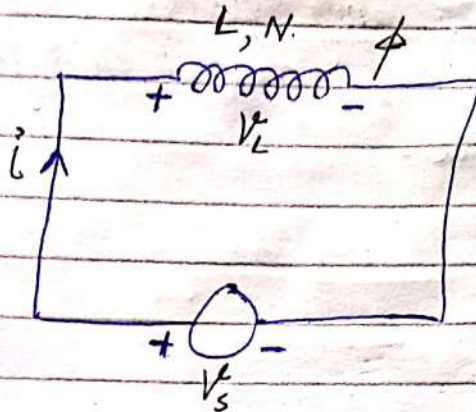
$$P_R = \frac{V_R^2}{R}$$

## Inductor

A straight conductor - even acts as inductor.

conductor  $N=1$ .

To enhance the property of inductivity we make a coil of  $N$  turns.



By increasing  $i$ ,  $\phi$  increases and vice versa.

If current is constant,  $\phi$  is also constant and if it is time varying,  $\phi$  will also be varying.

$$\phi \propto i$$

$$\phi = \frac{L}{N} i \quad \frac{L}{N} \rightarrow \text{constant}$$

For a straight conductor or for a coil of 1 turn i.e.  $N=1$

$$L = \frac{\phi N}{i} \Rightarrow L = \frac{\phi}{i}$$

$$\text{From } \phi = \frac{L}{N} i \Rightarrow N\phi = Li$$

$$V_L = N \frac{d\phi}{dt} = L \frac{di}{dt}$$

$$\textcircled{*} \quad V_L = L \frac{di}{dt} = V_s$$

$V_L = N \frac{d\phi}{dt} \rightarrow$  voltage across inductor from Faraday's law.

If DC is applied, no voltage will be induced i.e.  $V_L = 0$

This inductor behaves as short circuit for DC voltage.

To find current apply KVL;

$$V_S = V_L$$

As  $V_S = V_L = L \frac{di}{dt} \Rightarrow di = \frac{1}{L} V_S dt$

koracademy.com or

$$i = \frac{1}{L} \int V_S dt$$

koracademy.com

Also  $P_S = P_L$

$$P_L = V_L i \Rightarrow$$

$$P_L = L i \frac{di}{dt}$$

$$P_L dt = L i dt \Rightarrow dW = L i di$$

$$W = L \int i di$$

$$W = \frac{L i^2}{2}$$

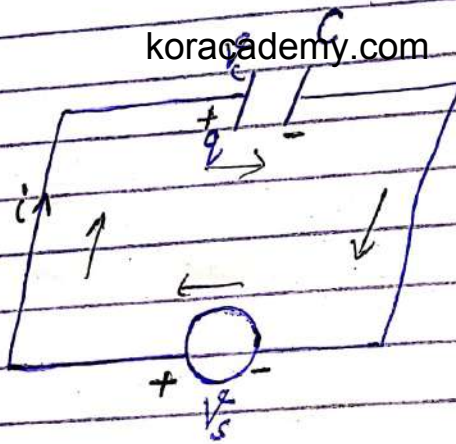
## Capacitor:

Passive circuit element.  
Also known as energy storing element.



Capacitance is the property of a capacitor to store electrical energy!

$$C = \frac{AE}{d}$$



When we apply a voltage through a passive circuit element, it results in a current through the passive circuit element. And the nature of the current depends on the nature of the applied voltage. as in this case time varying.

Applying KVL

$$V_s = V_C$$

$V_s \rightarrow$  Active circuit element.

$C \rightarrow$  Passive circuit element.

$$q \propto V_s \quad \text{or} \quad q \propto V_c$$

$$q = C V_s$$

Differentiating

$$\frac{dq}{dt} = C \frac{dV_s}{dt}$$

$$i = C \frac{dV_c}{dt}$$

If we apply a DC voltage across this capacitor, the voltage will not change;

$$\text{ie } \frac{dV_c}{dt} = 0 \Rightarrow i = 0$$

It means the DC current cannot flow through a capacitor.

In other words this capacitor offers infinite resistance to the flow of DC.

Or this capacitor behaves like an open circuit when connected across DC.

Finding voltage across the capacitor.

$$dV_c = \frac{1}{C} i dt$$

Integrate

$$V_c = \frac{1}{C} \int i dt$$



According to law of conservation of energy

$$P_s = P_c$$

$$V_s i = V_c i$$

$$P_c = V_c i$$

$$P_c = V_c \times C \frac{dV_c}{dt}$$

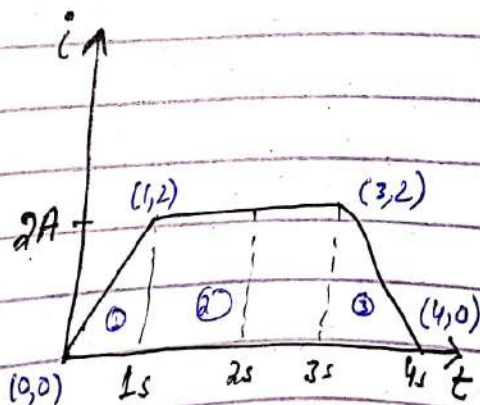
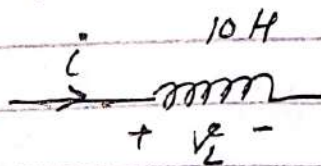
$$P_c dt = V_c \cdot C dV_c$$

$$dW = C V_c dV_c \quad \text{Integrate}$$

$$W = C \int V_c dV_c$$

$$W = \frac{1}{2} C V_c^2$$

Example 1



$V_L, P, W.$

Region 1.       $0 \leq t \leq 1$

$$\text{from } y = mx \Rightarrow i = mt$$

$$\text{for } m \text{ from } \frac{y_2 - y_1}{x_2 - x_1} \Rightarrow m = \frac{i_2 - i_1}{t_2 - t_1}$$

$$= \frac{2 - 0}{1 - 0} \Rightarrow m = 2 \Rightarrow i = 2t$$

$$V_L = L \frac{di}{dt} = 10 \times 2 = \boxed{20V}$$

$$P_L = V_L i = \boxed{40t}$$

$$w = \frac{1}{2} L i^2 = 20t^2$$

$$\text{At } t = 0 \Rightarrow w = 0 \text{ Ws}$$

$$\text{At } t = 0.5 \Rightarrow w = 5 \text{ Ws}$$

$$\text{At } t = 1 \Rightarrow w = 20 \text{ Ws}$$

Region 2.       $1 \leq t \leq 3$

$$i = 2A$$

$$V_L = L \frac{di}{dt} = 0$$

$$P_L = V_L i = 0$$

$$w = \frac{1}{2} L i^2 = 5 \times 4 = 20 \text{ Ws}$$

Region 3.  $3 \leq t \leq 4$

From  $y = mx + c$   
ie A line not passing through origin.  
c is y intercept.

Here  $i = mt + c$

$$m = \frac{i_2 - i_1}{t_2 - t_1} = \frac{0 - 2}{4 - 3} \Rightarrow m = -2$$

from graph and using eq

$$0 = -2 \times 4 + c \Rightarrow c = 8$$

$$\Rightarrow i = -2t + 8$$

$$V_L = L \frac{di}{dt} = 10 \times -2 = -20V$$

$$P_L = V_L i = 40t - 160$$

$$\text{At } t=3 \Rightarrow P_L = -40$$

$$\text{At } t=3.5 \Rightarrow P_L = -20$$

$$\text{At } t=4 \Rightarrow P_L = 0$$

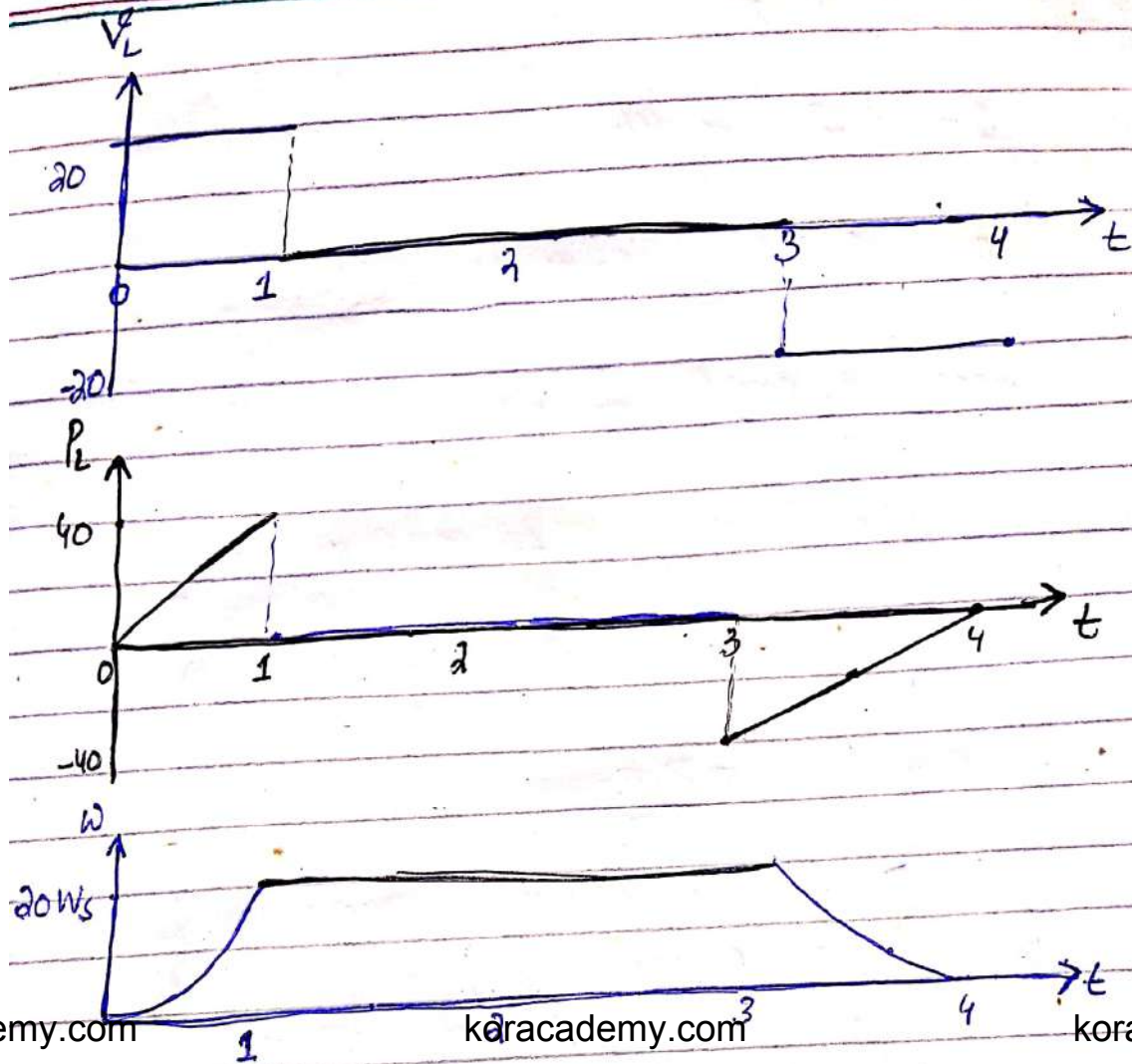
$$W = \frac{1}{2} L i^2 = \frac{1}{2} \times 10 \times (-2t + 8)^2$$

$$W = 5(-2t + 8)^2$$

$$\text{At } t=3 \Rightarrow W = 20$$

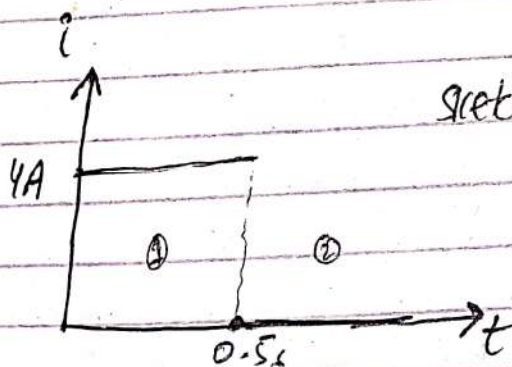
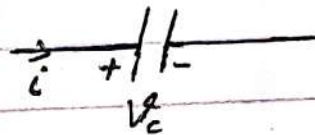
$$\text{At } t=3.5 \Rightarrow W = 5W$$

$$\text{At } t=4 \Rightarrow W = 0$$



Example 2:

$C = 0.1 \text{ F}$



Sketch  $V_C, q, P_C, \omega_C$  as a function of time.

Region 1  $0 \leq t \leq 0.5$

$i = 4 \text{ A}$

$$V_C = \frac{1}{C} \int i dt = \frac{1}{0.1} \int 4 dt = 40 \int dt$$

$$V_c = 40t$$

$$q = C V_c = 4t$$

$$P_c = V_c i = 160t$$

$$W = \frac{1}{2} C V_c^2 = \frac{1}{2} \times 0.1 \times 1600 t^2$$

$$W = 80 t^2$$

$$\text{At } t=0 \Rightarrow W=0$$

$$\text{At } t=0.2 \Rightarrow W=5$$

$$\text{At } t=0.5 \Rightarrow W=20$$

Region 2

$t \geq 0.5$

Under ideal condition, when a capacitor is charged i.e. it has energy; it will store this energy and does not lose it.

$$\text{i.e. } V_c = 20 \text{ V}$$

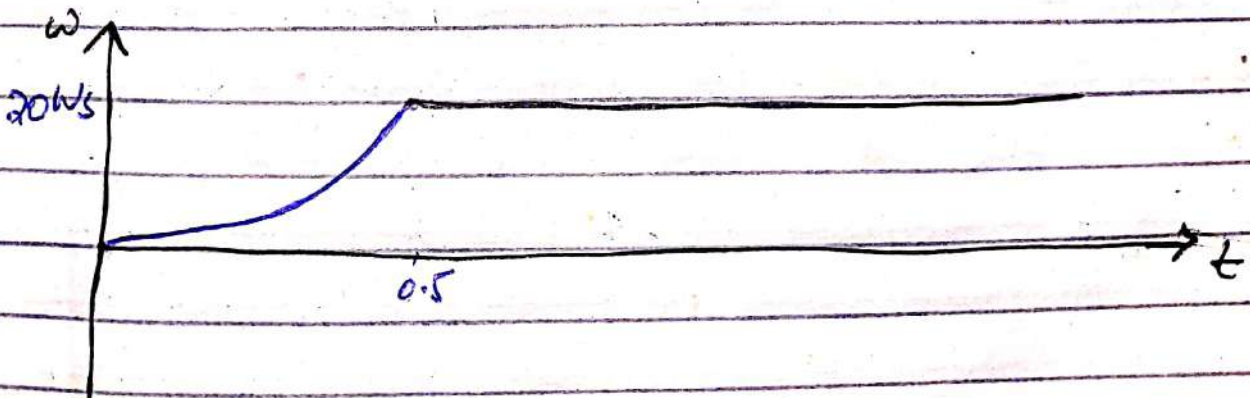
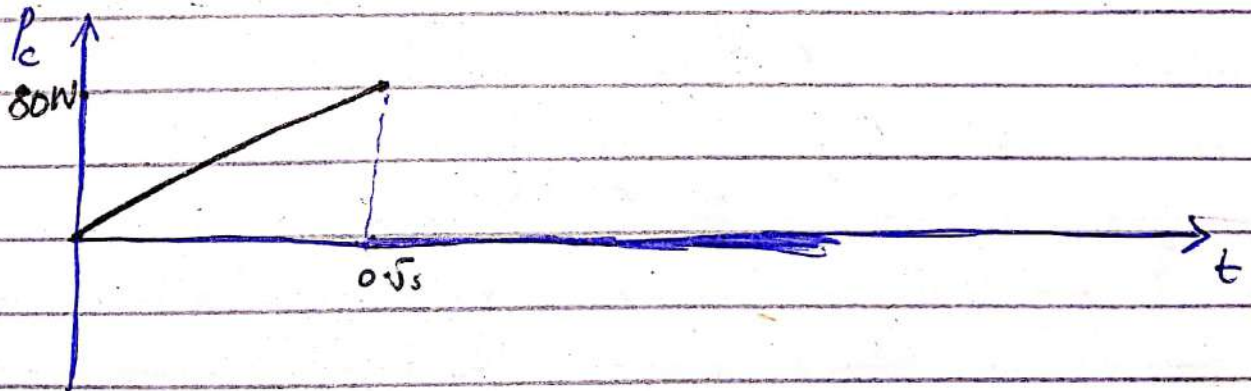
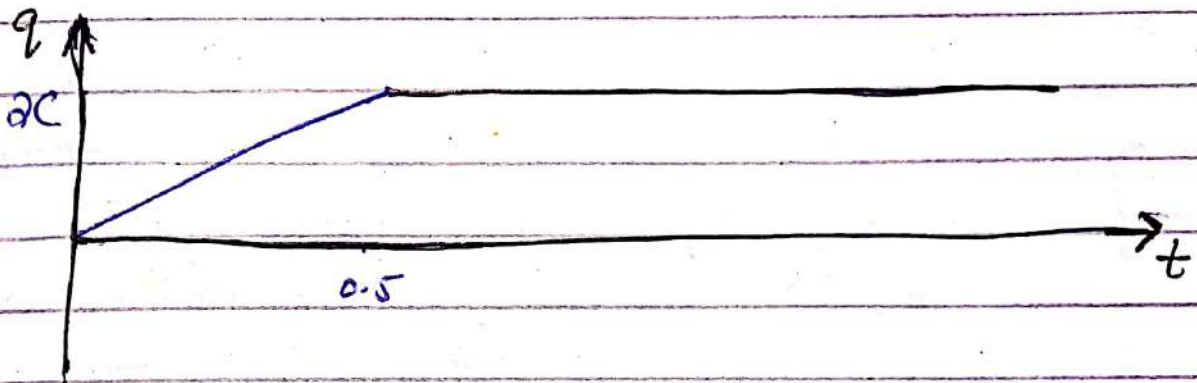
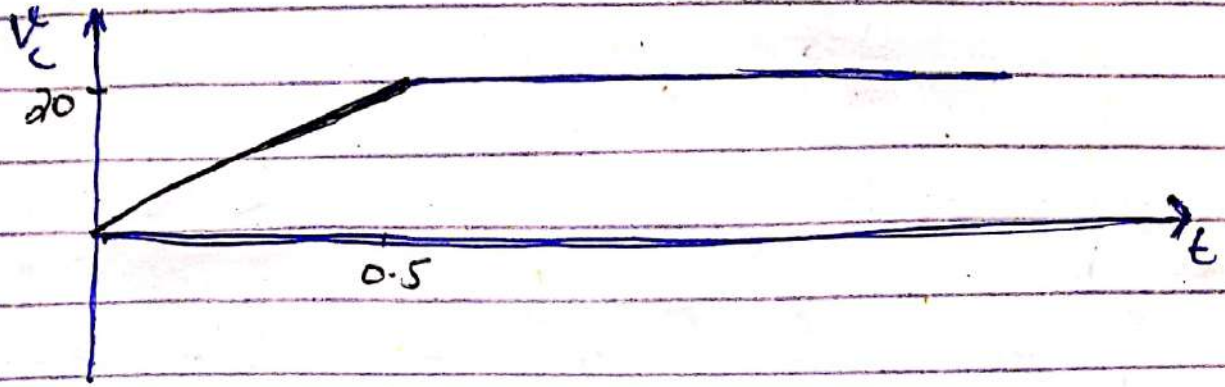
And the charges will also remain on the plates of capacitor forever, i.e.  $q = 2C$

Also we know that  $i = 0$

$$P_c = V_c i = 0$$

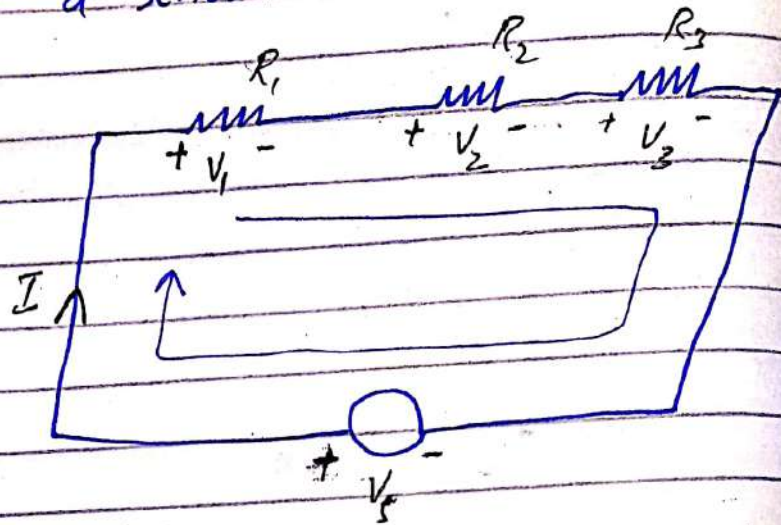
Also the energy remains the same.

$$W = 20 \text{ Ws.}$$



# KIRCHHOFF'S Voltage Law

Consider a series circuit - Resistors in Series



According to KVL

$\sum$  voltage rises =  $\sum$  voltage drops

$$V_s = V_1 + V_2 + V_3$$

$$\textcircled{2} \quad 0 = V_1 + V_2 + V_3 + (-V_s)$$

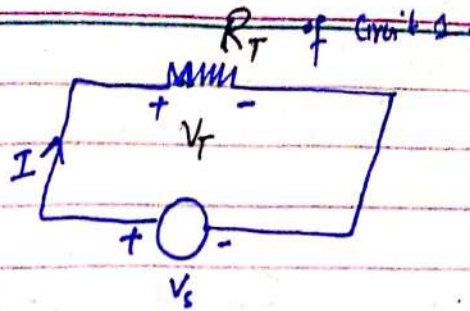
"The algebraic sum of all the voltage drops in a loop in a specific direction is always equal to zero."

$$\text{As } V_1 = IR_1, \quad V_2 = IR_2, \quad V_3 = IR_3$$

$$\Rightarrow V_s = IR_1 + IR_2 + IR_3$$

$$V_s = I (R_1 + R_2 + R_3) \rightarrow (1)$$

$$I = \frac{V_s}{R_1 + R_2 + R_3}$$



Apply KVL

$$V_S = V_T$$

From ohm's law  $V_T = I R_T$

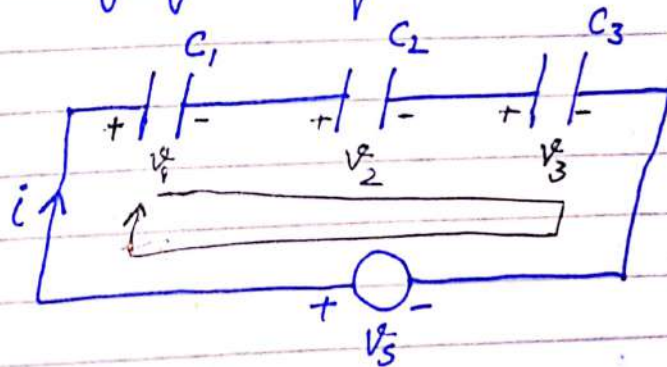
$$\Rightarrow V_S = I R_T \rightarrow \text{(I)}$$

For Eq (I) = Eq (II)

$$R_T = R_1 + R_2 + R_3$$

### Capacitors In Series.

As Capacitor blocks DC so we apply a time varying voltage.



Applying KVL

$$V_S = V_1 + V_2 + V_3$$

As  $\Sigma$  Voltage rises =  $\Sigma$  voltage drops

$$\textcircled{1} \quad 0 = V_1 + V_2 + V_3 + (-V_S)$$



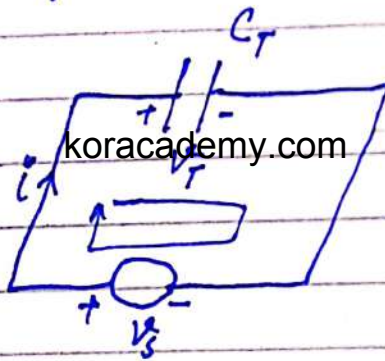
We know that

$$V_1 = \frac{1}{C_1} \int i dt$$

$$V_2 = \frac{1}{C_2} \int i dt \quad V_3 = \frac{1}{C_3} \int i dt$$

$$\Rightarrow V_s = \frac{1}{C_1} \int i dt + \frac{1}{C_2} \int i dt + \frac{1}{C_3} \int i dt$$

Let capacitors are replaced by a single equivalent capacitor.



The current in a capacitive circuit can be changed either by changing the source voltage or the capacitance of the circuit.

Apply KVL to second circuit

$$V_s = V_T$$

$$\text{As } V_T = \frac{1}{C_T} \int i dt$$

$$\text{or } V_s = \frac{1}{C_T} \int i dt \rightarrow \text{(II)}$$

$$V_s = \frac{1}{C_1} + \frac{1}{C_2} + \frac{1}{C_3} \quad (\int idt) \rightarrow \omega$$

Eq I = Eq II if;

$$\frac{1}{C_T} = \frac{1}{C_1} + \frac{1}{C_2} + \frac{1}{C_3}$$

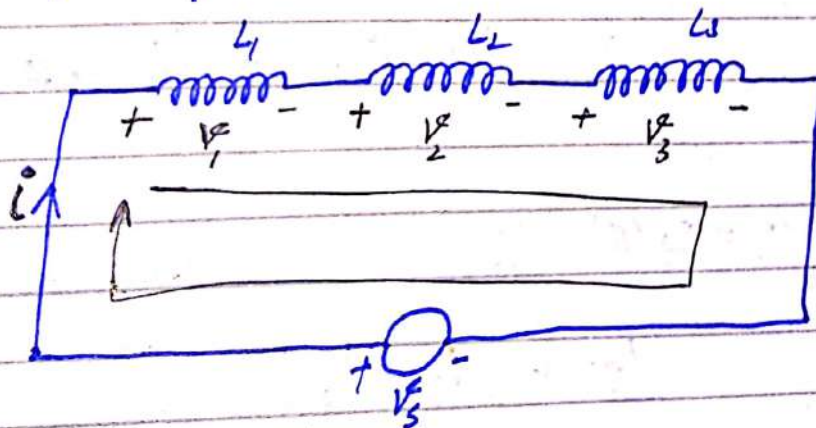
## Inductors In Series:

Assume that the mutual inductance of any two inductors is zero.

Apply a time varying voltage.

Under ideal condition, there will be infinite current in the circuit if DC voltage is applied.

Inductor do not provide opposition (resistance) to the flow of DC.



Apply KVL

$\Sigma$  voltage rise =  $\Sigma$  voltage drops

$$V_s = V_1 + V_2 + V_3$$

$$\Rightarrow 0 = V_1 + V_2 + V_3 + (-V_s)$$

"The algebraic sum of all the voltage drops is zero."

$$V_1 = L_1 \frac{di}{dt}$$

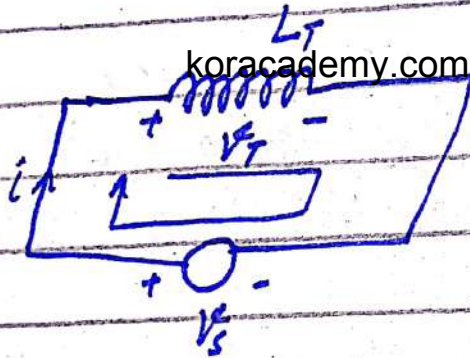
$$V_2 = L_2 \frac{di}{dt}$$

$$V_3 = L_3 \frac{di}{dt}$$

$$\Rightarrow V_S = L_1 \frac{di}{dt} + L_2 \frac{di}{dt} + L_3 \frac{di}{dt}$$

$$V_S = L_1 + L_2 + L_3 \left( \frac{di}{dt} \right) \rightarrow (I)$$

Replace the combination of inductors with a single equivalent inductor.



Apply KVL

$$V_S = V_T$$

$$\textcircled{II} \quad V_S = V_T = L_T \frac{di}{dt} \rightarrow (II)$$

$$\text{Eq (I)} = \text{Eq (II)} \quad \text{If}$$

$$L_T = L_1 + L_2 + L_3$$

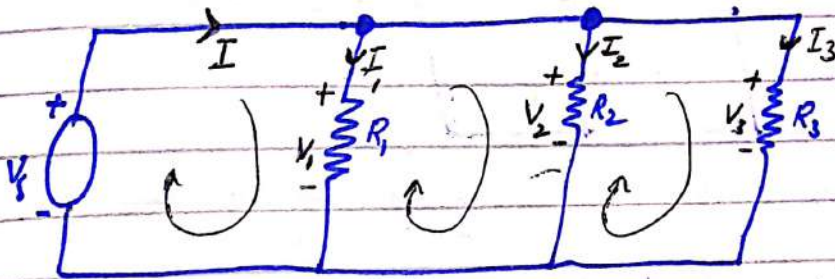
## Kirchoff's Current Law.

In parallel circuit, the voltage across all circuit elements is same.

In series circuit, the current across all the circuit elements is same.

### Resistors In Parallel:

Apply constant voltage.



There is no loop current. basically in any loop but for simple derivation we assume loop currents in the loops in clockwise direction.

Apply KVL

$$\text{In loop 1: } V_s = V_1$$

$$\text{In loop 2: } V_1 = V_2$$

$$\text{In loop 3: } V_2 = V_3$$

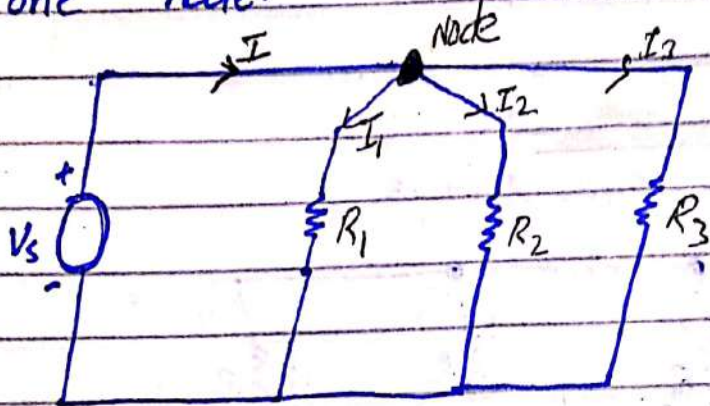
This implies that  $V_s = V_1 = V_2 = V_3$

Junction or Node  $\rightarrow$  point  $\rightarrow$  joins two or more circuit elements  $\rightarrow$  current divides at node.

If these points (where current divides) have the same voltage, they are treated as a single node. If there voltages

are different, they are different nodes.

In the given circuit we have two points where current is dividing but they have the same potential so they are only one node.



KCL states that the sum of all the currents moving towards a node will be equal to sum of all the currents moving away from the node.

$$I = I_1 + I_2 + I_3$$

Total current ( $I$ )  $\rightarrow$  line current

$I_1, I_2, I_3 \rightarrow$  branch current

$$0 = I_1 + I_2 + I_3 + (-I)$$

$\Rightarrow$  KCL can also be stated as;

"The algebraic sum of all the currents moving away from a node is always equal to zero."

$$I_1 = \frac{V_1}{R_1} = \frac{V_s}{R_1}$$

$$I_2 = \frac{V_2}{R_2} = \frac{V_s}{R_2}$$

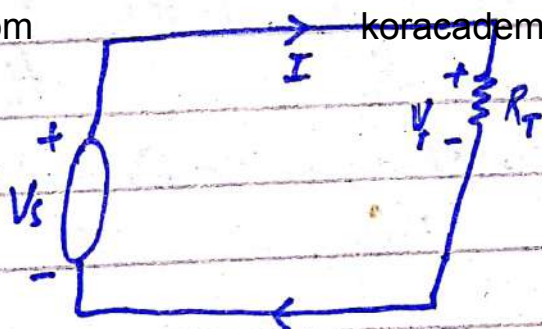
$$I_3 = \frac{V_3}{R_3} = \frac{V_s}{R_3}$$

$$\Rightarrow I = \frac{V_s}{R_1} + \frac{V_s}{R_2} + \frac{V_s}{R_3}$$

$$I = V_s \left( \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} \right) \rightarrow (I)$$

Replace the parallel combination by a single equivalent resistor.

Assume that the single resistor is equal to the total resistance of circuit 2.



According to KVL, in circuit 2.

$$V_s = V_T$$

From Ohm's law  $V_T = IR_T$

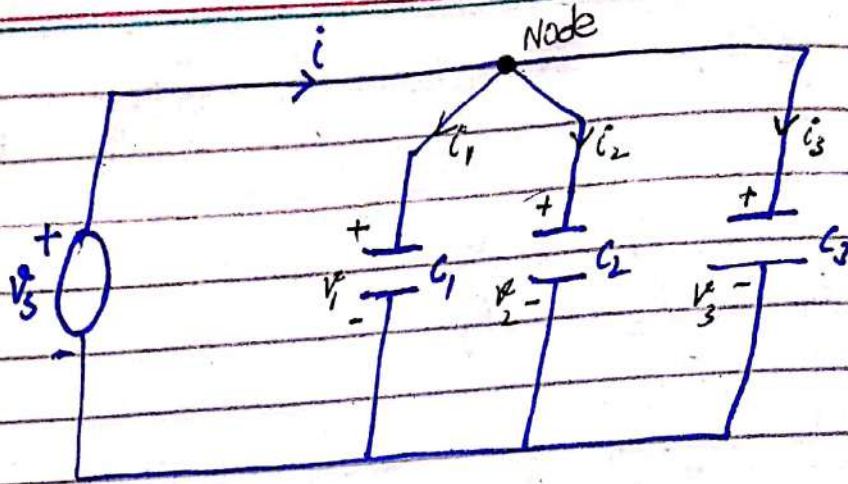
$$\text{or } V_s = IR_T \Rightarrow I = V_s \left( \frac{1}{R_T} \right) \rightarrow (II)$$

For Eq (I) = Eq (II) if

$$\frac{1}{R_T} = \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}$$

Capacitors In Parallel

Apply time varying voltage.



Applying KCL

$$i = i_1 + i_2 + i_3$$

We know that in parallel,  $V_s = V_1 = V_2 = V_3$

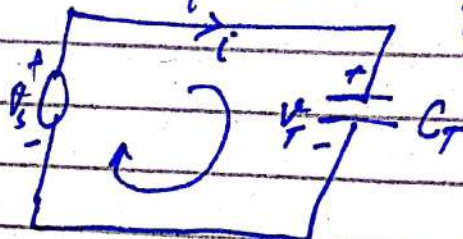
$$\text{Also } i_1 = C_1 \frac{dV_1}{dt} \quad i_2 = C_2 \frac{dV_2}{dt} \quad i_3 = C_3 \frac{dV_3}{dt}$$

$$\Rightarrow i_1 = C_1 \frac{dV_s}{dt} \quad i_2 = C_2 \frac{dV_s}{dt} \quad i_3 = C_3 \frac{dV_s}{dt}$$

$$\Rightarrow i = C_1 \frac{dV_s}{dt} + C_2 \frac{dV_s}{dt} + C_3 \frac{dV_s}{dt}$$

$$\Rightarrow i = (C_1 + C_2 + C_3) \frac{dV_s}{dt} \rightarrow \text{Eq (1)}$$

Replace the parallel combination by a single capacitor; assuming that the capacitance of this single capacitor is equal to the total capacitance of circuit.



Apply KVL

$$V_s = V_T$$

We can say that current in circuit 2 is the current in the circuit.

$$i = C_T \frac{dV_T}{dt} = C_T \frac{dV_s}{dt} \rightarrow \text{Eq. (II)}$$

For Eq (I) to be equal to Eq (II):

$$C_T = C_1 + C_2 + C_3$$

Example 1.1:

$$V_s = 12V \quad I = 20A \quad t = 10 \text{ hours}$$

70% chemical energy is stored in the battery.  
Remaining 30% of energy is lost.

- (a) Cost of charging of the electricity costs 5 cents / kW hr.  
(b) The amount of energy which is

$$W = Pt$$

$$P = V_s I = 12 \times 20 = 240$$

$$W = 240 \times 10 = 2400 \text{ whr} = 2.4 \text{ kWhr}$$

The cost of 1 kW = 5 cents

Cost of 2.4 kW = 12 cents.

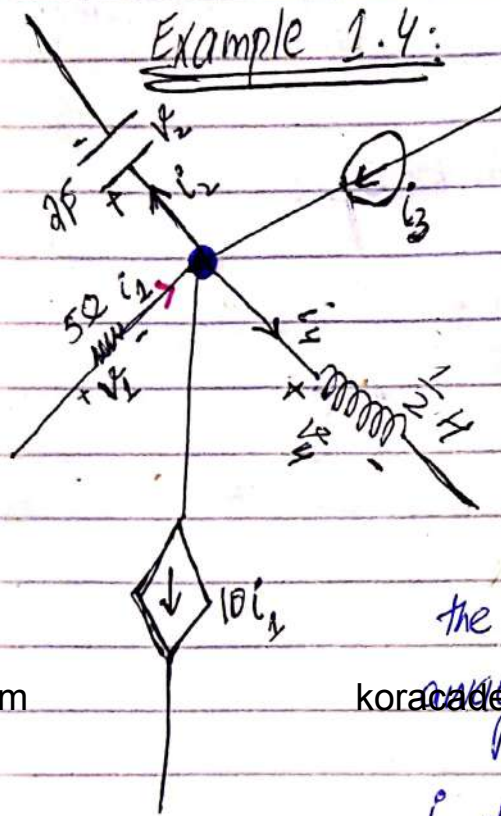
$$(b) 70\% = \frac{7}{100} \times 2400 \text{ whr}$$



$$= 1680 \times 3600 \text{ Wsec}$$

$$= 6048000 \text{ J.}$$

### Example 1.4:



Data

$$V_1 = 10e^{-at} \text{ Volts}$$

$$V_2 = 2e^{-at} \text{ Volts}$$

$$i_3 = 6e^{-at} \text{ Amp}$$

$$V_4 = ?$$

According to KCL, the current flowing towards the node is equal to the sum of currents flowing away from the node.

$$i_1 + i_3 = i_2 + i_4 + 10i_1$$

$$i_4 = i_1 + i_3 - i_2 - 10i_1 \rightarrow (I)$$

$$V_1 = i_1 R_1 \Rightarrow i_1 = \frac{V_1}{R} = \frac{10e^{-at}}{5}$$

$$i_2 = 2e^{-at} \quad \downarrow \frac{1}{R}$$

$$i_2 = C \frac{dV_2}{dt} = 2 \times 4e^{-2t} \quad i_2 = -8e^{-at} \text{ Amp}$$

This negative sign shows the direction of the current in the circuit is wrong.  
(But we don't change it).

Eq (1)  $\Rightarrow i_4 = 2e^{-2t} + 6e^{-2t} + 8e^{-2t} - 20e^{-2t}$

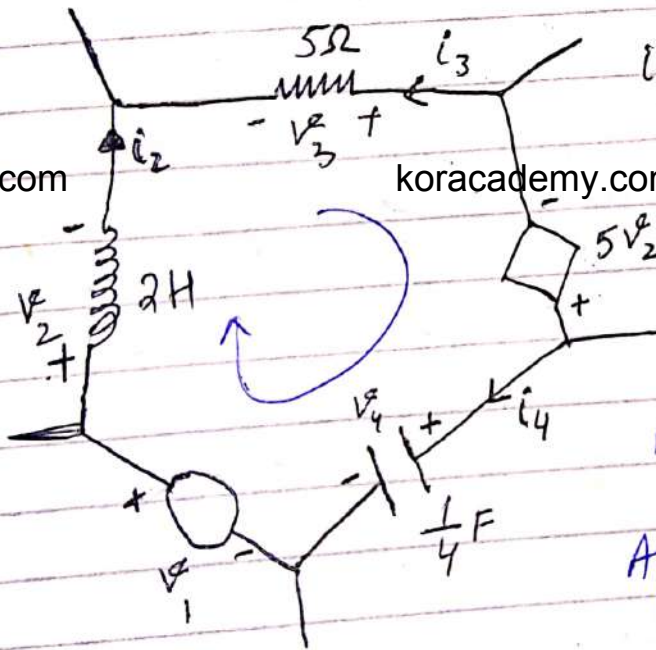
$i_4 = -4e^{-2t}$

Again negative sign  $\rightarrow$  assumed direction is wrong.  $\rightarrow$  actually the current flows in the opposite direction.

$v_4 = L \frac{di_4}{dt} = \frac{1}{2} \times 8e^{-2t} = \underline{\underline{4e^{-2t}}}$  Ans

Example 1.5.

$v = 12 \sin \omega t$  volts  
 $i_2 = 4 \sin \omega t$  A  
 $i_3 = 2 \sin \omega t + 4 \cos \omega t$  A  
 $i_4 = ?$



We assume that there is a loop current flowing in clockwise direction.

Apply KVL;

$\Sigma$  voltage rises =  $\Sigma$  voltage drop.  
 $5v_2 + v_1 + v_3 = v_2 + v_4$

(1)  $v_4 = 5v_2 + v_1 + v_3 - v_2 \rightarrow (1)$

$v_2 = L \frac{di_2}{dt} = 2 \times 8 \cos \omega t \Rightarrow v_2 = 16 \cos \omega t$

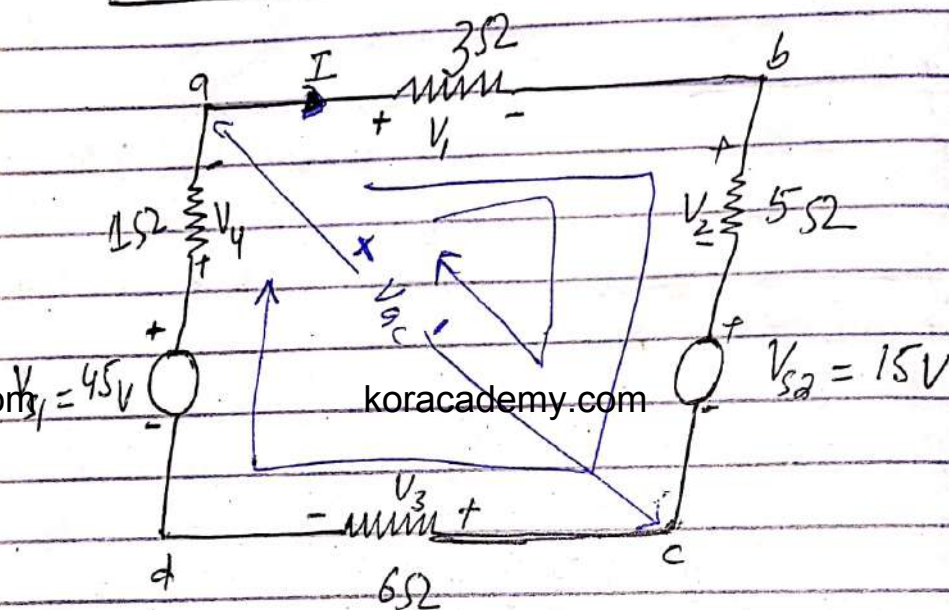
$v_1 = iR = 5 \times 2 = 10 \sin \omega t - 20 \cos \omega t$  V

$$\text{Eq (D)} \Rightarrow V_4 = 80 \cos at + 12 \sin at + 10 \sin at - 20 \cos at - 16 \cos at$$

$$V_4 = 22 \sin at + 44 \cos at.$$

$$i_4 = C \frac{dV_4}{dt} \Rightarrow \boxed{i_4 = 11 \cos at - 22 \sin at} \quad \text{Ans.}$$

### Exercise 2: 2.21



Calculate

- The current from a to b.
- The voltage  $V_{ac}$ .
- The power supplied by  $V_{s2}$ .

Apply KVL  $\Sigma$  voltage rise =  $\Sigma$  voltage drop.

$$V_{s1} = V_4 + V_1 + V_2 + V_{s2} + V_3$$

$$V_1 + V_2 + V_3 + V_4 = V_{s1} - V_{s2}$$

$$V_1 + V_2 + V_3 + V_4 = 30 \text{ V}$$

In series circuit, the current through all circuit elements is the same.

$$3I + 5I + 6I + I = 30$$

$$15I = 30$$

$$\Rightarrow I = 2A.$$

If voltage across source is voltage rise it will deliver power and if voltage across source is voltage drop it will absorb power.

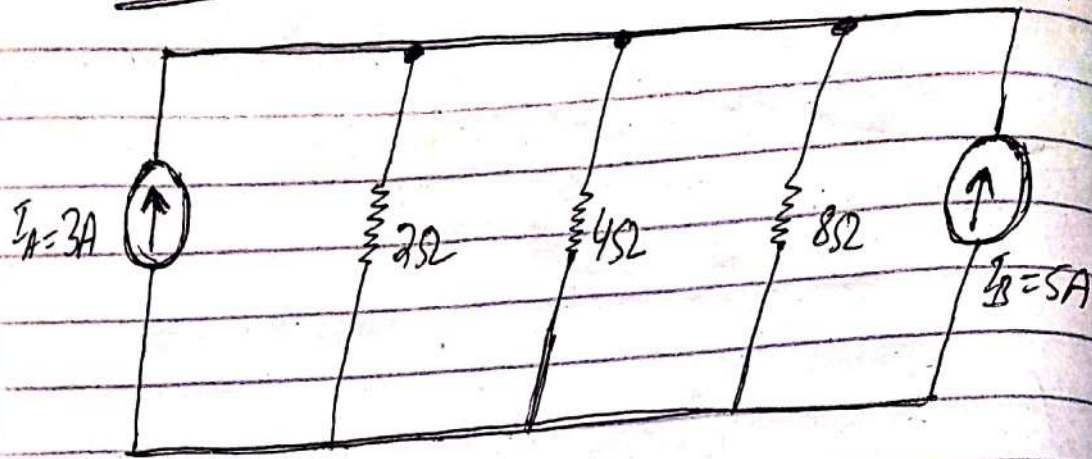
$$(c) P_s = V_s I = 45 \times 2 = 90W.$$

(b) Suppose there are two triangular loops (abc and acd). We consider the upper loop i.e. abc. and apply KVL to it.

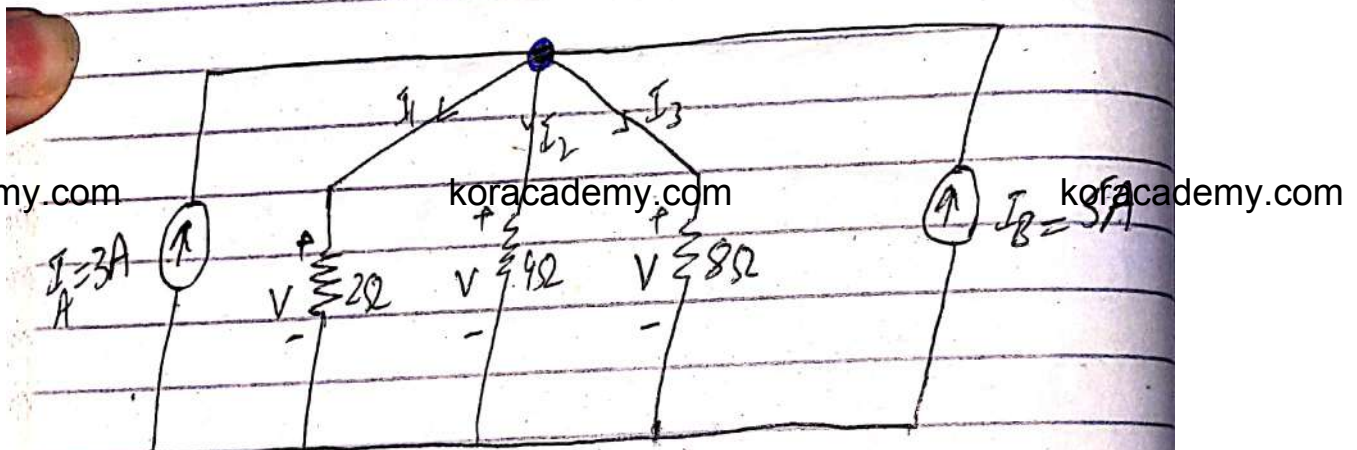
$$V_{ac} = V_1 + V_2 + V_{s2}$$

$$V_{ac} = 6 + 10 + 15 = \underline{31 \text{ volts.}}$$

(d) Power absorbed by segment BC = 50 W  
ie by resistors  $I^2 R$   
and the source also.



There are three points in this circuit where the current divides, but they are at the same potential, so it is only a single node.



In parallel circuit, voltage is same.

- (a)  $V = ?$  (b)  $I_1, I_2, I_3 = ?$   
 (c) Power supplied by  $I_B$ .

Apply KCL.

$$I_A + I_B = I_1 + I_2 + I_3$$

$$8 = \frac{V}{2} + \frac{V}{4} + \frac{V}{8}$$

$$\Rightarrow 9.14 \text{ volts.}$$

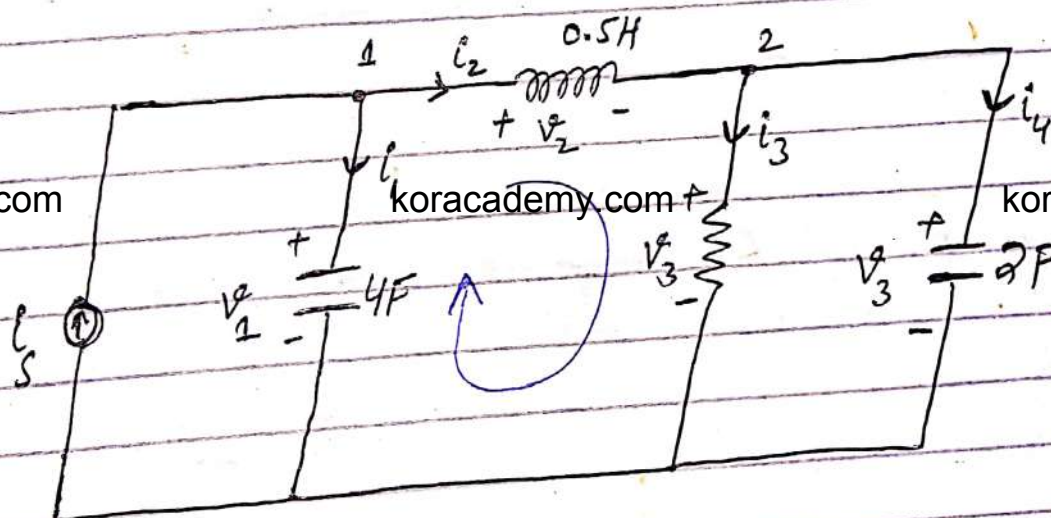
$$(b) I_1 = \frac{V}{2} = \frac{9.14}{2} = 4.57 \text{ A}$$

$$I_2 = \frac{V}{4} = \frac{9.14}{4} = 2.285 \text{ A}$$

$$I_3 = \frac{V}{8} = \frac{9.14}{8} = 1.1425$$

$$(c) P_B = V \cdot I_B = 9.14 \times 5 = 45.75 \text{ Watts.}$$

Q29.



$$V_3 = 5e^{-t} \text{ volts}$$

$$i_4 = C \frac{dV_3}{dt} = -10e^{-t} \text{ A}$$

$$i_3 = \frac{V_3}{R} = 15e^{-t} \text{ A}$$

Apply KCL to Node 2.

$$i_2 = i_3 + i_4 = 5e^{-t} \text{ A}$$

$$V_2 = L \frac{di_2}{dt} = -2.5e^{-t} \text{ volts}$$

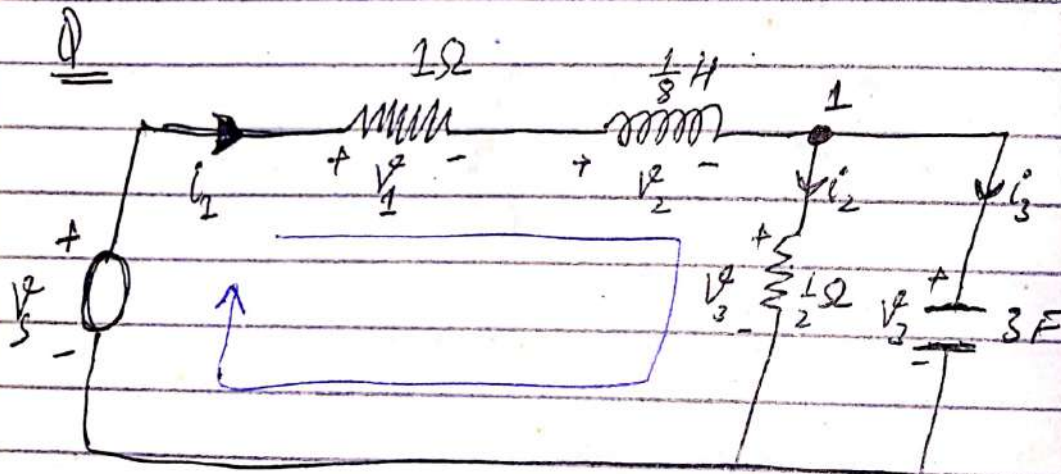
Let we suppose a loop current (which is actually not) and apply KVL.

$$V_1 = V_2 + V_3 = 2.5e^{-t} \text{ volts}$$

$$i_1 = C \frac{dV_1}{dt} = -10e^{-t} \text{ Amp.}$$

Apply KCL to node 1.

$$i_3 = i_2 + i_1 = -5e^{-t} \text{ Amp.}$$



$$i_2 = -20e^{-2t} \text{ Amp}$$

$$V_3 = i_2 R = -10e^{-2t} \text{ volts}$$

$$i_3 = C \frac{dV_3}{dt} = 60e^{-2t} \text{ Amp}$$

Apply KCL to node

$$i_1 = i_2 + i_3 = 40 e^{-2t} \text{ Amp}$$

$$V_1 = i_1 R = 40 e^{-2t} \text{ Volts}$$

$$V_2 = L \frac{di_1}{dt} = -10 e^{-2t} \text{ Volts.}$$

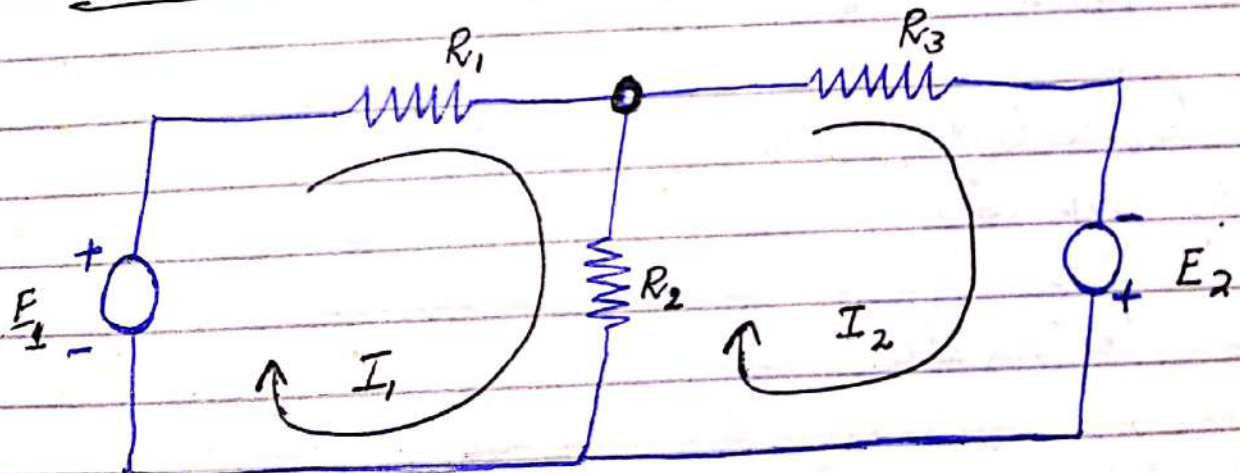
Apply KVL to the loop.

$$V_s = V_1 + V_2 + V_3$$

$$= 40 e^{-2t} - 10 e^{-2t} - 10 e^{-2t}$$

$$V_s = \underline{20 e^{-2t} \text{ Volts.}}$$

### Standard Loop Equations:



The current through  $R_2$  is  $I_1 - I_2$  whenever we consider it a member of loop 1. and is  $I_2 - I_1$ , when it is considered a member of loop 2.



Apply KVL to loop 1.

$$\sum \text{voltage rise} = \sum \text{voltage drop}$$

$$E_1 = I_1 R_1 + (I_1 - I_2) R_2$$

$$E_1 = I_1 R_1 + I_1 R_2 - I_2 R_2$$

$$E_1 = (R_1 + R_2) I_1 + (-R_2) I_2$$

$R_{11} = R_1 + R_2 \rightarrow$  The total self resistance of loop 1.

$$R_{12} = (-R_2)$$

$$R_{11} I_1 + R_{12} I_2 = E_1 \rightarrow \text{eq (A)}$$

Apply KVL to loop 2.

$$E_2 = I_2 R_3 + (I_2 - I_1) R_2$$

$$E_2 = I_2 R_3 + I_2 R_2 - I_1 R_2$$

$$E_2 = (R_3 + R_2) I_2 + (-R_2) I_1$$

$R_{22} = R_3 + R_2 \rightarrow$  the total self resistance of loop 2.

$$R_{21} = (-R_2)$$

$$R_{21} I_1 + R_{22} I_2 = E_2 \rightarrow \text{eq (B)}$$

Eq (A) and Eq (B) are the standard loop equations.

The no. of standard loop equations depends upon the number of loops in the circuit.

For 3 loops:

$$R_{11}I_1 + R_{12}I_2 + R_{13}I_3 = E_1$$

$$R_{21}I_1 + R_{22}I_2 + R_{23}I_3 = E_2$$

$$R_{31}I_1 + R_{32}I_2 + R_{33}I_3 = E_3$$

For n loops:

$$R_{11}I_1 + R_{12}I_2 + R_{13}I_3 + \dots + R_{1n}I_n = E_1$$

$$R_{21}I_1 + R_{22}I_2 + R_{23}I_3 + \dots + R_{2n}I_n = E_2$$

$$R_{31}I_1 + R_{32}I_2 + R_{33}I_3 + \dots + R_{3n}I_n = E_3$$



$$R_{n1}I_1 + R_{n2}I_2 + R_{n3}I_3 + \dots + R_{nn}I_n = E_n$$

Eq(A) and Eq (B) in matrix form;

$$\begin{bmatrix} R_{11} & R_{12} \\ R_{21} & R_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} E_1 \\ E_2 \end{bmatrix}$$

$$[R][I] = [V].$$

$$\begin{aligned} \text{Size of } R \text{ matrix} &= 2 \times 2 \rightarrow 2 \text{ loops.} \\ &= 3 \times 3 \rightarrow 3 \text{ loops} \\ &\vdots \end{aligned}$$

All the diagonal elements of R matrix will be positive.

The off diagonal elements may either be positive or negative depending on the direction of current.

If the loop currents are in opposite direction  $\rightarrow$  then negative.  
 If in the same direction  $\rightarrow$  positive.

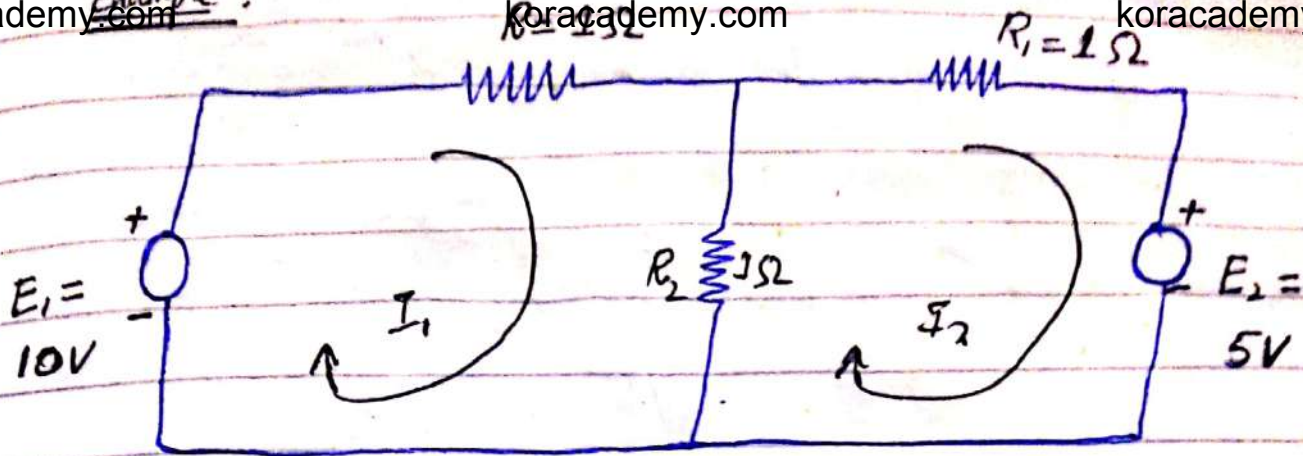
-ve to +ve  $\rightarrow$  voltage rise  $\rightarrow$   
 $\rightarrow$  + sign  $\rightarrow$  with E.

+ve to -ve  $\rightarrow$  voltage drop  $\rightarrow$  -ve  
 sign with E.

Cramer's Rule for finding current.

$$I_1 = \frac{\begin{vmatrix} E_1 & R_{12} \\ E_2 & R_{22} \end{vmatrix}}{\begin{vmatrix} R_{11} & R_{12} \\ R_{21} & R_{22} \end{vmatrix}}$$

$$I_2 = \frac{\begin{vmatrix} R_{11} & E_1 \\ R_{21} & E_2 \end{vmatrix}}{\begin{vmatrix} R_{11} & R_{12} \\ R_{21} & R_{22} \end{vmatrix}}$$



$$R_{11} = 1 + 1 = 2 \Omega$$

$$R_{12} = R_{21} = -R_2 = -1 \Omega$$

$$R_{22} = 1 + 1 = 2 \Omega$$

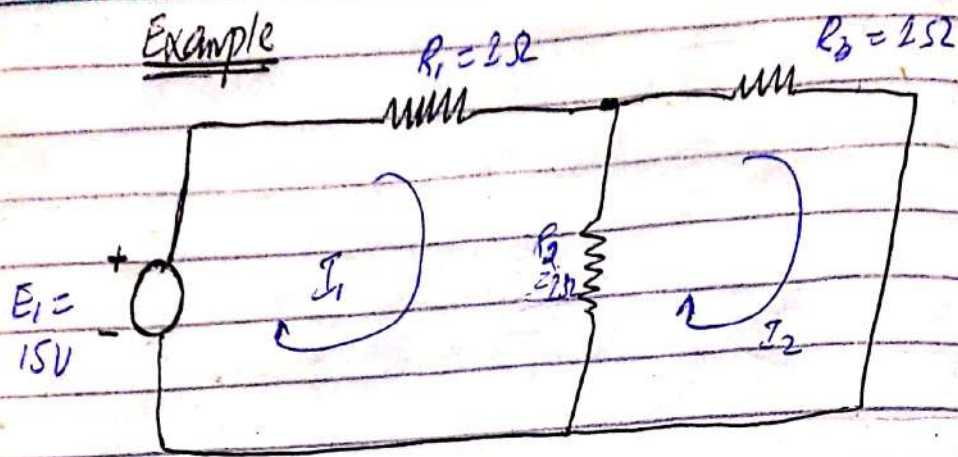
$E_1 \rightarrow$  voltage rise  $\rightarrow 10V$   
 $E_2 \rightarrow$  voltage drop  $\rightarrow -5V$

$$I_1 = \frac{\begin{vmatrix} E_1 & R_{12} \\ E_2 & R_{22} \end{vmatrix}}{\begin{vmatrix} R_{11} & R_{12} \\ R_{21} & R_{22} \end{vmatrix}} = \frac{\begin{vmatrix} 10 & -1 \\ -5 & 2 \end{vmatrix}}{\begin{vmatrix} 2 & -1 \\ -1 & 2 \end{vmatrix}} = \frac{15}{3} = \underline{\underline{5A}}$$

In matrix form

$$\begin{bmatrix} 2 & -1 \\ -1 & 2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} E_1 = 10 \\ -5 \end{bmatrix}$$

$$I_2 = \frac{\begin{vmatrix} R_{11} & E_1 \\ R_{21} & E_2 \end{vmatrix}}{\begin{vmatrix} R_{11} & R_{12} \\ R_{21} & R_{22} \end{vmatrix}} = \frac{\begin{vmatrix} 2 & 10 \\ -1 & -5 \end{vmatrix}}{3} = \underline{\underline{0A}}$$

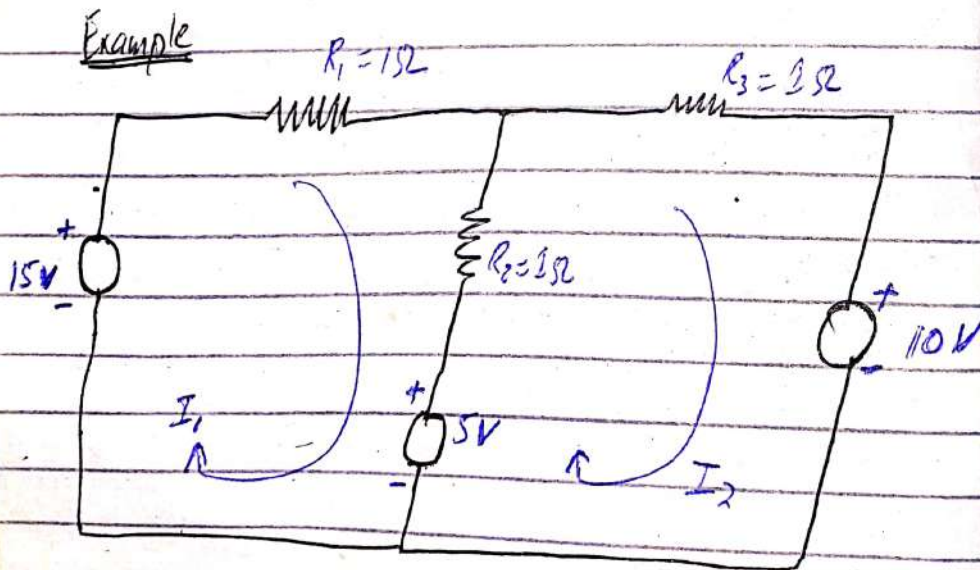


In matrix form:

$$\begin{bmatrix} 2 & -1 \\ -1 & 2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 15 \\ 0 \end{bmatrix}$$

$$I_1 = \frac{\begin{vmatrix} 15 & -1 \\ 0 & 2 \end{vmatrix}}{\begin{vmatrix} 2 & -1 \\ -1 & 2 \end{vmatrix}} = \frac{30}{3} = \underline{\underline{10A}}$$

$$I_2 = \frac{\begin{vmatrix} 2 & 15 \\ -1 & 0 \end{vmatrix}}{3} = \frac{15}{3} = \underline{\underline{5A}}$$



$E_1 \rightarrow$  total V in ~~loop~~ loop 1  
 $E_2 \rightarrow$  total V in loop 2

$$E_1 = 15 - 5 = 10V$$

$$E_2 = -10 + 5 = -5V$$

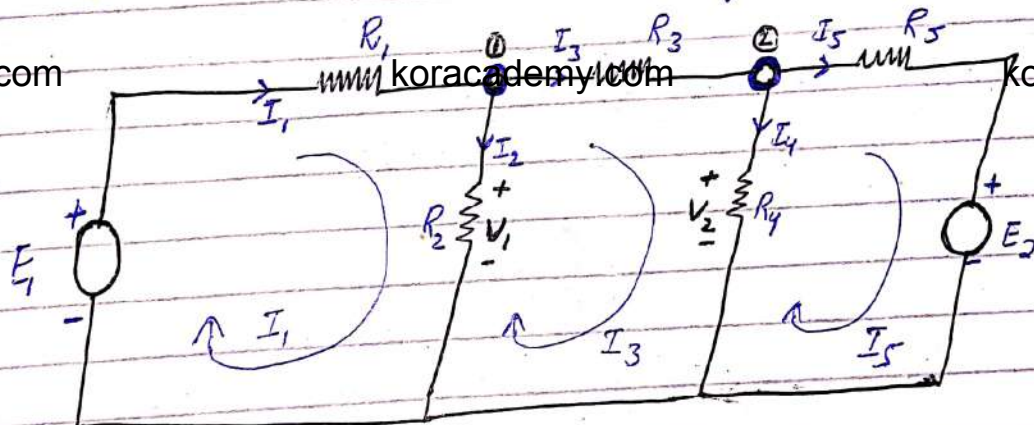
In matrix form

$$\begin{bmatrix} 2 & -1 \\ -1 & 2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 10 \\ -5 \end{bmatrix}$$

$$\therefore \text{Example 2} \quad I_1 = 5A \quad I_2 = 0A$$

### Standard Node Equations:

Consider a circuit having 2 nodes.



Apply KCL to node 1.

$$I_1 = I_2 + I_3 \rightarrow \text{eq (i)}$$

Apply KCL to node 2.

$$I_3 = I_4 + I_5 \rightarrow \text{eq (ii)}$$

Apply KVL to loop 1.

$$E_1 = I_1 R_1 + V_1$$

$$I_1 = \frac{E_1 - V_1}{R_1}$$

$$I_2 = \frac{V_2}{R_2}$$

Apply KVL to loop 2.

$$V_1 = I_3 R_3 + V_2$$

$$I_3 = \frac{V_1 - V_2}{R_3}$$

$$I_4 = \frac{V_2}{R_4}$$

Apply KVL to  $V_3$

$$V_2 = I_5 R_5 + E_2$$

$$I_5 = \frac{V_2 - E_2}{R_5}$$

$$\text{eq(i)} \Rightarrow \frac{E_1 - V_1}{R_1} = \frac{V_1}{R_2} + \frac{V_1 - V_2}{R_3}$$

$$\frac{E_1}{R_1} = \left( \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} \right) V_1 + \left( \frac{-1}{R_3} \right) V_2$$

$\frac{E_1}{R_1}$  = voltage  $\rightarrow$  gives unit of current  $\rightarrow$   
 $R_1$  resistance replace by  $I_{G1} \rightarrow$  it is  
 assumed  $\rightarrow$  it does not exist in any part  
 of the circuit.

$$\text{conductance } G = \frac{1}{R} \quad \text{U (mho)}$$

$$I_{g1} = (G_1 + G_2 + G_3)V_1 + (-G_3)V_2$$

The conductance of all resistors that are connected to node 1.

$$G_{11} = G_1 + G_2 + G_3$$

$G_3 \rightarrow$  common resistor  $\rightarrow G_{12}$

$$\boxed{G_{11}V_1 + G_{12}V_2 = I_{g1}} \rightarrow \text{eq (A)}$$

$$\text{eq (ii)} \Rightarrow \frac{V_1 - V_2}{R_3} = \frac{V_2}{R_4} + \frac{V_2 - E_2}{R_5}$$

$$\frac{E_2}{R_5} = \left(\frac{-1}{R_3}\right)V_1 + \left(\frac{1}{R_3} + \frac{1}{R_4} + \frac{1}{R_5}\right)V_2$$

$$I_{g2} = (-G_3)V_1 + (G_3 + G_4 + G_5)V_2$$

$G_{21} \rightarrow$  common resistor  $\rightarrow G_3$

Total conductance in loop 2.

$$G_{22} = G_3 + G_4 + G_5$$

$$\boxed{I_{g2} = G_{21}V_1 + G_{22}V_2} \rightarrow \text{eq (B)}$$

The number of node equations is equal to the number of nodes.



For 3 nodes

$$\begin{aligned} G_{11}V_1 + G_{12}V_2 + G_{13}V_3 &= I_1 \\ G_{21}V_1 + G_{22}V_2 + G_{23}V_3 &= I_2 \\ G_{31}V_1 + G_{32}V_2 + G_{33}V_3 &= I_3 \end{aligned}$$

For n nodes

$$\begin{aligned} G_{11}V_1 + G_{12}V_2 + G_{13}V_3 + \dots + G_{1n}V_n &= I_1 \\ G_{21}V_1 + G_{22}V_2 + G_{23}V_3 + \dots + G_{2n}V_n &= I_2 \\ G_{31}V_1 + G_{32}V_2 + G_{33}V_3 + \dots + G_{3n}V_n &= I_3 \end{aligned}$$

$$\boxed{G_{11}V_1 + G_{12}V_2 + G_{13}V_3 + \dots + G_{1n}V_n = I_1}$$

$$\boxed{G_{21}V_1 + G_{22}V_2 + G_{23}V_3 + \dots + G_{2n}V_n = I_2}$$

$$\boxed{G_{31}V_1 + G_{32}V_2 + G_{33}V_3 + \dots + G_{3n}V_n = I_3}$$

In matrix form;

$$\begin{bmatrix} G_{11} & G_{12} \\ G_{21} & G_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

$$[G][V] = [I]$$

Size of G matrix depends on number of nodes. The diagonal elements of G is always positive. The off diagonal elements will always be negative.

$$I_{s1} = \frac{E_1}{R_1} = \frac{\text{voltage of voltage source connected to node 1}}{\text{the resistance of series resistor}}$$

for standard node equation; there must be a series resistor s/w source and node.

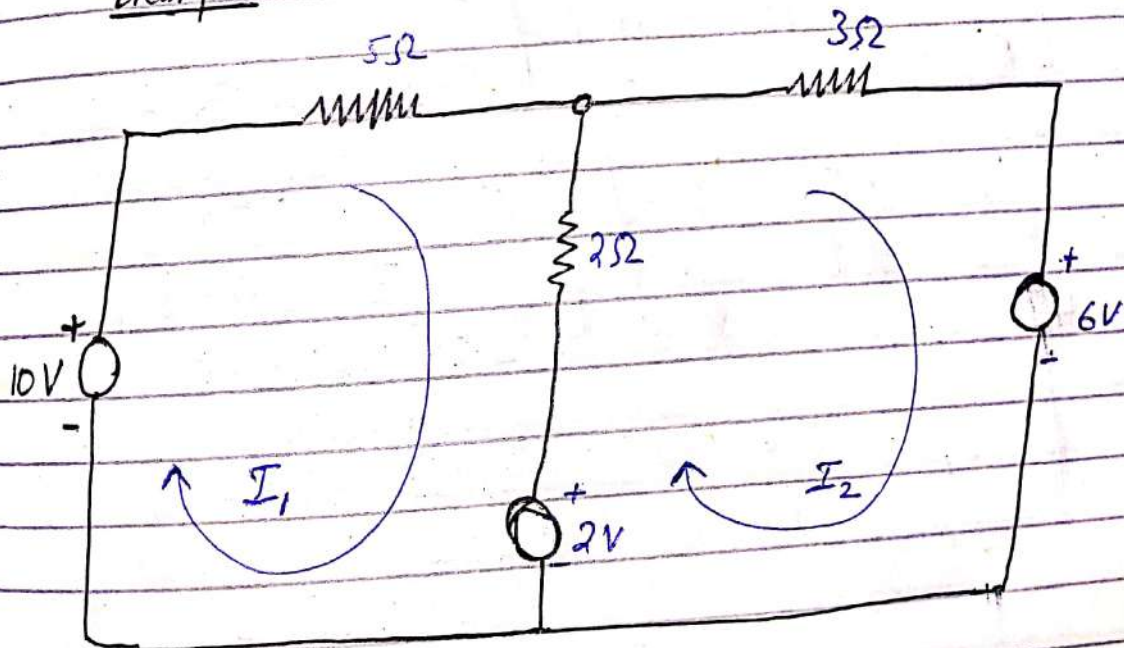
$$I_{s2} = \frac{E_2}{R_s}$$

$E_2$  and second node are connected with the help of series resistor  $R_s$ .

The + and - sign with  $I_{s1}$  and  $I_{s2}$  depends on the polarities.

if + polarity towards node  $\rightarrow$  + sign  
 if - polarity towards node  $\rightarrow$  - sign.

### Example

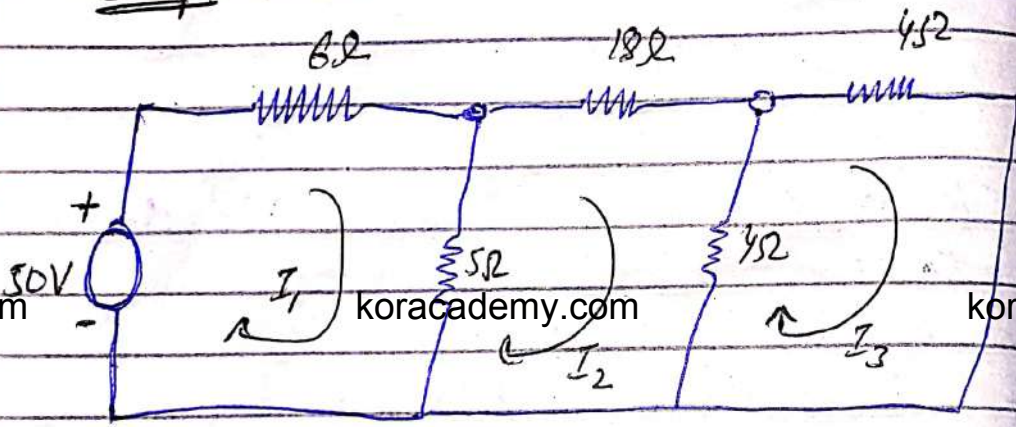


$$\begin{bmatrix} 7 & -2 \\ -2 & 5 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 8 \\ -4 \end{bmatrix}$$

$$I_1 = \frac{\begin{vmatrix} 8 & -2 \\ -4 & 5 \end{vmatrix}}{\begin{vmatrix} 7 & -2 \\ -2 & 5 \end{vmatrix}} = \frac{32}{31} = 1.032 \text{ A}$$

$$I_2 = \frac{\begin{vmatrix} 7 & 8 \\ -2 & -4 \end{vmatrix}}{31} = \frac{-12}{31} = -0.387 \text{ A}$$

Example



$$\begin{bmatrix} R_{11} & R_{12} & R_{13} \\ R_{21} & R_{22} & R_{23} \\ R_{31} & R_{32} & R_{33} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} E_1 \\ E_2 \\ E_3 \end{bmatrix}$$

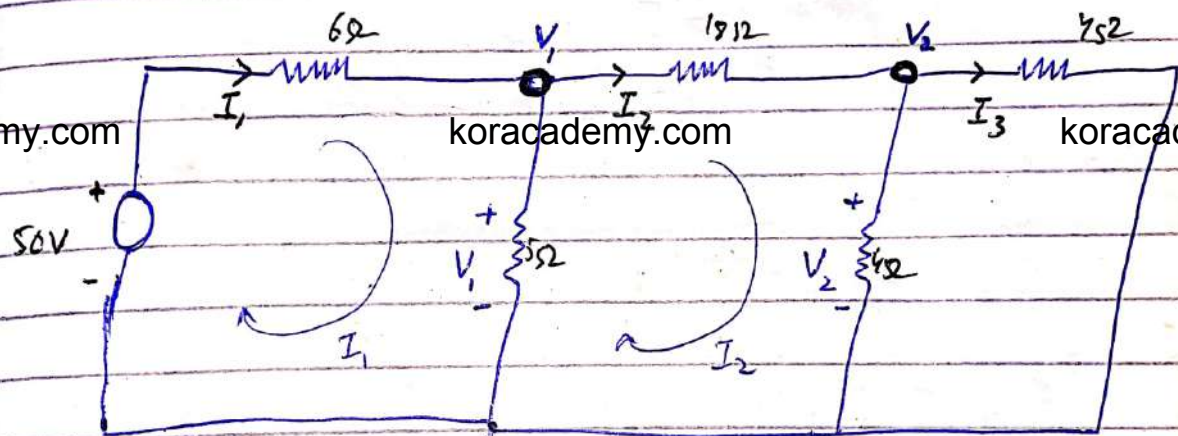
$$\begin{bmatrix} 11 & -5 & 0 \\ -5 & 27 & -4 \\ 0 & -4 & 8 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 50 \\ 0 \\ 0 \end{bmatrix}$$

$$I_1 = \frac{\begin{vmatrix} 50 & -5 & 0 \\ 0 & 27 & -4 \\ 0 & -4 & 8 \end{vmatrix}}{\begin{vmatrix} 11 & -5 & 0 \\ -5 & 27 & -4 \\ 0 & -4 & 8 \end{vmatrix}} = \frac{19000}{8000} = 5 \text{ A}$$

$$I_2 = \frac{\begin{vmatrix} 11 & 50 & 0 \\ -5 & 0 & -4 \\ 0 & 0 & 8 \end{vmatrix}}{2,000} = \frac{2,000}{2,000} = 1 \text{ A}$$

$$I_3 = \frac{\begin{vmatrix} 11 & -5 & 50 \\ -5 & 27 & 0 \\ 0 & -4 & 0 \end{vmatrix}}{2,000} = \frac{1,000}{2,000} = 0.5 \text{ A}$$

The same example with standard node equations.



$$\begin{bmatrix} G_{11} & G_{12} \\ G_{21} & G_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} I_{g1} \\ I_{g2} \end{bmatrix}$$

$$\begin{bmatrix} \frac{1}{6} + \frac{1}{5} + \frac{1}{18} & -\frac{1}{18} \\ -\frac{1}{18} & \frac{1}{18} + \frac{1}{4} + \frac{1}{4} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} 50/6 \\ 0 \end{bmatrix}$$

$$\begin{bmatrix} 0.422 & -0.0555 \\ -0.0555 & 0.555 \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} 8.33 \\ 0 \end{bmatrix}$$

Apply Cramer's rule to calculate  $V_1$  and  $V_2$ .

$$V_1 = \begin{vmatrix} 8.33 & -0.0555 \\ 0 & 0.555 \end{vmatrix} \\ \hline \begin{vmatrix} 0.422 & -0.0555 \\ -0.0555 & 0.555 \end{vmatrix} = 20 \text{ Volts}$$

$$V_2 = \begin{vmatrix} 0.422 & 8.33 \\ -0.0555 & 0 \end{vmatrix} = 2 \text{ Volts.} \\ \hline \begin{vmatrix} 0.422 & -0.0555 \\ -0.0555 & 0.555 \end{vmatrix}$$

Apply KVL to loop 1

$$50 = 6I_1$$

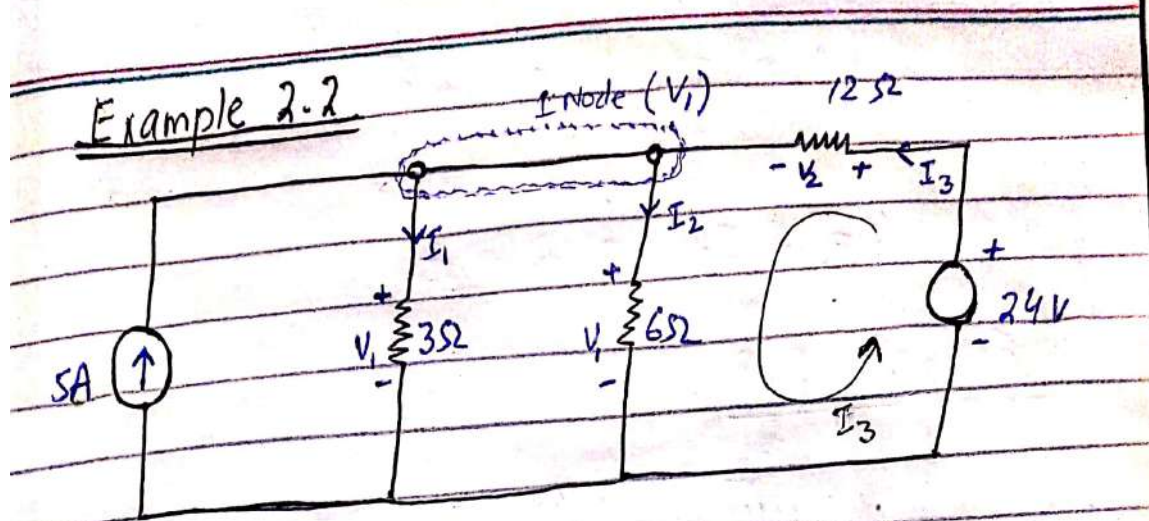
$$I_1 = \frac{50 - V_1}{6} = \frac{50 - 20}{6}$$

$$\boxed{I_1 = 5 \text{ A}}$$

$$\text{KVL to loop 2} \Rightarrow V_1 = 18I_2 + V_2$$

$$2 - 20 = 18I_2 \Rightarrow \boxed{I_2 = 1 \text{ A}}$$

$$I_3 = \frac{V_2}{4} = \frac{2}{4} \Rightarrow \boxed{I_3 = 0.5 \text{ A}}$$



$$[G_{11}] [V_1] = [I_{g1}]$$

$G_{11}$  → the conductance of resistors connected to node 1.

If current is moving towards junction will be positive and vice versa. (for current source).

$$I_{g1} = 5 + \frac{24}{12} = 7A$$

$$\left[ \frac{1}{3} + \frac{1}{6} + \frac{1}{12} \right] [V_1] = 7$$

$$V_1 = 12V$$

$$I_1 = \frac{V_1}{3} = \frac{12}{3} \Rightarrow I_1 = 4A$$

$$I_2 = \frac{V_1}{6} = \frac{12}{6} \Rightarrow I_2 = 2A$$

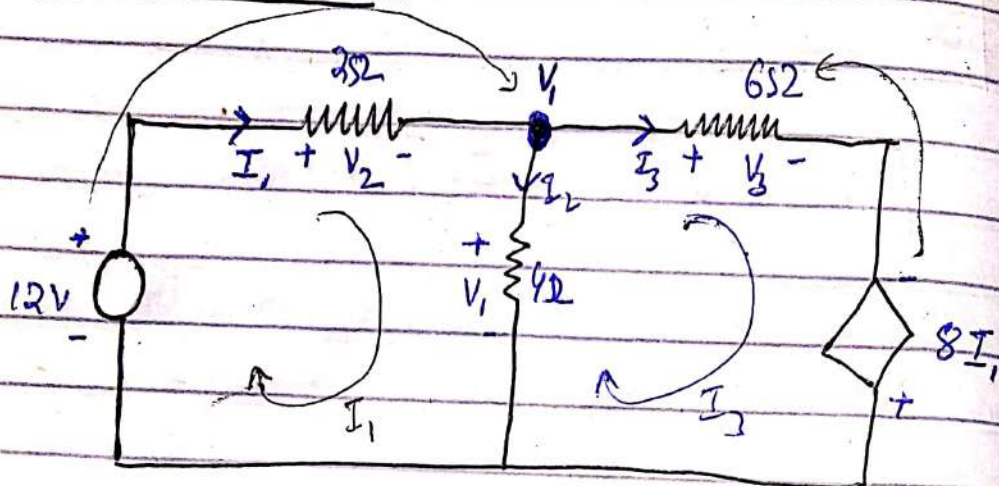
For calculating  $I_3$ , apply KVL to loop 3

$$24 = 12I_3 + V_1$$

$$I_3 = \frac{24 - V_1}{12} = 1 \text{ A}$$

$$V_2 = 12 I_3 = 12 \text{ volts.}$$

Example 2.3 :



$$[G_{ii}] [V_i] = [I_i]$$

$$\left[ \frac{1}{2} + \frac{1}{4} + \frac{1}{6} \right] [V_1] = \left[ \frac{12}{2} - \frac{8I_1}{6} \right]$$

$$\left[ \frac{1}{2} + \frac{1}{4} + \frac{1}{6} \right] [V_1] = \left[ 6 - \frac{8}{6} I_1 \right] \rightarrow \text{eq (1)}$$

Apply KVL to loop 1 to find  $I_1$

$$12 = 2I_1 + V_1$$

$$\Rightarrow I_1 = \frac{12 - V_1}{2} \rightarrow \text{eq (2)}$$

Put the value of  $I_1$  in Eq (1)

$$\left[ \frac{1}{2} + \frac{1}{4} + \frac{1}{6} \right] [V_1] = 6 - \frac{8}{6} \left( \frac{12 - V_1}{2} \right)$$

$$\left( \frac{1}{2} + \frac{1}{4} + \frac{1}{6} - \frac{4}{6} \right) [V_1] = -2$$

$$V_1 = -8 \text{ volts.}$$

$$\text{eq (ii)} \Rightarrow I_1 = \frac{12 + 8}{2} = 10 \text{ A}$$

$$I_2 = \frac{V_1}{4} = -2 \text{ A}$$

Apply KVL to loop 2 to find  $I_3$

$$V_1 + 8I_1 = 6I_3$$

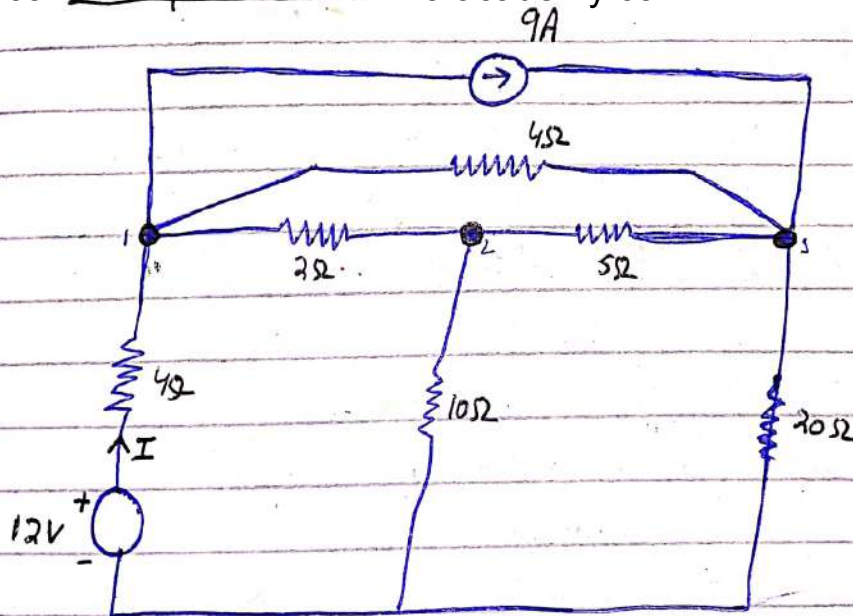
$$\frac{-8 + 80}{6} = I_3 = 12 \text{ A}$$

$$V_2 = 2I_1 = 2 \times 10 = 20 \text{ volts}$$

$$V_3 = 6I_3 = 72 \text{ volts}$$

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$$\begin{bmatrix} G_{11} & G_{12} & G_{13} \\ G_{21} & G_{22} & G_{23} \\ G_{31} & G_{32} & G_{33} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix} = \begin{bmatrix} I_{s1} \\ I_{s2} \\ I_{s3} \end{bmatrix}$$

$$G_{11} = \frac{1}{4} + \frac{1}{4} + \frac{1}{2} = 1 \text{ S}$$



$$G_{12} = G_{21} = -\frac{1}{2} = -0.5 \text{ S}$$

$$G_{13} = G_{31} = -\frac{1}{4} = -0.25 \text{ S}$$

$$G_{22} = \frac{1}{2} + \frac{1}{10} + \frac{1}{5} = 0.8 \text{ S}$$

$$G_{23} = G_{32} = -\frac{1}{5} = -0.2 \text{ S}$$

$$G_{33} = \frac{1}{4} + \frac{1}{5} + \frac{1}{20} = 0.5 \text{ S}$$

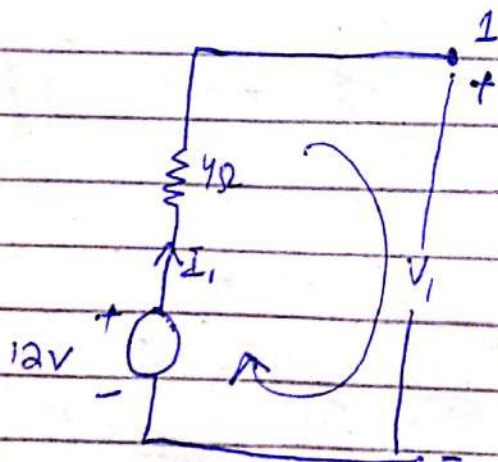
$$I_{s1} = \frac{12}{4} - 9 = -6 \text{ A}$$

$$I_{s2} = 0 \text{ A}$$

$$I_{s3} = 9 \text{ A}$$

$$\begin{bmatrix} 1 & -0.5 & -0.25 \\ -0.5 & 0.8 & -0.2 \\ -0.25 & -0.2 & 0.5 \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix} = \begin{bmatrix} -6 \\ 0 \\ 9 \end{bmatrix}$$

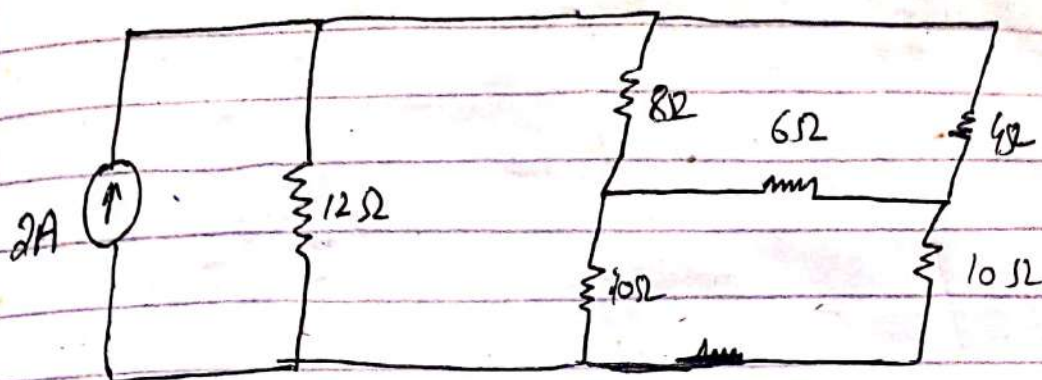
By cramer's rule  $V_1 = 4 \text{ VOLTS}$



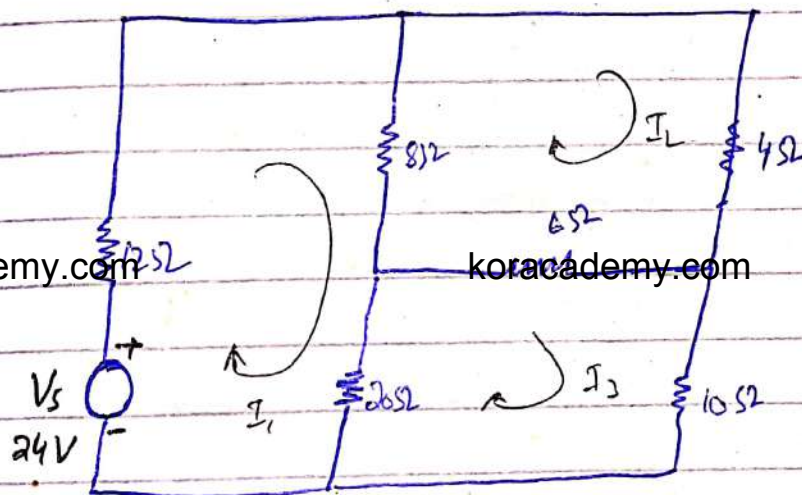
Apply KVL to this loop

$$12 = 4I + V_1$$

$$I = \frac{12 - V_1}{4} = \frac{8}{4} = 2 \text{ A}$$



We convert the current source into a voltage source.



$$\begin{bmatrix} R_{11} & R_{12} & R_{13} \\ R_{21} & R_{22} & R_{23} \\ R_{31} & R_{32} & R_{33} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} E_1 \\ E_2 \\ E_3 \end{bmatrix}$$

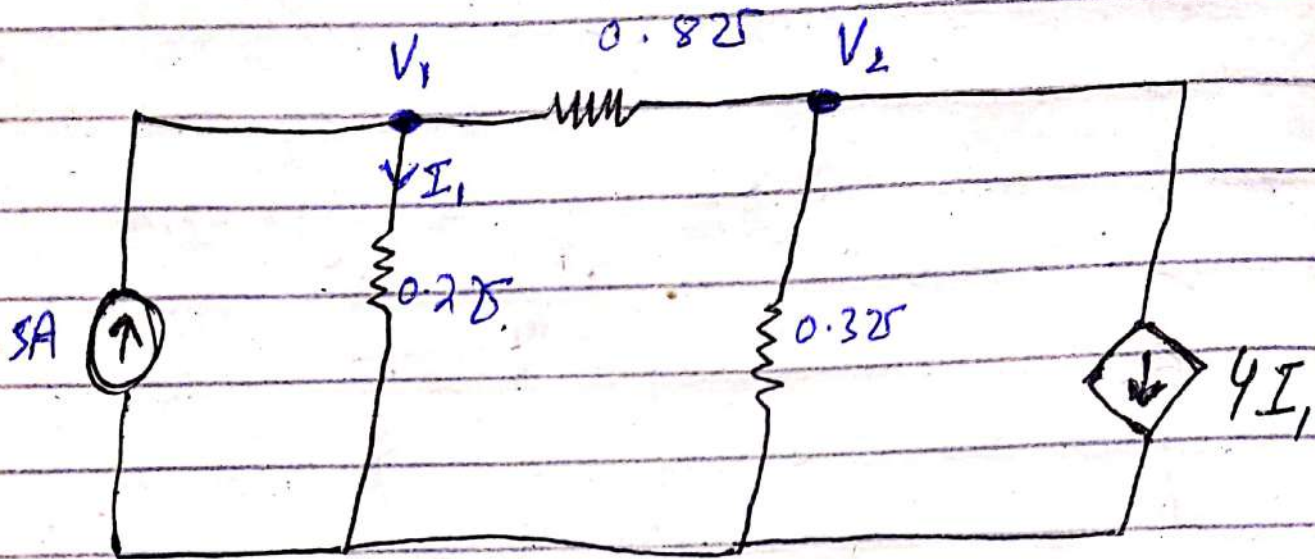
$$\begin{bmatrix} 40 & -8 & -20 \\ -8 & 18 & -6 \\ -20 & -6 & 36 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 24 \\ 0 \\ 0 \end{bmatrix}$$

By Cramer's rule.

$$I_1 = 1.125 \text{ A} \quad I_2 = 0.75 \text{ A} \quad I_3 = 0.75 \text{ A}$$

Example 2.7.

$$V_{0.8V} = ?$$



$$\begin{bmatrix} G_{11} = 1 & G_{12} = -0.8 \\ G_{21} = -0.8 & G_{22} = 1.1 \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} I_{s1} = 5 \\ I_{s2} = -4I_1 \end{bmatrix}$$

$\rightarrow = -0.8V_1$

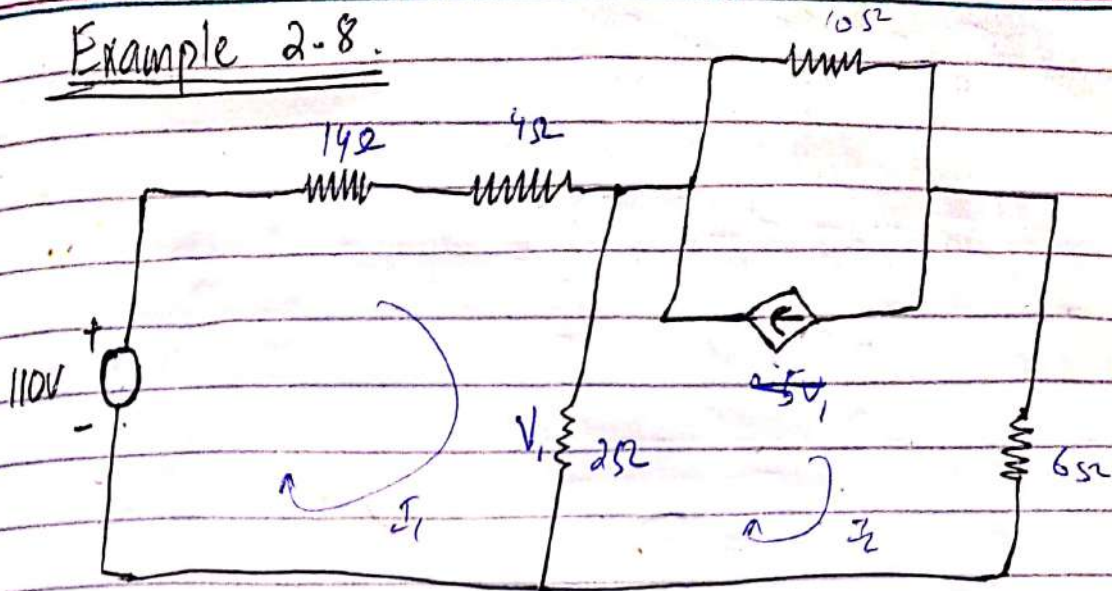
$$I_1 = \frac{V_1}{R} = 9V_1 = (0.2)V_1$$

$$\begin{aligned} V_1 - 0.8V_2 &= 5 \rightarrow \textcircled{A} \\ -0.8V_1 + 1.1V_2 &= -0.8V_1 \rightarrow \textcircled{B} \end{aligned}$$

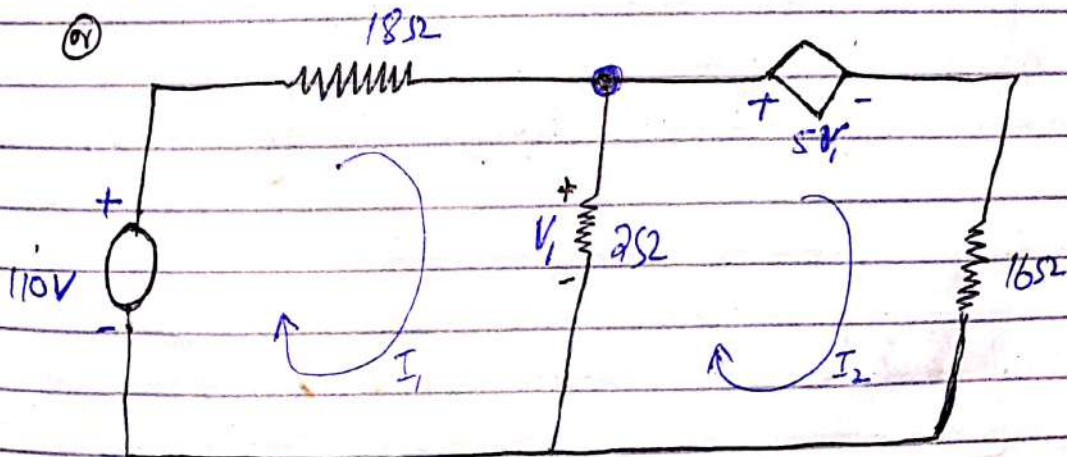
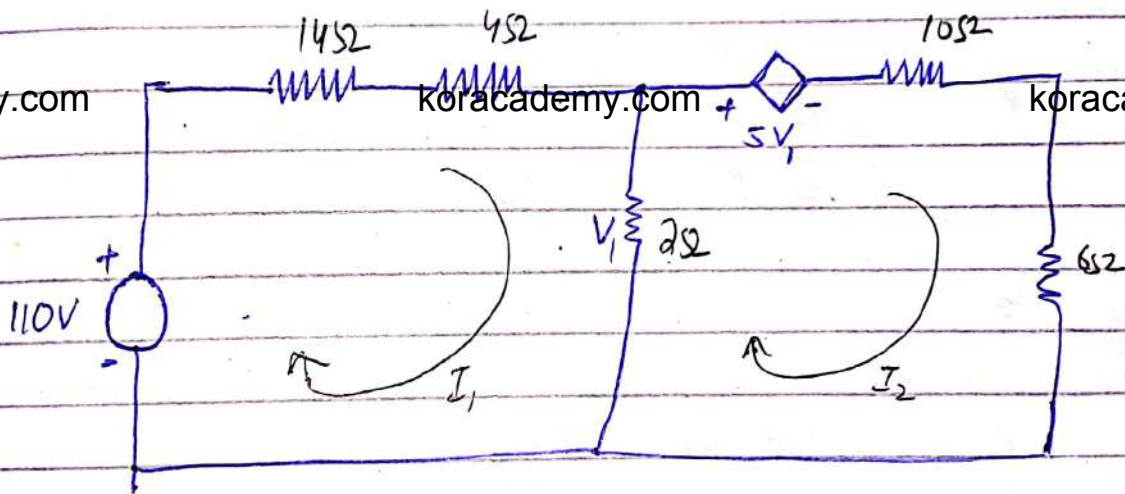
$$\begin{aligned} V_2 &= 0 \\ V_1 &= 5V \end{aligned}$$

$$V_{0.8V} = V_1 - V_2 = \underline{\underline{5V}}$$

Example 2-8.



When there is a dependent current source and a resistor is connected to it in parallel, it can be merged to a dependent voltage source.



$$[G_{ii}][V_i] = [I_{si}]$$

$$I_{S1} = \frac{110}{18} + \frac{5V_1}{16}$$

$$\left[ \frac{1}{18} + \frac{1}{2} + \frac{1}{16} \right] [V_1] = \left[ \frac{110}{18} + \frac{5V_1}{16} \right]$$

$$0.306 V_1 = \frac{110}{18}$$

$$V_1 = 20V$$

Apply KVL to loop 1.

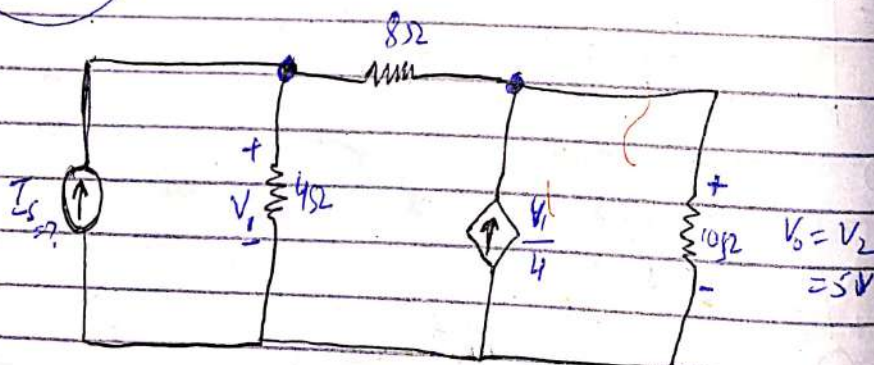
$$110 = 18I_1 + V_1$$

$$\Rightarrow I_1 = \frac{110 - 20}{18} \Rightarrow I_1 = 5A$$

Apply KVL to loop 2.

$$V_1 + 5V_1 + 16I_2 \Rightarrow I_2 = -5A$$

Q1.33



$$\begin{bmatrix} \frac{1}{4} + \frac{1}{8} & -\frac{1}{8} \\ -\frac{1}{8} & \frac{1}{8} + \frac{1}{10} \end{bmatrix} \begin{bmatrix} V_1 \\ 5 \end{bmatrix} = \begin{bmatrix} I_s \\ \frac{V_1}{4} \end{bmatrix}$$

$$\begin{bmatrix} 0.375 & -0.125 \\ -0.125 & 0.225 \end{bmatrix} \begin{bmatrix} V_1 \\ 5 \end{bmatrix} = \begin{bmatrix} I_s \\ 0.25V_1 \end{bmatrix}$$

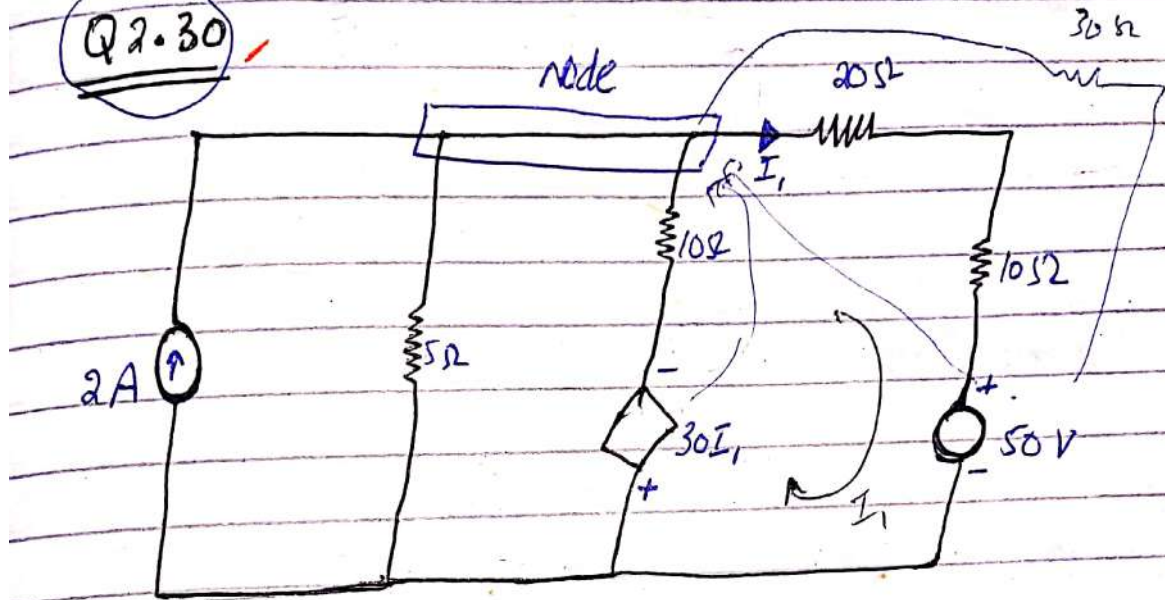
$$0.375 V, \text{ koracademy.com } I_s \rightarrow \textcircled{1}$$

$$-0.125 V, + 1.125 = 0.25 V, \rightarrow \textcircled{2}$$

$$\textcircled{2} \Rightarrow V_1 = 3 V$$

$$\textcircled{1} \Rightarrow I_s = 0.5 A$$

Q2.30



Replace the series combination of  $20\Omega$  and  $10\Omega$  by a single equivalent resistor

$$[G_{11}] [V_1] = [I_s]$$

$$\left[ \frac{1}{5} + \frac{1}{10} + \frac{1}{30} \right] [V_1] = \left[ 2 + \frac{5}{3} - 3I_1 \right]$$

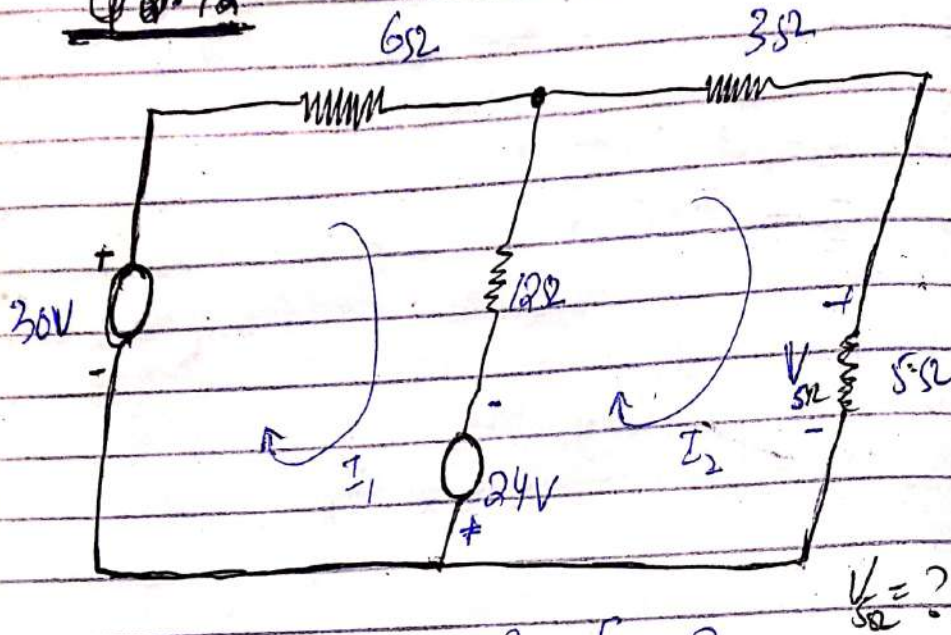
Apply KVL to loop 3

$$V_1 = 30I_1 + 50 \Rightarrow I_1 = \frac{V_1 - 50}{30}$$

$$\left[ \frac{1}{5} + \frac{1}{10} + \frac{1}{30} \right] [V_1] = \left[ 2 + \frac{5}{3} - \left( \frac{V_1 - 50}{30} \right) \right]$$

$$\Rightarrow \underline{V_1 = 20 \text{ VOLT}}$$

Q.2.12

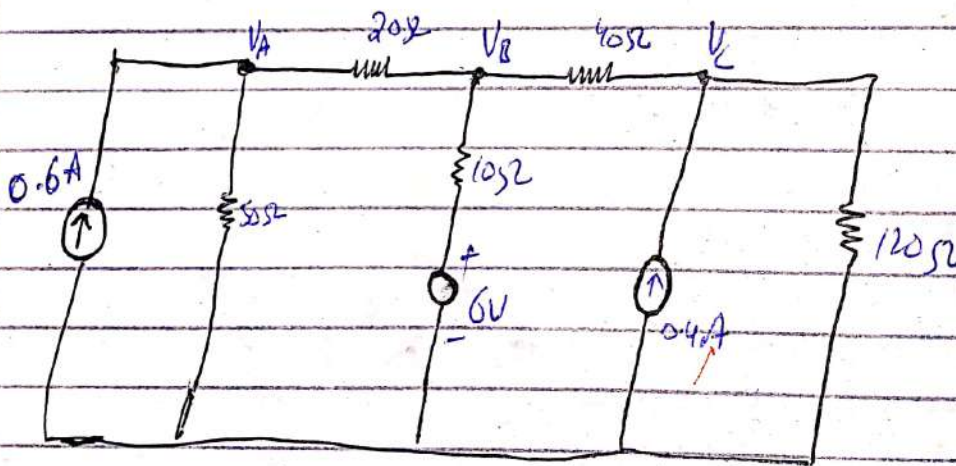


$$\begin{bmatrix} 18 & -12 \\ -12 & 20 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 54 \\ -24 \end{bmatrix}$$

$I_2 = 1A$

$V_{5\Omega} = I_2 R = 1 \times 5 = 5 \text{ V}$

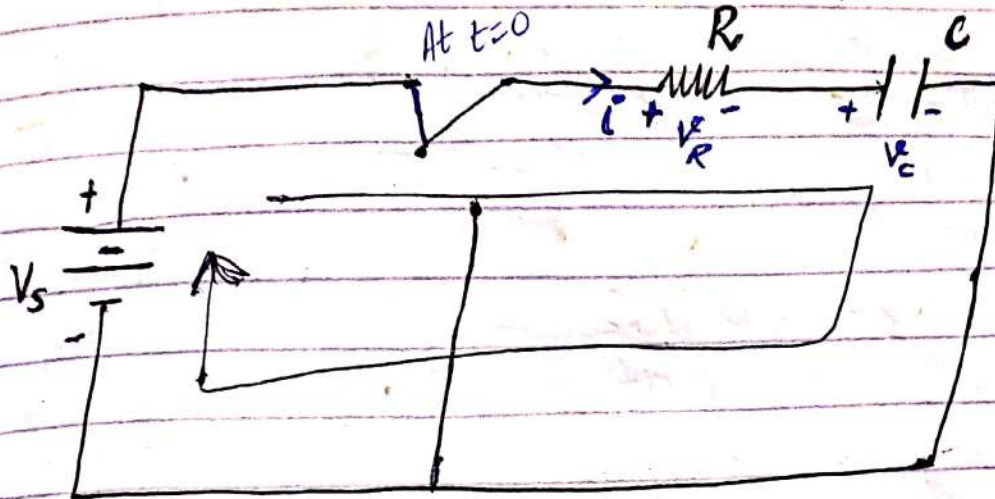
Q.2.20



$$\begin{bmatrix} 1/50 + 1/20 & -1/20 & 0 \\ -1/20 & 1/20 + 1/10 + 1/40 & -1/40 \\ 0 & -1/40 & 1/40 + 1/120 \end{bmatrix} \begin{bmatrix} V_A \\ V_B \\ V_C \end{bmatrix} = \begin{bmatrix} 0.6 \\ 0.6 \\ 0.4 \end{bmatrix}$$

$V_B = ?$   
 By mesh rule.

# Charging and Discharging of a Capacitor



When the capacitor is fully charged, then it behaves DC. Let initially there is no voltage across this capacitor.

At  $t=0$ ,

$V_C = 0$

When the switch is closed, current flows through the circuit (call it charging current). When the capacitor fully charges, there is no current in the circuit; so call it a time varying current.

Apply KVL to the only loop.

$$V_s = V_R + V_C \rightarrow \textcircled{1}$$

$$V_R = iR$$

Same current in all element (as series circuit)

$$i = C \frac{dV_C}{dt}$$



$$V_c = RC \frac{dV_c}{dt}$$

RC  $\rightarrow$  time constant =  $\tau$   
 fully charging  $5\tau - 6\tau$

$$0 \Rightarrow V_s = RC \frac{dV_c}{dt} + V_c$$

$$\int \frac{-dV_c}{V_s - V_c} = \int \frac{-dt}{RC}$$

multiply by -1  
and integrate.

$$\ln(V_s - V_c) = \frac{-t}{RC} + K \rightarrow \textcircled{2}$$

Use the initial condition i.e., At  $t=0$ ,  $V_c=0$

$$\ln(V_s) = 0 + K$$

$$\textcircled{2} \Rightarrow \ln(V_s - V_c) = \frac{-t}{RC} + \ln(V_s)$$

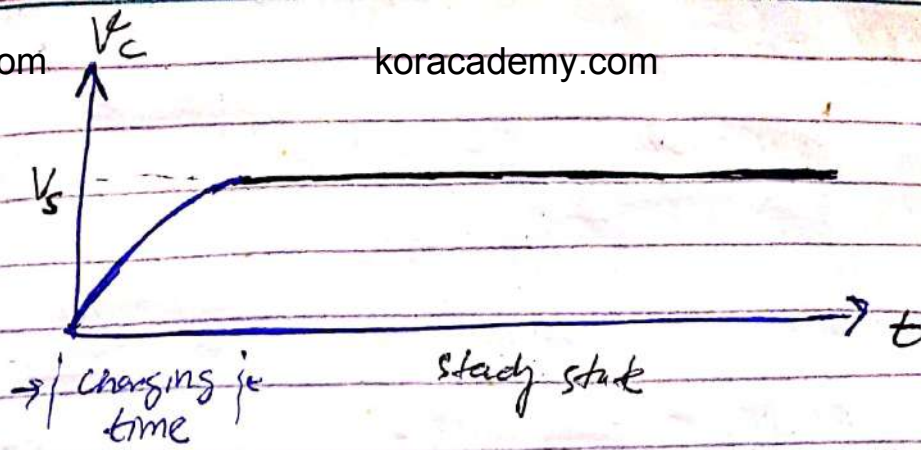
$$\ln\left(\frac{V_s - V_c}{V_s}\right) = \frac{-t}{RC}$$

Antilog

$$\frac{V_s - V_c}{V_s} = e^{-t/RC}$$

$$V_c = V_s - V_s e^{-t/RC}$$

$$V_c = V_s (1 - e^{-t/RC})$$



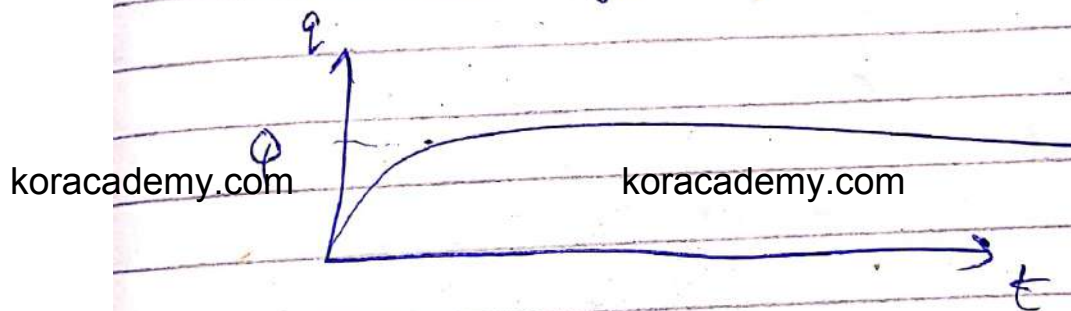
multiply by  $c$

$$cV_c = cV_s (1 - e^{-t/RC})$$

$$q = Q (1 - e^{-t/RC})$$

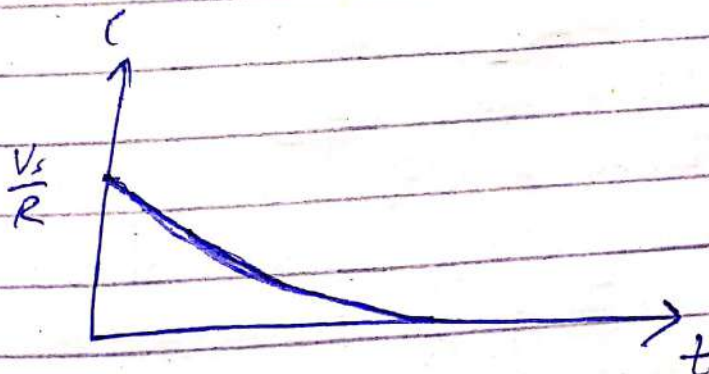
$q \rightarrow$  instantaneous charge

$Q \rightarrow$  max const charge



$$i = \frac{dq}{dt} = \frac{cV_s}{RC} e^{-t/RC}$$

$$i = \frac{V_s}{R} e^{-t/RC}$$



$$R = 1 \times 10^6 \Omega \quad C = 2 \times 10^{-6} F \quad V_s = 100V$$

- (i) The initial value of  $i$ .  
 (ii) The time constant.  
 (iii) The charging current after 2s.  
 (iv) The  $V_c$  after 2s.  
 (v) The time taken by  $C$  so that  $V_c = 80V$

$$(i) \quad i = \frac{V_s}{R} e^{-t/RC} \quad \text{At } t=0, e^0 = 1$$

$$\Rightarrow i = \frac{V_s}{R} = \frac{100}{1} \times 10^{-6} = \boxed{100 \mu A}$$

$$(ii) \quad \tau = RC = \boxed{2s}$$

$$(iii) \quad t = 2s \quad \Rightarrow i = \frac{100}{1} \times 10^{-6} \times e^{-1}$$

$$\Rightarrow \boxed{i = 36.78 \mu A}$$

$$(iv) \quad V_c = V_s (1 - e^{-t/RC}) = 100 (1 - e^{-1})$$

$$\boxed{V_c = 63.21 \text{ volts}}$$

$$(v) \quad V_c = V_s (1 - e^{-t/RC})$$

$$80 = 100 (1 - e^{-t/2})$$

$$1 - e^{-t/2} = 0.8 \quad \Rightarrow \quad e^{-t/2} = 0.2$$

Taking log

$$-\frac{t}{2} = \ln(0.2)$$

$$\downarrow$$

$$-1.6$$

$$\Rightarrow \boxed{t = 3.2s}$$

(vi) The voltage  $V_c$  after  $5\tau = 10\text{sec}$

$$V_c = 100(1 - e^{-5}) = \boxed{99.3 \text{ volts}}$$

### Example 2

$$R = 1 \times 10^6 \Omega \quad C = 2 \times 10^{-6} \text{F} \quad V_s = 100 \text{V}$$

$$\text{At } t=0, \quad V_c = V_0 = 10 \text{ volts.}$$

$$\ln(V_s - V_c) = \frac{-t}{RC} + K \rightarrow (i)$$

Initial condition,  $V_c = V_0$  at  $t=0$

$$\ln(V_s - V_0) = 0 + K$$

$$(i) \Rightarrow \ln(V_s - V_c) - \ln(V_s - V_0) = \frac{-t}{RC}$$

$$\ln\left(\frac{V_s - V_c}{V_s - V_0}\right) = \frac{-t}{RC}$$

Take Antilog

$$\frac{V_s - V_c}{V_s - V_0} = e^{-t/RC}$$

$$V_s - V_c = (V_s - V_0)e^{-t/RC}$$

$$\boxed{V_c = V_s - (V_s - V_0)e^{-t/RC}}$$

$$\boxed{C V_c = C V_s - C(V_s - V_0)e^{-t/RC}}$$

Differentiate it w.r.t  $t$

$$\boxed{i = \frac{V_s - V_0}{R} e^{-t/RC}}$$

## Example 2 solution

$$(i) \quad \tau = RC = 2 \text{ sec}$$

$$(ii) \quad i = \frac{V_s - V_0}{R} e^{-t/RC}$$

$$\text{At } t=0 \Rightarrow e^0 = 1$$

$$i = \frac{V_s - V_0}{R} = \boxed{90 \mu\text{A}}$$

(iii) Current after 2s

$$i = \frac{90}{1} \times 10^{-6} \times e^{-2/2} = \boxed{33 \mu\text{A}}$$

(iv)  $V_c$  after 2s.

$$V_c = V_s - (V_s - V_0) e^{-t/RC}$$

$$V_c = 100 - 90 e^{-1} = \boxed{66.89 \text{ V}}$$

## Phasor

Phasor is a complex quantity.  
They behave like vectors.  
They may be vector or scalar depending on the original quantity they are representing. If the original quantity is scalar, the phasor representing it will also be a scalar and hence so for vectors.

Phasor  $\rightarrow$  complex quantity  
 $\downarrow$   
 $Z = a + jb$

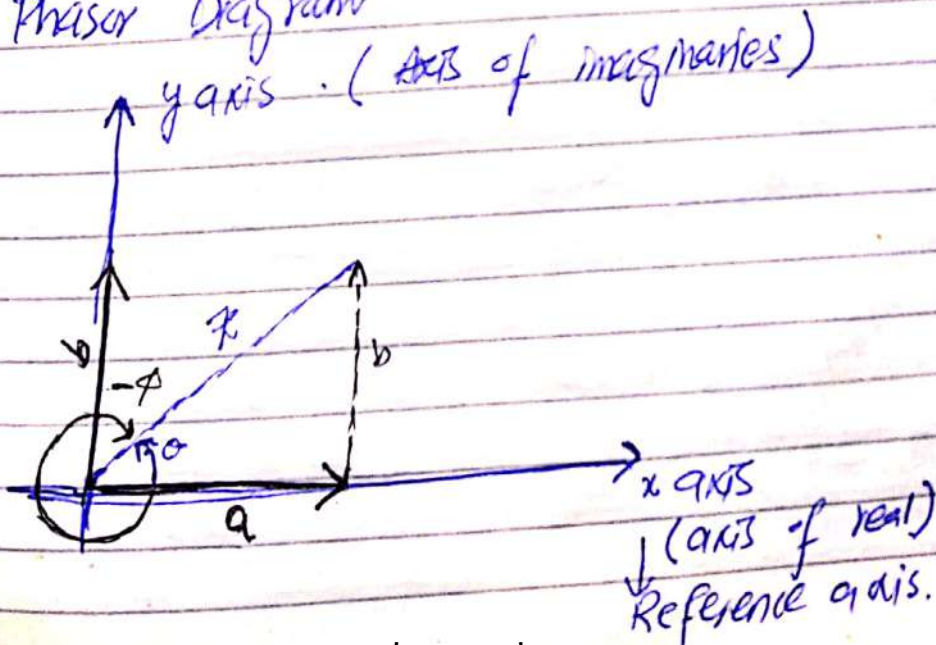
$\rightarrow$  Real component = a  
 $\rightarrow$  Imaginary component = b

$$j = i = \sqrt{-1} \rightarrow \text{operator}$$

$$j^2 = i^2 = -1$$

if the entire phasor behaves as vector quantity, both its components will behave like vectors.

### Phasor Diagram



Any phasor (vector) has two important parameters; the magnitude and direction (angle with the  $x$  axis).  
 clockwise angle  $\rightarrow$  -ve anticlockwise  $\rightarrow$  +ve

### Representation of A Phasor.

- (i) Rectangular form (ii) Trigonometric or circular form  
 (iii) Exponential form.  
 (iv) Polar form.

(i) Rectangular form

Real component and imaginary component.

$$Z = a + bj$$

(ii) Trigonometric form

$|Z|$  = magnitude of the phasor.  
 $\theta$   $\rightarrow$  Angle of the phasor.

$$a = |Z| \cos \theta \quad b = |Z| \sin \theta$$

$$Z = a + jb \Rightarrow Z = |Z| \cos \theta + j |Z| \sin \theta$$

$$Z = |Z| [\cos \theta + j \sin \theta]$$

$$|Z| = \sqrt{a^2 + b^2}$$

$$\tan \theta = \frac{b}{a} \Rightarrow \theta = \tan^{-1} \left( \frac{b}{a} \right)$$

### (iii) Exponential Form

Consider the trigonometric form.

$$z = |z| [\cos \theta + j \sin \theta]$$

From Euler's equations:

$$\cos \theta + j \sin \theta = e^{j\theta}$$

$$\cos \theta - j \sin \theta = e^{-j\theta}$$

$$\rightarrow z = |z| e^{j\theta}$$

### (iv) Polar Form

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The magnitude and angle are given independently.

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$$|z| \angle \theta$$

(this is not a product)

Example

$$z = 3 + j4 \rightarrow \text{rectangular}$$

$$|z| = 5 \quad \theta = 53.1^\circ$$

$$z = 5 [\cos 53.1^\circ + j \sin 53.1^\circ] \rightarrow \text{trigonometric}$$

$$z = 5 e^{j53.1^\circ} \rightarrow \text{exponential}$$

$$z = 5 \angle 53.1^\circ \rightarrow \text{polar.}$$



## Addition of Phasors.

$$Z_1 = |Z_1| e^{j\theta_1} = |Z_1| \angle \theta_1 = a_1 + jb_1$$

$$Z_2 = |Z_2| e^{j\theta_2} = |Z_2| \angle \theta_2 = a_2 + jb_2$$

$$\begin{aligned} Z_0 &= Z_1 + Z_2 = |Z_1| e^{j\theta_1} + |Z_2| e^{j\theta_2} \\ &= |Z_1| \angle \theta_1 + |Z_2| \angle \theta_2 \end{aligned}$$

$$\begin{aligned} Z_0 &= Z_1 + Z_2 = (a_1 + jb_1) + (a_2 + jb_2) \\ &= (a_1 + a_2) + j(b_1 + b_2) \quad \text{Rectangular form.} \end{aligned}$$

$$|Z_0| = \sqrt{(a_1 + a_2)^2 + (b_1 + b_2)^2}$$

$$\theta_0 = \tan^{-1} \frac{(b_1 + b_2)}{(a_1 + a_2)}$$

$$Z_0 = |Z_0| e^{j\theta_0} = |Z_0| \angle \theta_0$$

## Subtraction of Phasors

$$Z_5 = (a_1 - a_2) + (b_1 - b_2)j$$

$$|Z_5| = \sqrt{(a_1 - a_2)^2 + (b_1 - b_2)^2}$$

$$\theta_5 = \tan^{-1} \frac{(b_1 - b_2)}{(a_1 - a_2)}$$

$$Z_5 = |Z_5| e^{j\theta_5} = |Z_5| \angle \theta_5$$

## Multiplication of Phasors

$$Z_1 = |Z_1| e^{j\theta_1} = |Z_1| \angle \theta_1 = a_1 + jb_1$$

$$Z_2 = |Z_2| e^{j\theta_2} = |Z_2| \angle \theta_2 = a_2 + jb_2$$

$$Z_m = Z_1 \times Z_2 = (a_1 + jb_1)(a_2 + jb_2)$$

$$= (a_1 a_2 - b_1 b_2) + j(a_1 b_2 + a_2 b_1)$$

$$|Z_m| = \sqrt{(a_1 a_2 - b_1 b_2)^2 + (a_1 b_2 + a_2 b_1)^2}$$

$$\theta_m = \tan^{-1} \frac{(a_1 b_2 + a_2 b_1)}{(a_1 a_2 - b_1 b_2)}$$

$$Z_m = |Z_m| \angle \theta_m = |Z_m| e^{j\theta_m}$$

In exponential form

$$\begin{aligned} Z_m &= Z_1 \times Z_2 = |Z_1| e^{j\theta_1} \times |Z_2| e^{j\theta_2} \\ &= |Z_1| |Z_2| e^{j\theta_1 + j\theta_2} = \underbrace{|Z_1| |Z_2|}_{|Z_m|} e^{j(\theta_1 + \theta_2)} \end{aligned}$$

in polar form

$$\begin{aligned} Z_m &= |Z_1| \angle \theta_1 \times |Z_2| \angle \theta_2 \\ &= |Z_1| |Z_2| \angle \theta_1 + \theta_2 \end{aligned}$$

## Division of Phasors

$$Z_d = \frac{Z_1}{Z_2} = \frac{(a_1 + jb_1)}{(a_2 + jb_2)} \times \frac{(a_2 - jb_2)}{(a_2 - jb_2)}$$

$$= \frac{(a_1 a_2 + b_1 b_2)}{(a_2^2 + b_2^2)} + j \frac{(a_2 b_1 - a_1 b_2)}{(a_2^2 + b_2^2)}$$

In exponential form

$$Z_d = \frac{Z_1}{Z_2} = \frac{|Z_1| e^{j\theta_1}}{|Z_2| e^{j\theta_2}} = \frac{|Z_1|}{|Z_2|} e^{j\theta_1 - j\theta_2}$$

$$= \frac{|Z_1|}{|Z_2|} e^{j(\theta_1 - \theta_2)}$$

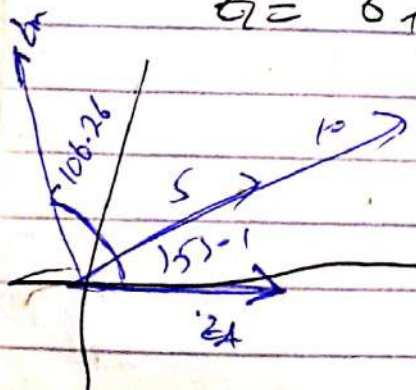
In polar form

$$Z_d = \frac{Z_1}{Z_2} = \frac{|Z_1| \angle \theta_1}{|Z_2| \angle \theta_2} = \frac{|Z_1|}{|Z_2|} \angle \theta_1 - \theta_2$$

Q1

$$Z_1 = 3 + j4 = 5 \angle 53.1^\circ$$

$$Z_2 = 6 + j8 = 10 \angle 53.1^\circ$$



$$Z_4 = Z_2 - Z_1 = 9 + j12$$

$$= 11.18 \angle 53.1^\circ$$

$$Z_s = Z_2 - Z_1 = (6 + j8) - (3 + j4) = 3 + j4$$

$$Z_s = 5 \angle 53.1^\circ$$

Q3.  $Z_m = Z_1 \times Z_2 = (3 + j4)(6 + j8)$   
 $= (18 - 32) + j(24 + 24) = -14 + j48$

$$|Z_m| = \sqrt{196 + 2304} = 50$$

$$\theta_m = \tan^{-1} \frac{48}{-14} = 73.73^\circ$$

$$\Rightarrow 180 - 73.73 \Rightarrow \theta_m = 106.26^\circ$$

$$Z_m = 50 \angle 106.26^\circ$$

In polar form;  $Z_m = 5 \angle 53.1^\circ \times 10 \angle 53.1^\circ$   
 $Z_m = 50 \angle 106.26^\circ$

Q4.  $Z_d = \frac{Z_2}{Z_1} = \frac{10 \angle 53.1^\circ}{5 \angle 53.1^\circ} = 2 \angle 0^\circ = 2$

# AC FUNDAMENTALS

## Generation of AC voltage

### Faraday's law

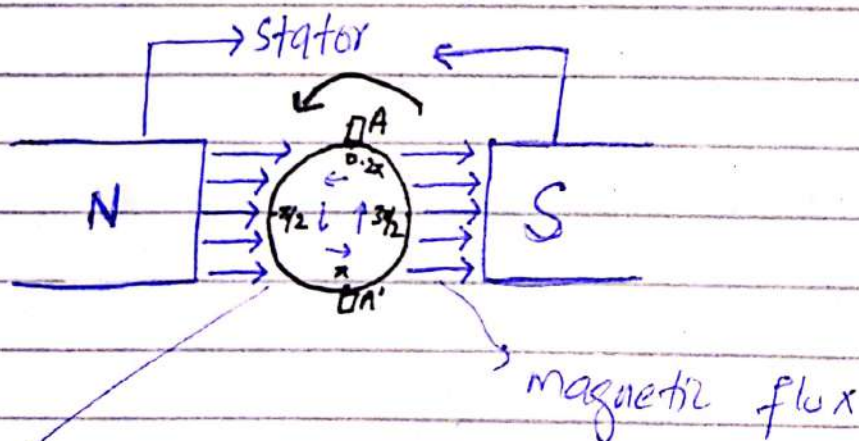
① When a conductor moves in a constant magnetic field

④ Whenever a conductor cuts the magnetic flux, voltage is induced across that conductor.

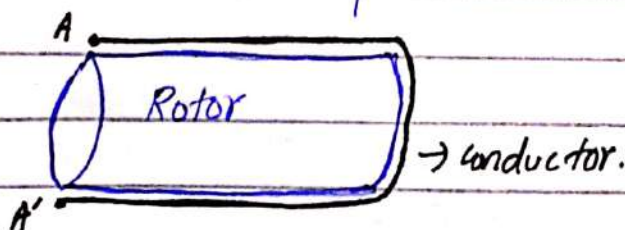
② If we place a conductor in a time varying magnetic field; voltage is induced across that conductor.

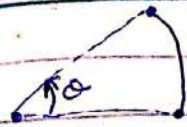
→ If we want to generate voltage we need:

a) magnetic field    b) Conductor (winding)



→ Rotor → the part that rotates.  
 Stator → The part in static condition.



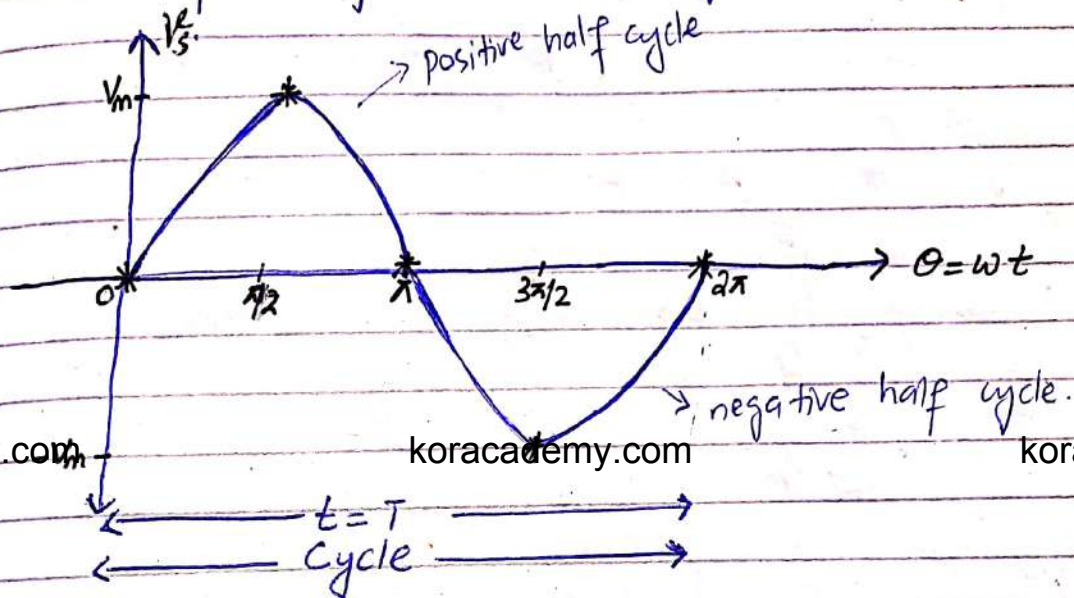


$\omega \rightarrow$  angular velocity in rad/sec.  
 $\theta \rightarrow$  angular displacement in "t" seconds.

$$\omega = \frac{\theta}{t} \rightarrow \text{rad/sec}$$

$$\theta = \omega t$$

### Waveform of the AC voltage



At  $\theta = \omega t = 0 \Rightarrow V_s = 0$ .

The direction of motion of the conductor is parallel to the flux, hence the conductor does not cut any line  $\rightarrow$  no voltage.

At  $\theta = \pi/2$  rad  $V_s \Rightarrow V_m = \text{Maximum voltage or peak value.}$

Direction of motion perpendicular to flux,  $\&$  cut maximum lines and hence maximum voltage.

At  $\theta = \omega t = \pi$  rad  $\Rightarrow V_s = 0$

Direction of motion of conductor is  $\parallel$  to flux  $\rightarrow$  does not cut any line  $\rightarrow$  no voltage

At  $\theta = 3\pi/2$   $-V_m \rightarrow$  maximum or peak value

Direction of motion  $\perp \rightarrow$  maximum lines  
So maximum voltage.

The waveform satisfies the equation;

$$V_s = V_m \sin \omega t$$

which is the instantaneous equation for the AC voltage.

$V_m \rightarrow$  maximum or peak value  $\rightarrow$  which is constant

The time taken by one complete cycle of the AC voltage is known as time period - T.

Number of cycles per second is called frequency.  $f$  (cycles/sec  $\odot$  Hz).

In Pakistan it is 50 Hz.

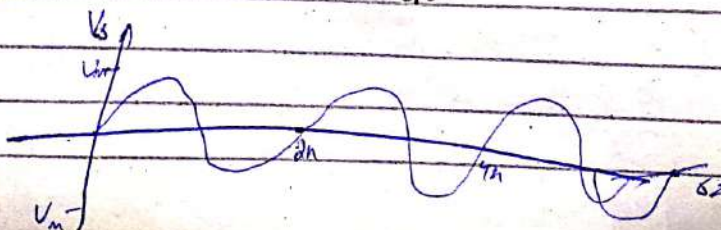
$\rightarrow$   $\odot$  The time taken by waveform to repeat itself.

In T seconds (time period) we get 1 cycle of AC voltage.

$$\text{In 1 second we get} = \frac{1 \times 1}{T} = \frac{1}{T} \text{ cycles.}$$

$$\text{mean frequency} \leftarrow f = \frac{1}{T} \text{ cycles/sec}$$

The time period of AC voltage in Pakistan is 0.02 seconds.



The cycle repeats itself at  $\omega t = 2\pi, 4\pi, 6\pi, 8\pi, \dots, 2m\pi$  where  $m = 1, 2, 3, \dots$

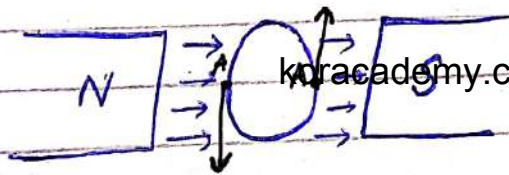
when  $m = 1$ ,  $\omega = 2\pi \Rightarrow t = T$ ,  $\omega T = 2\pi$

Angular frequency  $\leftarrow \omega = 2\pi f \leftarrow \omega = \frac{2\pi}{T}$

The  $\omega$  in the instantaneous expression is the angular frequency.  $v_s = V_m \sin \omega t$

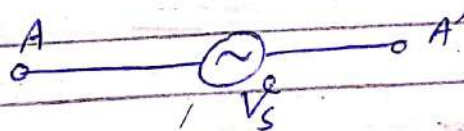
Root Mean Squared Value or Effective value of AC current.

### Fleming's Right Hand Rule

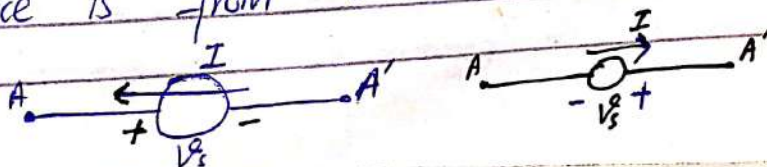


thumb  $\rightarrow$  direction of motion of conductor.  
 index  $\rightarrow$  direction of magnetic field / flux.  
 middle  $\rightarrow$  direction of induced current in the conductor.

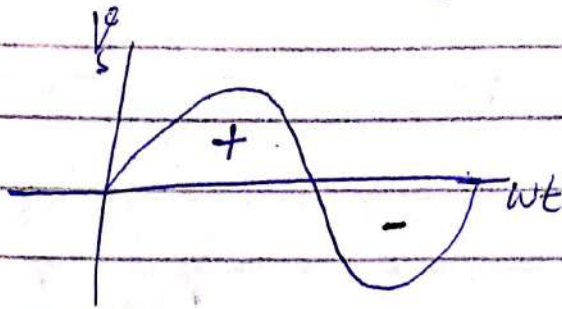
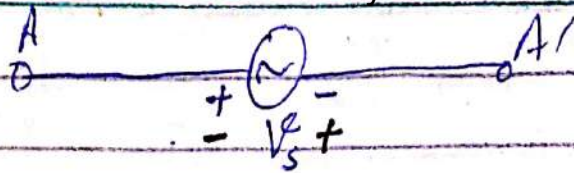
The conductor behaves as source.



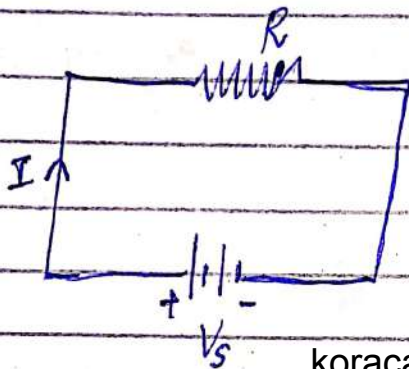
The conventional direction of current inside the source is from -ve to +ve.



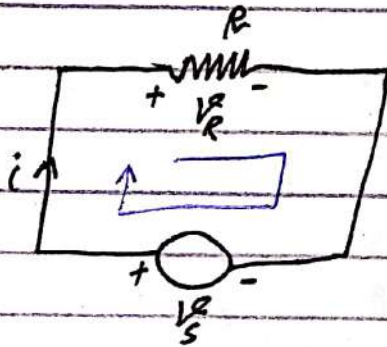




## RMS Value or Effective Value of AC



$$P_R = I^2 R$$



$V_s = V_m \sin \omega t$   
Apply KVL to this loop.

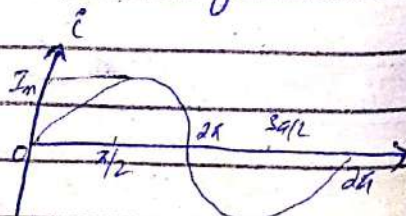
$$V_R = V_s = V_m \sin \omega t$$

$$V_s = V_R = iR$$

$$i = \frac{V_s}{R} \Rightarrow i = \frac{V_m}{R} \sin \omega t$$

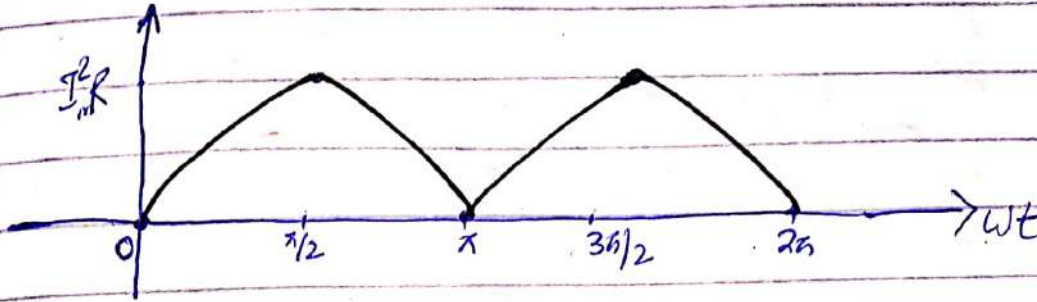
$I_m = \frac{V_m}{R} \rightarrow$  max or peak value of AC current

$$i = I_m \sin \omega t$$

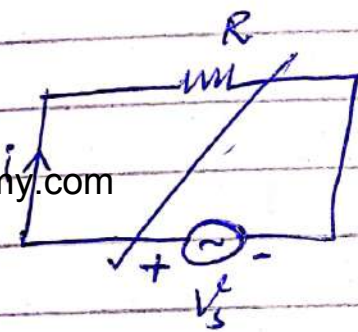


$\Rightarrow P_R = I_m^2 R \sin^2 \omega t$

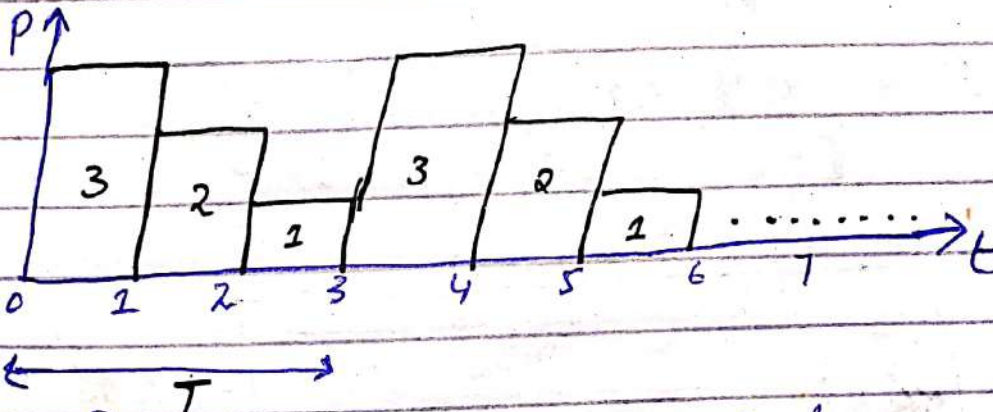
$I_m^2 R \rightarrow$  max or peak value of AC power.



RMS Value of AC current



$V_s = \frac{V_R}{R} = V_m \sin \omega t$



$P_{average} = \frac{3+2+1+3+2+1}{6} = 2 \text{ watts}$

for one cycle i.e.  $T = 3 \text{ s}$ .

$P_{av} = \frac{3+2+1}{3} = 2 \text{ Watts}$

$$P_{avg} = \frac{1}{\pi} \int_0^{\pi} P_R dt$$

$$P_{avg} = \frac{1}{\pi} \int_0^{\pi} P_R dt$$

$$\text{As } P_R = i^2 R \Rightarrow P_{avg} = \frac{1}{\pi} \int_0^{\pi} i^2 R dt$$

The effective value of AC current is equal to DC current and effective value of AC voltage is equal to DC voltage if  $P_R = P_{avg}$

$$I^2 R = \frac{1}{\pi} \int_0^{\pi} i^2 R dt$$

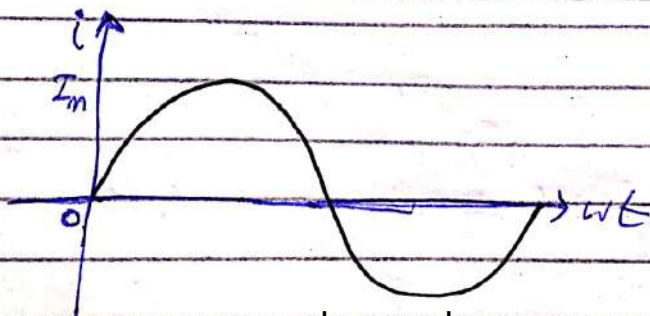
let us consider two cycles

$$I^2 R = \frac{1}{2\pi} \int_0^{2\pi} i^2 R dt$$

As R is constant -  $2\pi$

$$I^2 = \frac{1}{2\pi} \int_0^{2\pi} i^2 dt$$

$$I_{\text{effective}} = \sqrt{\frac{1}{2\pi} \int_0^{2\pi} i^2 dt} \rightarrow \text{RMS value}$$



$$I_{\text{mean}} = \frac{1}{2\pi} \int_0^{2\pi} i \, dwt$$

If we have time along the horizontal axis,

$$I_{\text{rms}} = \sqrt{\frac{1}{T} \int_0^T i^2 \, dt}$$

Put  $i = I_m \sin wt$  in  $I_{\text{eff}}$  or  $I_{\text{rms}}$ .

$$I_{\text{rms}} = \sqrt{\frac{1}{2\pi} \int_0^{2\pi} I_m^2 \sin^2 wt \, dwt}$$

As  $I_m$  is a constant value

$$I_{\text{rms}} = \sqrt{\frac{I_m^2}{2\pi} \int_0^{2\pi} \sin^2 wt \, dwt}$$

$$\text{As } \sin^2 wt = \frac{1 - \cos 2wt}{2}$$

$$I_{\text{rms}} = \sqrt{\frac{I_m^2}{4\pi} \left\{ \int_0^{2\pi} dwt - \int_0^{2\pi} \cos 2wt \, dwt \right\}}$$

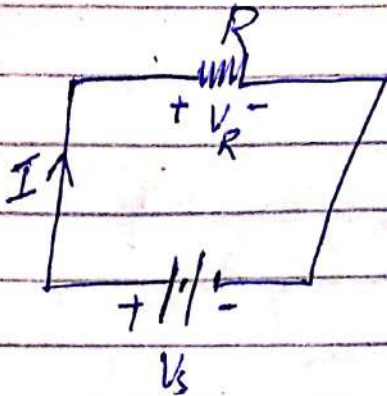
$\downarrow \frac{\sin wt}{2} \rightarrow \text{limits} \rightarrow 0$

$$I_{\text{rms}} = \sqrt{\frac{I_m^2}{4\pi} (wt) \Big|_0^{2\pi}}$$

$$I_{\text{rms}} = \sqrt{\frac{I_m^2}{2}} = \frac{I_m}{\sqrt{2}}$$

$$\textcircled{a} \quad I_{\text{rms}} = 0.707 I_m$$

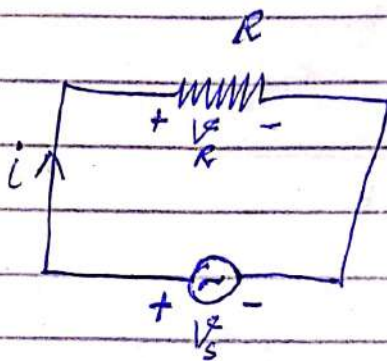
## RMS Value of AC Voltage.



According to KVL

$$V_s = V_R$$

$$P_R = \frac{V_s^2}{R} = \frac{V_R^2}{R}$$

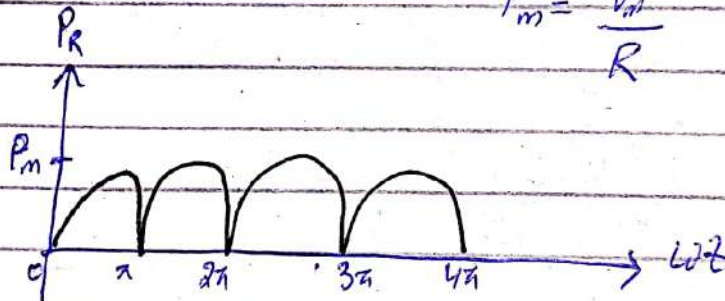


$$V_s = V_R = V_m \sin \omega t$$

$$P_R = \frac{V_R^2}{R} = \frac{V_s^2}{R}$$

$$P_R = \frac{V_m^2}{R} \sin^2 \omega t$$

$$P_m = \frac{V_m^2}{R}$$



$$P_{avg} = \frac{1}{\pi} \int_0^{\pi} P_R d\omega t$$

$$P_{avg} = \frac{1}{\pi} \int_0^{\pi} \frac{V_s^2}{R} d\omega t$$

$$\underline{P_R = P_{avg}} \rightarrow \text{for AC to be equal to DC}$$

$$\frac{V_e^2}{R} = \frac{1}{\pi} \int_0^{\pi} \frac{V_s^2}{R} d\omega t$$

(R is constant)

We can consider it for two cycles;

$$V_s^2 = \frac{1}{2\pi} \int_0^{2\pi} V_s^2 dt$$

$$V_{s\text{ rms}} = \frac{1}{2\pi} \int_0^{2\pi} V_s^2 dt$$

If we have time along horizontal axis;

$$V_{s\text{ rms}} = \frac{1}{T} \int_0^T V_s^2 dt$$

Put value of  $V_s$  in equation of  $V_{s\text{ rms}}$ .

$$V_{s\text{ rms}} = \sqrt{\frac{1}{2\pi} \int_0^{2\pi} V_m^2 \sin^2 \omega t dt}$$

$$V_{s\text{ rms}} = \sqrt{\frac{V_m^2}{2\pi} \int_0^{2\pi} \sin^2 \omega t dt}$$

$$\text{As } \sin^2 \omega t = \frac{1 - \cos 2\omega t}{2}$$

$$V_{s\text{ rms}} = \sqrt{\frac{V_m^2}{4\pi} \left\{ \int_0^{2\pi} dt - \int_0^{2\pi} \cos 2\omega t dt \right\}}$$

$$V_{s\text{ rms}} = \sqrt{\frac{V_m^2}{4\pi} (wt) \Big|_0^{2\pi}}$$

$$\frac{V_s}{\text{rms}} = \sqrt{\frac{V_m^2}{2}} = \frac{V_m}{\sqrt{2}} = 0.707 V_m$$

Example 1

$$V_{s,rms} = 220 \text{ V} \quad f = 50 \text{ Hz}$$

(Main AC source)  $V_s = ?$

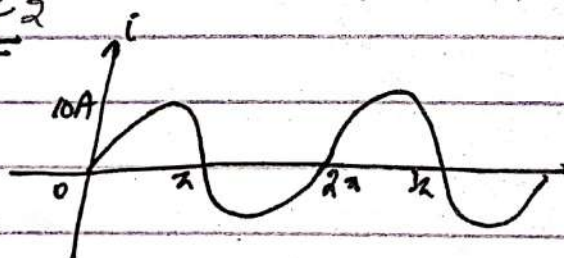
$$V_s = V_m \sin \omega t \rightarrow (i)$$

$$V_{s,rms} = \frac{V_m}{\sqrt{2}} \Rightarrow V_m = V_{s,rms} \sqrt{2}$$

$$V_m = 220 \sqrt{2} \text{ volts}$$

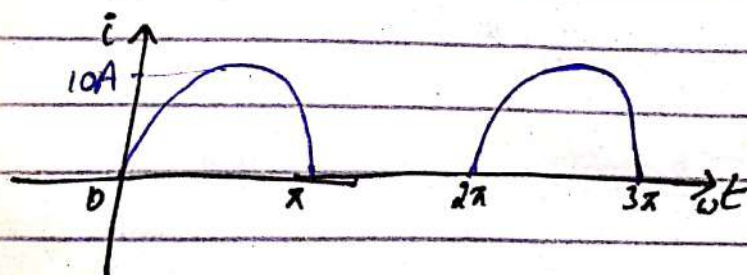
$$\omega = 2\pi f = 100\pi \text{ rad/sec}$$

$$(i) \Rightarrow \boxed{V_s = 220 \sqrt{2} \sin 100\pi t}$$

Example 2

calculate the rms  $I_m = 10 \text{ A}$

$$I_{rms} = \frac{I_m}{\sqrt{2}} = \frac{10}{\sqrt{2}} = 7.07 \text{ A}$$

Example 3

$$I_m = 10 \text{ A}$$

$$I_{avg} = ?$$

$$I_{rms} = ?$$

$$i = 0$$

$$\pi \leq \omega t \leq 2\pi$$

$$I_{avg} = \frac{1}{2\pi} \int_0^{2\pi} i \, d\omega t$$

$$I_{avg} = \frac{1}{2\pi} \left[ \int_0^{\pi} i \, d\omega t + \int_{\pi}^{2\pi} i \, d\omega t \right]$$

As  $i=0$  in this interval

$$I_{avg} = \frac{10}{2\pi} \int_0^{\pi} \sin \omega t \, d\omega t$$

$$I_{avg} = \frac{10}{2\pi} \left[ -\cos \omega t \right]_0^{\pi} = \frac{10}{2\pi} \times 2$$

$$I_{avg} = 3.18 \text{ A}$$

$$I_{rms} = \sqrt{\frac{1}{2\pi} \int_0^{2\pi} i^2 \, d\omega t}$$

$$I_{rms} = \left[ \frac{1}{2\pi} \left\{ \int_0^{\pi} i^2 \, d\omega t + \int_{\pi}^{2\pi} i^2 \, d\omega t \right\} \right]^{1/2}$$

$$I_{rms} = \sqrt{\frac{100}{2\pi} \int_0^{\pi} \left( \frac{1 - \cos 2\omega t}{2} \right) \, d\omega t}$$

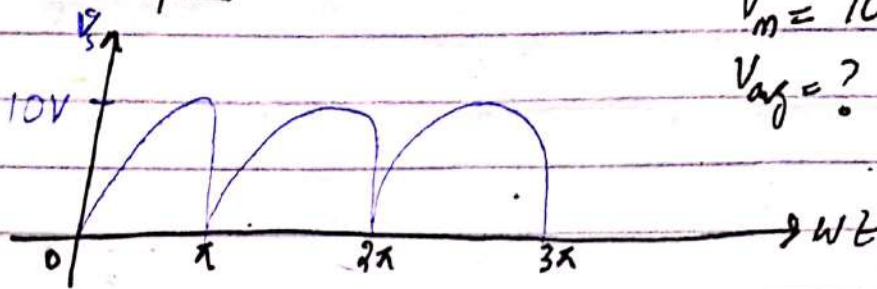
$$I_{rms} = \frac{100}{\sqrt{2\pi}} \left[ \omega t \right]_0^{\pi}$$

$$I_{rms} = \sqrt{25} = 5 \text{ A}$$



AC voltmeter reads rms value.

DC Ammeter reads the constant and average value.

Example 4

$$V_m = 10 \text{ V}$$

$$V_{avg} = ? \quad V_{rms} = ?$$

$$v_s = 10 \sin \omega t \quad 0 \leq \omega t \leq \pi$$

$$v_s = 10 \sin(\omega t - \pi) \quad \rightarrow \pi \leq \omega t \leq 2\pi$$

or  $v_s = -10 \sin \omega t$

$$V_{avg} = \frac{1}{\pi} \int_0^{\pi} v_s \, d\omega t = \frac{10}{\pi} \int_0^{\pi} \sin \omega t \, d\omega t$$

$$= \frac{10}{\pi} \left[ -\cos \omega t \right]_0^{\pi} = \frac{20}{\pi}$$

$$V_{avg} = 6.36 \text{ V}$$

$$V_{rms} = \sqrt{\frac{1}{\pi} \int_0^{\pi} v_s^2 \, d\omega t} = \sqrt{\frac{100}{\pi} \int_0^{\pi} \sin^2 \omega t \, d\omega t}$$

$$= \sqrt{\frac{100}{2\pi} \left[ \int_0^{\pi} d\omega t - \int_0^{\pi} \cos 2\omega t \, d\omega t \right]}$$

$$= \sqrt{50} = 2\sqrt{2} = 7.07 \text{ volts.}$$

Example 5

$$I_{avg} = ? \quad I_{rms} = ?$$



This waveform is repeating itself after 2s so its time period is 2s.  
 $T = 2s$

$$i = 10 \text{ A} \quad 0 \leq t \leq 1$$

$$i = 0 \text{ A} \quad 1 \leq t \leq 2$$

$$I_{avg} = \frac{1}{T} \int_0^T i \, dt$$

$$I_{avg} = \frac{1}{2} \left[ \int_0^1 10 \, dt + \int_1^2 0 \, dt \right]$$

$$I_{avg} = \frac{10}{2} \left( t \Big|_0^1 + 0 \right) = \boxed{5 \text{ A}}$$

$$I_{rms} = \sqrt{\frac{1}{T} \int_0^T i^2 \, dt}$$

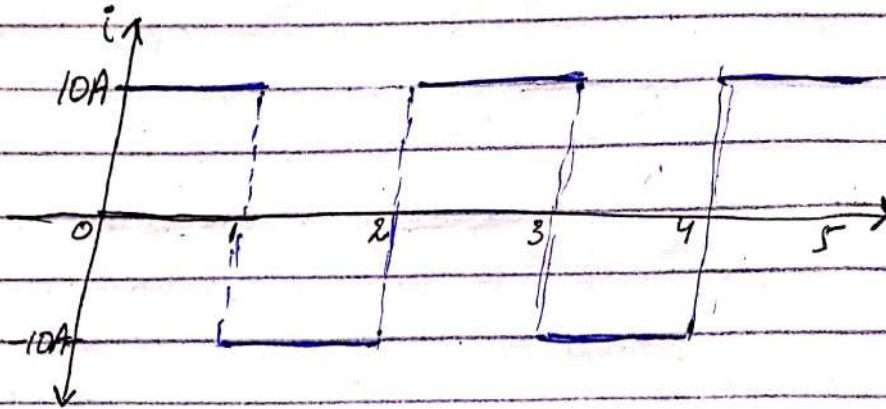
$$I_{rms} = \left[ \frac{1}{2} \left\{ \int_0^1 10^2 \, dt + \int_1^2 0^2 \, dt \right\} \right]^{1/2}$$

$$I_{rms} = \frac{100}{2} \left( t \Big|_0^1 + 0 \right) = \sqrt{50}$$

$$\boxed{I_{rms} = 7.07 \text{ A}}$$

Example 6:

$$I_{avg} = ? \quad I_{rms} = ? \quad T = 2 \text{ sec}$$



$$i = 10A \quad 0 \leq t \leq 1$$

$$i = -10A \quad 1 \leq t \leq 2$$

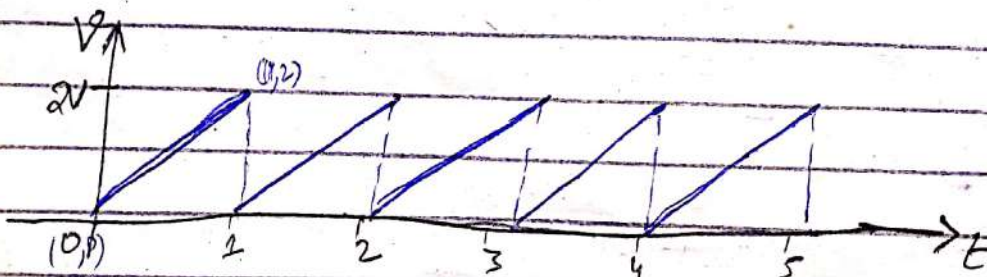
$$I_{rms} = \sqrt{\frac{1}{T} \int_0^T i^2 dt} = \sqrt{\frac{1}{2} \left\{ \int_0^1 10^2 dt + \int_1^2 (-10)^2 dt \right\}}$$

$$I_{rms} = \sqrt{\frac{100}{2} (t/2) + \frac{100}{2} (t/2)} = \sqrt{\frac{50 + 100}{-50}}$$

$$I_{rms} = 10A$$

$$I_{avg} = \frac{1}{T} \int_0^T i dt = \frac{1}{2} \left\{ \int_0^1 10 dt + \int_1^2 -10 dt \right\}$$

$$I_{avg} = 0A$$

Example 7:

$$T = 2 \text{ sec}$$

The eq for a straight line passing through the origin  $V = mt$

$$m = \frac{V_2 - V_1}{t_2 - t_1} = \frac{2 - 0}{1 - 0} = 2$$

$$V = 2t \quad 0 \leq t \leq 1$$

Similarly  $V = 2(t-1) \quad 1 \leq t \leq 2$

$$V = 2(t-2) \quad 2 \leq t \leq 3$$

$$V = 2(t-3) \quad 3 \leq t \leq 4$$

$$V = 2(t-4) \quad 4 \leq t \leq 5$$

$$V_{\text{avg}} = \frac{1}{T} \int_0^T V dt$$

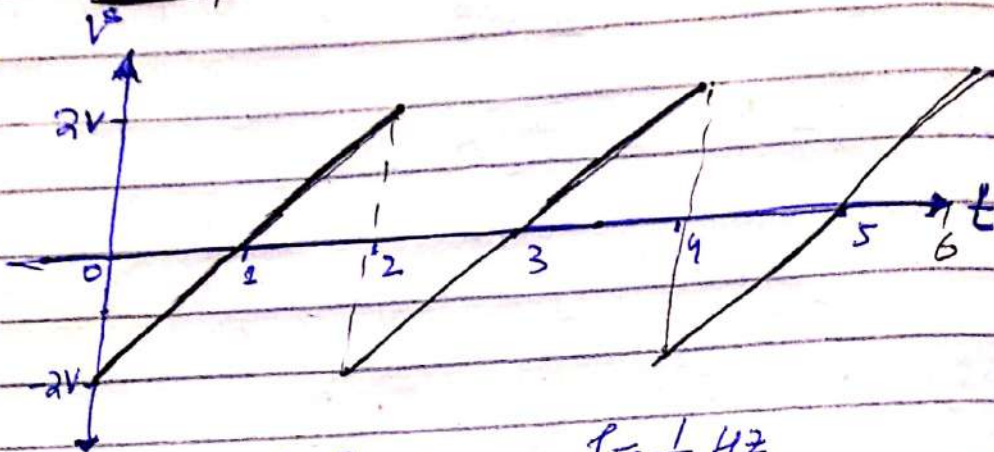
$$V_{\text{avg}} = \frac{1}{1} \int_0^1 2t dt = 2 \left( \frac{t^2}{2} \right) \Big|_0^1 = \boxed{1 \text{ volt}}$$

↓  
reading by  
DC voltmeter.

$$V_{\text{rms}} = \sqrt{\frac{1}{T} \int_0^T V^2 dt}$$

$$V_{\text{rms}} = \sqrt{\frac{1}{1} \int_0^1 (2t)^2 dt} = \sqrt{4 \left( \frac{t^3}{3} \right) \Big|_0^1} = \sqrt{\frac{4}{3}}$$

$$\boxed{V_{\text{rms}} = 1.154 \text{ V}}$$

Example 8

$$T = 2 \text{ sec}$$

$$f = \frac{1}{2} \text{ Hz}$$

$$V = mt + C$$

$$(y = mx + c)$$

$$m = \frac{V_2 - V_1}{t_2 - t_1} = \frac{2 - (-2)}{2} = 2$$

$$V = 2t + C$$

Take any point (t, V) from graph to find C.  
Let (1, 0)  $\Rightarrow$   $C = -2$

$$V = 2t - 2 \quad 0 \leq t \leq 2$$

$$V = 2(t-2) - 2 \quad 2 \leq t \leq 4$$

$$V = 2(t-4) - 2 \quad 4 \leq t \leq 6$$

$$V_{\text{avg}} = \frac{1}{T} \int_0^T V dt = \frac{1}{2} \int_0^2 (2t - 2) dt$$

$$V_{\text{avg}} = \frac{1}{2} \left[ \int_0^2 2t dt - \int_0^2 2 dt \right]$$

$$= \left( \frac{t^2}{2} \right)_0^2 - \left( \frac{t}{1} \right)_0^2 = \boxed{0 \text{ Volts}}$$

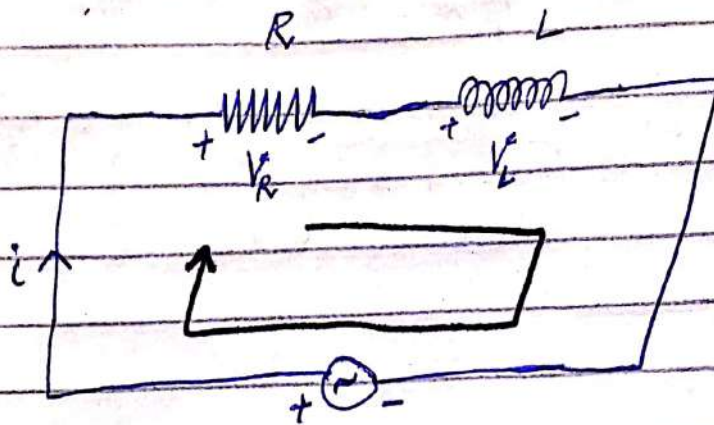
$$V_{\text{rms}} = \sqrt{\frac{1}{T} \int_0^T V^2 dt}$$

$$= \sqrt{\frac{1}{2} \int_0^2 (2t - 2)^2 dt} = \sqrt{\frac{1}{2} \int_0^2 (4t^2 + 4 - 8t) dt}$$

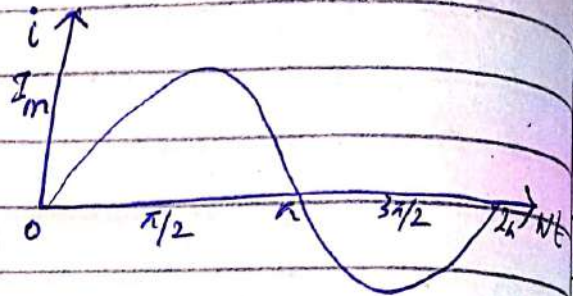
$$= \left[ \frac{1}{2} \left\{ 4 \left( \frac{t^3}{3} \right) \Big|_0^2 + 4(t) \Big|_0^2 - 8 \left( \frac{t^2}{2} \right) \Big|_0^2 \right\} \right]^{1/2}$$

$$= \sqrt{2 \times \frac{8}{3} + 4 - 8} = 1.15 \text{ volts.}$$

# R-L Series Circuit



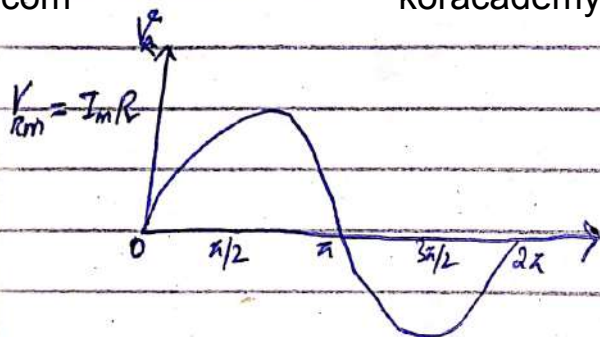
$$i = I_m \sin \omega t$$



$$V_R = iR$$

$$\Rightarrow V_R = I_m R \sin \omega t$$

$$V_{Rm} = I_m R \Rightarrow R = \frac{V_{Rm}}{I_m}$$



The voltage across the resistor is always in phase with the current.

$$\text{Also } V_L = L \frac{di}{dt} = L \frac{d(I_m \sin \omega t)}{dt}$$

$$V_L = I_m \omega L \cos \omega t$$

$$V_{Lm} = I_m \times (\omega L)$$

This ratio gives the resistance of the inductor.

$$\omega L = \frac{V_{Lm}}{I_m}$$

The resistance offered by inductor to AC is called inductive reactance.

which is represented by  $X_L$  and

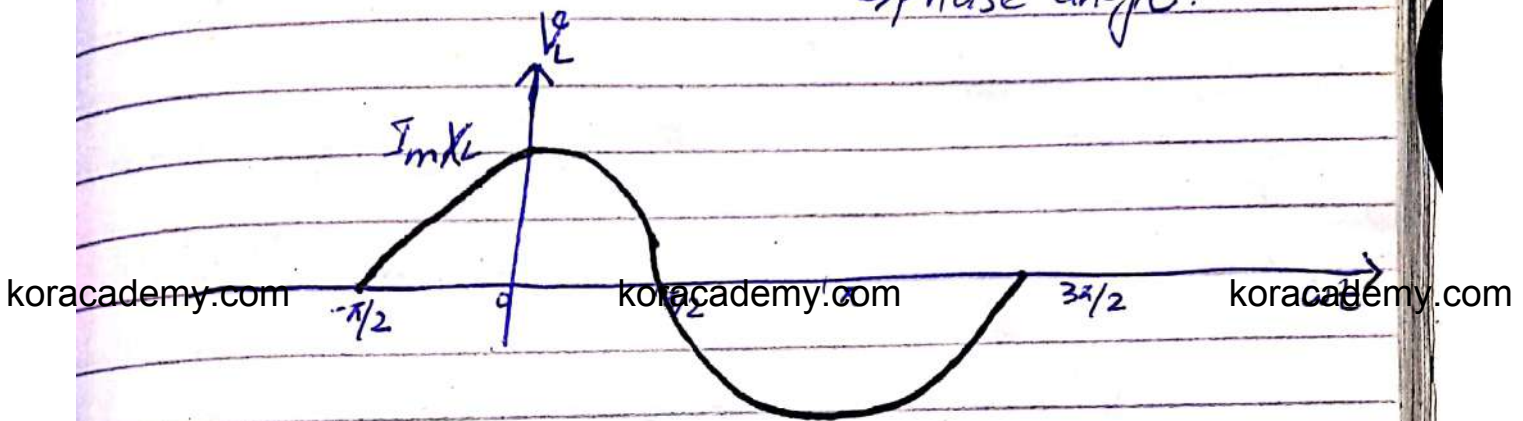
$$X_L = \omega L = 2\pi f L$$

Inductor does not offer any resistance to the flow of DC because its frequency is zero.

$$V_L = I_m X_L \cos \omega t$$

$$\textcircled{ii} V_L = I_m X_L \sin(\omega t + 90^\circ)$$

↳ phase angle.



The voltage across the inductor leads the current by  $90^\circ$ .

① the current in the inductor lags behind the voltage across it by  $90^\circ$ .  
There is a phase difference of  $90^\circ$ .

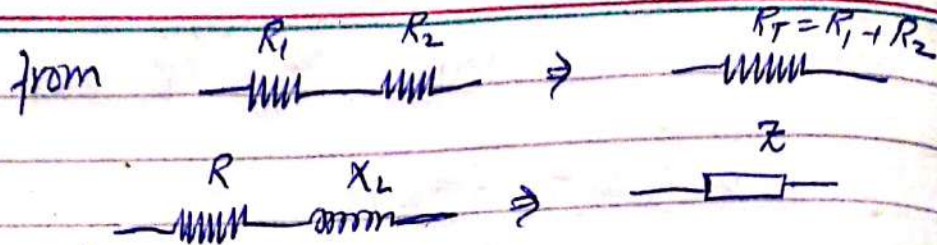
Apply KVL to the loop

$$V_s = V_R + V_L$$

$$V_s = I_m R \sin \omega t + I_m X_L \cos \omega t \rightarrow \textcircled{ii}$$

$$\text{Let } V_s = V_m \sin(\omega t + \phi)$$



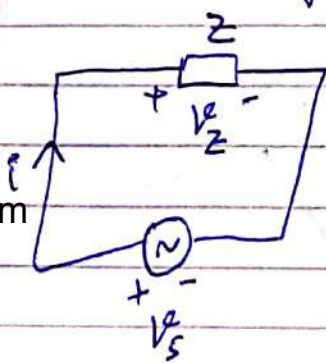


replace it theoretically.

total resistance  $Z \rightarrow$  impedance

Apply the same source across the passive circuit element so the same current will flow through it.

The voltage, current and total resistance of circuit 1 is equal to C2 and apply KVL to it.



$$\Rightarrow V_S = V_Z$$

$$V_m = V_{Zm} = I_m Z$$

$$V_S = I_m Z \sin(\omega t + \theta)$$

expand  $\sin(\alpha + \beta)$

$$V_S = I_m (Z \cos \theta) \sin \omega t + I_m (Z \sin \theta) \cos \omega t \rightarrow \text{⑥}$$

Eqs ④ and ⑥ gives the same source voltage; these equations are equal if

$$R = Z \cos \theta$$

and  $X_L = Z \sin \theta$

$$R^2 = Z^2 \cos^2 \theta \quad X_L^2 = Z^2 \sin^2 \theta$$

Add

$$R^2 + X_L^2 = Z^2 \cos^2 \theta + Z^2 \sin^2 \theta$$

$$Z^2 = R^2 + X_L^2$$

$$Z = \sqrt{R^2 + X_L^2}$$

where  $Z < R + X_L$

$$\frac{X_L}{R} = \frac{Z \sin \theta}{Z \cos \theta} \Rightarrow \theta = \tan^{-1} \left( \frac{X_L}{R} \right)$$

The instantaneous power taken by a resistor

$$P_R = i^2 R$$

$$P_R = I_m^2 R \sin^2 \omega t$$

$$P_R = \frac{1}{T} \int_0^T I_m^2 R \sin^2 \omega t \, d\omega t;$$

$$P_R = \frac{I_m^2 R}{T} \int_0^T \sin^2 \omega t \, d\omega t$$

$$\sin^2 \omega t = \frac{1 - \cos 2\omega t}{2}$$

$$P_R = \frac{I_m^2 R}{2T} \int_0^T (1 - \cos 2\omega t) \, d\omega t$$

$$P_R = \frac{I_m^2 R}{2T} \left\{ \int_0^T d\omega t - \int_0^T \cos 2\omega t \, d\omega t \right\}$$

$$P_R = \frac{I_m^2 R}{2T} (\omega t) \Big|_0^T$$

$$P_R = \frac{I_m^2 R}{2} = \left( \frac{I_m}{\sqrt{2}} \right)^2 R$$

$$P_R = I_{rms}^2 R$$

Power taken by inductor

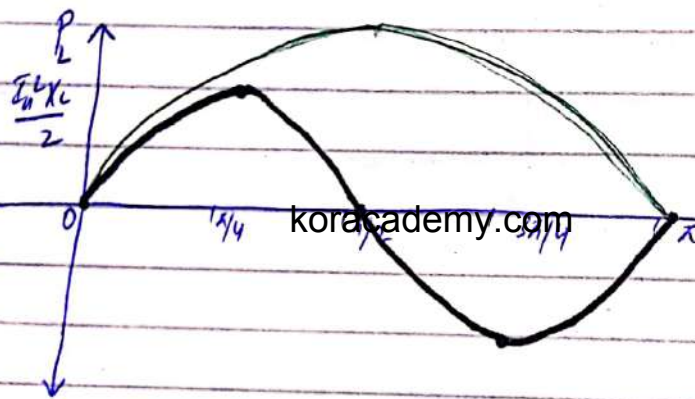
$$P_L = V_L \times i$$

$$P_L = (I_m X_L \cos \omega t) (I_m \sin \omega t)$$

$$P_L = \frac{I_m^2 X_L}{2} \sin \omega t \cos \omega t$$

$$2 \sin \omega t \cos \omega t = \sin 2\omega t$$

$$P_L = \frac{I_m^2 X_L}{2} \sin 2\omega t$$



When current in inductor increases, the inductor takes the power from voltage source. When the current decreases the same power is delivered back to the source.

The inductor does not consume this power so the average value of this power is zero which is calculated as,

$$P_L = \frac{1}{\pi} \int_0^{\pi} \frac{I_m^2 X_L}{2} \sin 2\omega t \, d\omega t$$

$$P_L = \frac{I_m^2 X_L}{2\pi} \int_0^{\pi} \sin 2\omega t \, d\omega t \quad P_L = 0$$

According to law of conservation of energy

$$P_S = P_R + P_L$$

$$P_S = I_{rms}^2 R$$

Example 1:

$$R = 3 \Omega \quad X_L = 4 \Omega \quad i = 10\sqrt{2} \sin 100\pi t \text{ A}$$

$$I_m = 10\sqrt{2} \text{ A} \quad \omega = 100\pi \text{ rad/sec}$$

$$V_R = I_m R \sin \omega t = 30\sqrt{2} \sin 100\pi t \text{ volts}$$

$$V_L = I_m X_L \sin(\omega t + 90^\circ) = 40\sqrt{2} \sin(100\pi t + 90^\circ)$$

$$V_S = I_m Z \sin(\omega t + \theta)$$

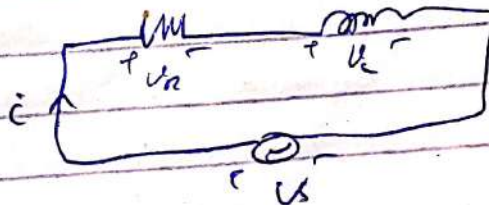
$$Z = \sqrt{R^2 + X_L^2} = \sqrt{9 + 16} \Rightarrow Z = 5 \Omega$$

$$\theta = \tan^{-1}\left(\frac{X_L}{R}\right) = \tan^{-1}\left(\frac{4}{3}\right) \Rightarrow \theta = 53.1^\circ$$

$$V_S = 50\sqrt{2} \sin(100\pi t + 53.1^\circ) \text{ volts.}$$

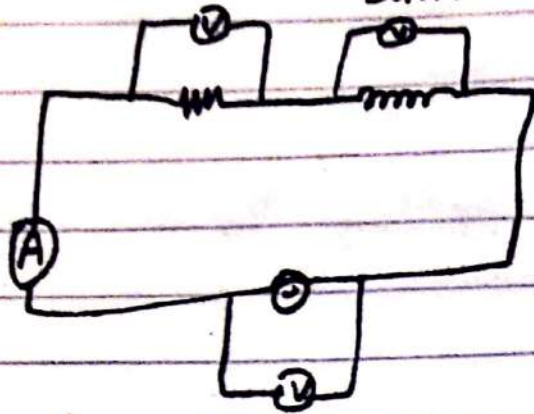
$$P_R = I_m^2 R \sin^2 \omega t = 600 \sin^2 100\pi t \text{ watts.}$$

$$P_L = \frac{I_m^2 X_L}{2} \sin 2\omega t = 400 \sin 200\pi t \text{ watts.}$$



Example 2

Same data as example 1.



Readings = ?

Ammeter reads  $I_{rms} = \frac{I_m}{\sqrt{2}} = 10A$

voltmeter across R  $V_R = \frac{30\sqrt{2}}{\sqrt{2}} = 30 \text{ volts.}$

voltmeter across  $Z$   $V_{rms} = \frac{40\sqrt{2}}{\sqrt{2}} = 40 \text{ volts.}$

$V_{rms} = \frac{50\sqrt{2}}{\sqrt{2}} = 50 \text{ volts.}$

Example 3

$R = 8\Omega$   $X_L = 6\Omega$   $i = 5\sqrt{2} \sin 100\pi t$

$I_{rms} = 5A$   $V_R = I_m R \sin \omega t = 40\sqrt{2} \sin 100\pi t$

$V_{R,rms} = 40V$

$V_L = I_m X_L \sin(\omega t + 90^\circ) = 30\sqrt{2} \sin(100\pi t + \pi/2)$

$V_{L,rms} = 30 \text{ volts.}$

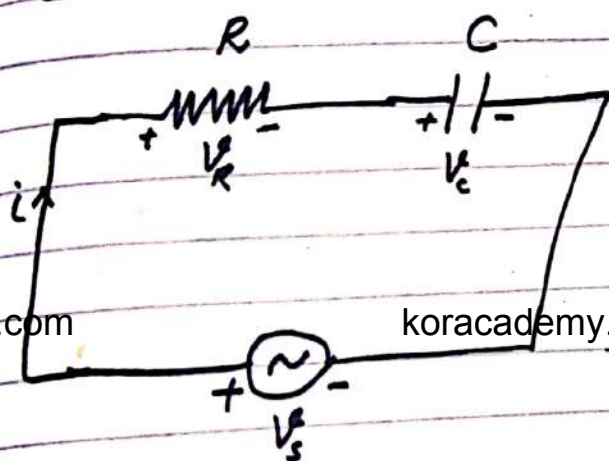
$$Z = \sqrt{4 + 36} = 10 \Omega$$

$$\theta = \tan^{-1} \left( \frac{6}{8} \right) = 36.8^\circ$$

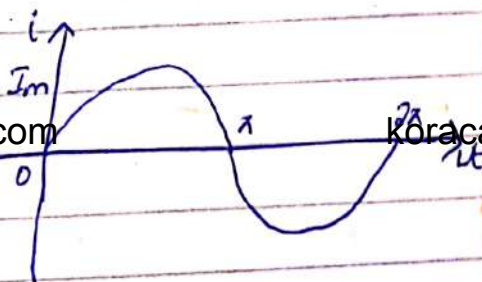
$$V_s = I_m Z \sin(\omega t + \theta) \\ = 50 \sqrt{2} \sin(100\pi t + 36.8^\circ)$$

$$V_{rms} = 50 \text{ volts.}$$

## R-C Series Circuit



$$i = I_m \sin \omega t$$



$$V_R = iR = I_m R \sin \omega t$$



The current and voltage through resistor are in phase with each other.

$$V_C = \frac{1}{C} \int i dt = \frac{1}{C} \int I_m \sin \omega t dt$$

$$V_C = -\frac{I_m}{\omega C} \cos \omega t = \frac{I_m}{\omega C} \sin(\omega t - 90^\circ)$$

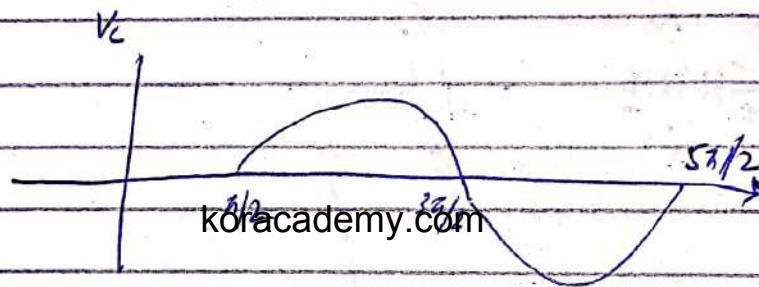
$$V_{cm} = \frac{I_m}{\omega C}$$

$$\frac{1}{\omega C} = \frac{V_{em}}{I_m} \quad \dots \quad (12)$$

This ratio gives the unit of resistance which is the resistance of the capacitor called as the capacitive reactance.

$$X_C = \frac{1}{\omega C} = \frac{1}{2\pi f C}$$

$$V_C = I_m X_C \sin(\omega t - 90^\circ)$$



The voltage across capacitor lags behind the current by  $90^\circ$  or the current leads voltage by  $90^\circ$ .

Apply KVL to this loop.

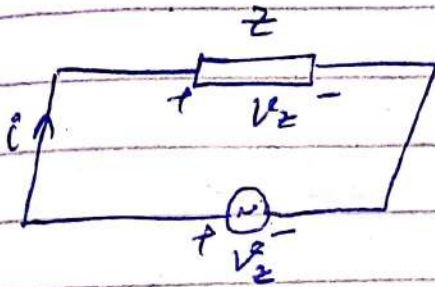
$$V_S = V_R + V_C$$

$$V_S = I_m R \sin \omega t - I_m X_C \cos \omega t \rightarrow (13)$$

This is AC voltage which varies sinusoidally with time. It will have proper maximum value and a proper phase angle which cannot be found out from this equation. So we write generally.

$$V_s = V_m \sin(\omega t - \theta)$$

Replace the combination of two circuit elements by a single circuit element whose resistance is equal to the total resistance of circuit 1.



Apply KVL to it.

$$V_s = V_Z$$

$$V_m = V_{Zm} = I_m Z$$

$$V_s = I_m Z \sin(\omega t - \theta)$$

expand  $\sin(\alpha - \beta)$

$$\textcircled{4} - V_s = I_m (Z \cos \theta) \sin \omega t - I_m (Z \sin \theta) \cos \omega t$$

$\textcircled{4} = \textcircled{5}$  as both represent both the source voltage.

These equations are equal if

$$R = Z \cos \theta \quad X_c = Z \sin \theta$$

$$\textcircled{6} \quad R^2 = Z^2 \cos^2 \theta \quad X_c^2 = Z^2 \sin^2 \theta$$

$$\text{Add} \quad R^2 + X_c^2 = Z^2 (\cos^2 \theta + \sin^2 \theta)$$

$$Z^2 = R^2 + X_c^2 \Rightarrow Z = \sqrt{R^2 + X_c^2}$$

$$\theta = \tan^{-1} \left( \frac{X_c}{R} \right)$$



Instantaneous power by resistor

$$P_R = I_m^2 R \sin^2 \omega t = i^2 R$$

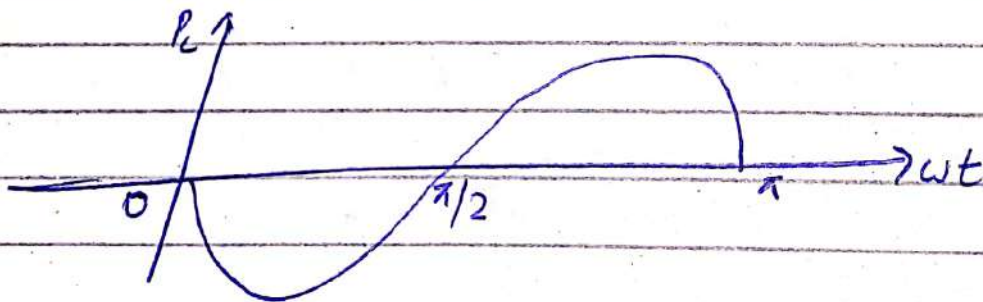
$$P_R = I_{rms}^2 R \rightarrow \text{average power.}$$

Instant power by capacitor

$$P_C = V_C i = - \frac{I_m^2 X_C^2 \sin^2 \omega t \cos \omega t}{2}$$

$$2 \sin \omega t \cos \omega t = \sin 2\omega t$$

$$P_C = - \frac{I_m^2 X_C^2 \sin 2\omega t}{2}$$



Average power.

$$P_C = \frac{1}{\pi} \int_0^{\pi} - \frac{I_m^2 X_C^2 \sin 2\omega t}{2} d\omega t$$

$$P_C = 0$$

According to law of conservation of energy

$$P_s = P_R + P_C$$

$$P_s = P_R$$

Example.

$$R = 6 \Omega \quad X_c = 8 \Omega \quad i = 10\sqrt{2} \sin 100\pi t \text{ A}$$

$$Z = \sqrt{R^2 + X_c^2} = \sqrt{36 + 64} = \sqrt{100} \Rightarrow Z = 10 \Omega$$

$$\theta = \tan^{-1} \frac{X_c}{R} = \tan^{-1} \frac{8}{6} \Rightarrow \theta = 53.1^\circ$$

$$V_R = I_m R \sin \omega t = 60\sqrt{2} \sin 100\pi t \text{ volts.}$$

$$V_C = I_m X_c \sin(\omega t - 90^\circ) = 80\sqrt{2} \sin(100\pi t - \pi/2)$$

$$V_s = I_m Z \sin(\omega t - 90^\circ) = 100\sqrt{2} \sin(100\pi t - 53.1^\circ)$$

$$P_R = I_m^2 R \sin^2 \omega t = 1200 \sin^2 100\pi t \text{ watt.}$$

$$P_C = -\frac{I_m^2 X_c}{2} \sin 2\omega t = 800 \sin 200\pi t \text{ watt.}$$

Consider the data of above example and calculate the rms values of the current, voltage across resistor, and the capacitor and average power supplied by source to entire circuit.

$$I_{rms} = \frac{I_m}{\sqrt{2}} = \frac{100\sqrt{2}}{\sqrt{2}} \Rightarrow I_{rms} = 100 \text{ A}$$

$$V_{Rrms} = 60 \text{ volts.} = \frac{60\sqrt{2}}{\sqrt{2}}$$

$$V_{C rms} = \frac{80\sqrt{2}}{\sqrt{2}} = 80 \text{ volts.}$$

$$V_{S rms} = \frac{100\sqrt{2}}{\sqrt{2}} = \frac{V_m}{\sqrt{2}} = 100 \text{ volts.}$$

$$P_s = P_R + P_C$$

$$P_C = 0 \quad P_R = I_{rms}^2 R$$

$$P_s = I_{rms}^2 R = 600 \text{ watt.}$$

AC ammeter reads rms value of current.

### Example

$$R = 4 \Omega \quad X_C = 3 \Omega \quad i = 5\sqrt{2} \sin 100\pi t \text{ A}$$

$$I_{rms} = 5 \text{ A}$$

$$Z = \sqrt{R^2 + X_C^2} = \sqrt{16 + 9} = 5 \Omega$$

$$\theta = \tan^{-1} \frac{X_C}{R} = \frac{3}{4} \Rightarrow \theta = 36.8^\circ$$

$$V_R = I_m R \sin \omega t = 20\sqrt{2} \sin 100\pi t \text{ volts.}$$

$$V_C = I_m X_C \sin(\omega t - 90^\circ) = 15\sqrt{2} \sin(100\pi t - \pi/2)$$

$$V_{R_{rms}} = 20 \text{ volts.} \quad V_{C_{rms}} = 15 \text{ volts.}$$

$$V_S = I_m Z \sin(\omega t - \theta) = 25\sqrt{2} \sin(100\pi t - 36.8^\circ)$$

$$P_R = I_m^2 R \sin^2 \omega t = 200 \sin^2 100\pi t \text{ watt.}$$

$$P_C = \frac{-I_m^2 X_C \sin 2\omega t}{2} = -75 \sin 2\omega t \text{ watts.}$$

$$V_{S_{rms}} = 25 \text{ volts.}$$

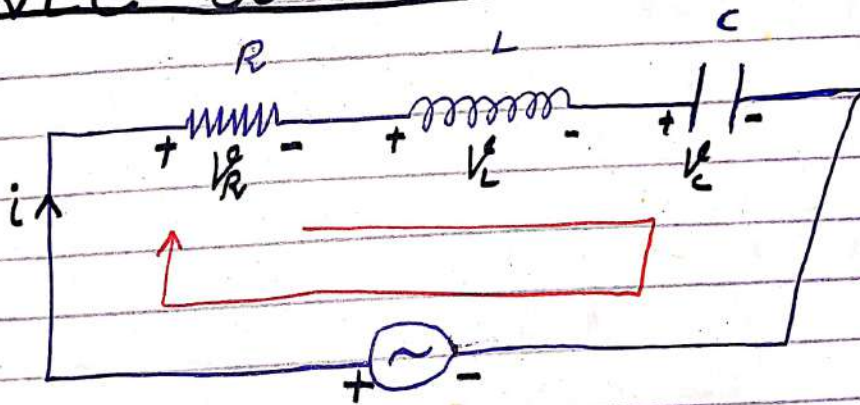
$$P_S = P_R + P_C$$

$$P_C = 0 \quad P_R = I_{rms}^2 R$$

$$P_S = 100 \text{ watts.}$$

②

## R.L.C. Series Circuit



$$i = I_m \sin \omega t$$

$$V_R = I_m R \sin \omega t \rightarrow \text{in phase with voltage}$$

$$V_L = I_m X_L (\sin(\omega t + 90^\circ)) \rightarrow \text{leads current by } 90^\circ$$

$$V_C = I_m X_C \sin(\omega t - 90^\circ) \rightarrow \text{lags behind current by } 90^\circ$$

waveforms of all 4.

Apply KVL to this loop

$$V_s = V_R + V_L + V_C$$

$$V_s = I_m R \sin \omega t + I_m X_L \cos \omega t - I_m X_C \cos \omega t$$

$$V_s = I_m R \sin \omega t + I_m (X_L - X_C) \cos \omega t \rightarrow \textcircled{a}$$

$$X = \text{net reactance} = (X_L - X_C)$$

This voltage is AC voltage so it has a proper maximum value and a phase angle which cannot be easily determined from this equation although it is correct and gives the ~~true~~ instantaneous value.

Generally  $V = V_m \sin(\omega t + \theta)$

if this angle  $\theta$  is 0 the source voltage is in phase with current and is a pure resistive circuit.

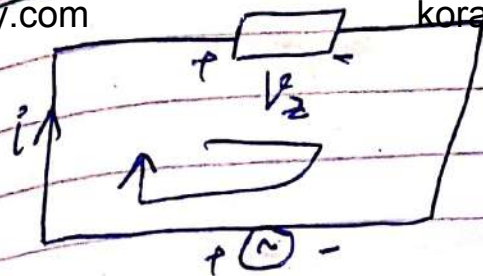
if  $\theta$  is positive circuit will be inductive.

if  $\theta$  is negative circuit will be capacitive.

Replace the series combination of these three circuit elements by a single passive circuit element theoretically.

It cannot be replaced practically due to the different nature of all the three elements.

let  $Z$  be the total equivalent resistance of all 3 elements.



Apply KVL  $V_s = V_z$

$$\Rightarrow V_m = V_m Z = I_m Z$$

$$V_s = I_m Z \sin(\omega t \pm \theta)$$

Expand  $\sin(\omega t \pm \theta)$

$$V_s = I_m (Z \cos \theta) \sin \omega t \pm I_m (Z \sin \theta) \cos \omega t \rightarrow \textcircled{a}$$

$$\textcircled{a} = \textcircled{b}$$

$$R = Z \cos \theta \Rightarrow R^2 = Z^2 \cos^2 \theta$$

$$(X_L - X_C) = Z \sin \theta \Rightarrow (X_L - X_C)^2 = Z^2 \sin^2 \theta$$

$$Z^2 = R^2 + (X_L - X_C)^2$$

$$Z = \sqrt{R^2 + (X_L - X_C)^2}$$

$$\tan \theta = \frac{X_L - X_C}{R} \quad \theta = \tan^{-1} \left( \frac{X_L - X_C}{R} \right)$$

(i) if  $X_L = X_C \rightarrow$  entire circuit will behave like resistive circuit and this condition is known as resonance. Here the source voltage will be in phase with current.

(ii) If  $X_L > X_C \rightarrow \theta$  is positive and the entire circuit will behave like

inductive circuit.

(iii) If  $X_L < X_C \Rightarrow \theta$  is negative and the circuit is capacitive.

Instantaneous power.

$$P_R = I_m^2 R \sin^2 \omega t$$

$$P_L = \frac{I_m^2 X_L \sin 2\omega t}{2}$$

$$P_C = -\frac{I_m^2 X_C \sin 2\omega t}{2}$$

Average  $P_R = I_{rms}^2 R$      $P_L = 0$      $P_C = 0$

With the series average value of power.

Law of conservation of energy.

$$P_S = P_L + P_R + P_C$$

$$P_S = P_R = I_{rms}^2 R$$

Example 1

$$R = 3 \Omega \quad X_L = 4 \Omega \quad X_C = 4 \Omega$$

$$i = 10\sqrt{2} \sin 100\pi t \text{ A}$$

$$I_{rms} = 10 \text{ A}$$

$$V_R = I_m R \sin \omega t = 30\sqrt{2} \sin 100\pi t \text{ volts.}$$

$$V_{R,rms} = 30 \text{ volts}$$

$$V_L = I_m X_L \sin(\omega t + \pi/2) = 40\sqrt{2} \sin(100\pi t + \pi/2)$$

$$V_{L_{rms}} = 40 \text{ volts.}$$

$$V_C = I_m X_C \sin(\omega t - \pi/2) = 40\sqrt{2} \sin(100\pi t - \pi/2)$$

$$V_{C_{rms}} = 40 \text{ volts.}$$

$$Z = \sqrt{R^2 + (X_L - X_C)^2} = R = 3 \Omega$$

$$\theta = \tan^{-1} \frac{X_L - X_C}{R} = 0$$

$$V_S = I_m Z \sin(\omega t \pm \theta) = V_R = 30\sqrt{2} \sin 100\pi t.$$

$$V_{S_{rms}} = 30 \text{ volts.}$$

under resonance condition,

→ Impedance = R

→ entire circuit offers minimum resistance and maximum current will flow.

$$\rightarrow V_S = V_R$$

$$P_R = 600 \sin^2 \omega t$$

$$P_C = -400 \sin^2 \omega t$$

$$P_L = 400 \sin^2 \omega t$$

watts.

Example

$$R = 8 \Omega$$

$$X_L = 16 \Omega$$

$$X_C = 10 \Omega$$

$$I = 10\sqrt{2} \sin 100\pi t$$

$$I_{rms} = 10 A$$

$$V_R = I_m R \sin \omega t = 80\sqrt{2} \sin 100\pi t$$

$$V_{R_{rms}} = 80 \text{ volts.}$$



$$V_L = I_m X_L \sin(\omega t + \pi/2) = 160\sqrt{2} \sin(100\pi t + \pi/2)$$

$$V_{L_{rms}} = 160 \text{ volts.}$$

$$V_C = I_m X_C \sin(\omega t - \pi/2) = 100\sqrt{2} \sin(100\pi t - \pi/2)$$

$$V_{C_{rms}} = 100 \text{ volts.}$$

$$Z = \sqrt{R^2 + (X_L - X_C)^2} = 10\Omega$$

$$\theta = \tan^{-1} \frac{X_L - X_C}{R} = 36.8^\circ$$

$$V_s = I_m Z \sin(\omega t \pm \theta) = 100\sqrt{2} \sin(100\pi t + 36.8^\circ) \text{ volts}$$

$$V_{s_{rms}} = 100 \text{ volts.}$$

$$P_R = 1600 \sin^2 100\pi t \text{ watt}$$

$$P_L = 1600 \sin 2\omega t$$

$$P_C = -100 \sin 2\omega t$$

### Example

$$R = 6\Omega \quad X_L = 10\Omega \quad X_C = 18\Omega$$

$$i = 10\sqrt{2} \sin 100\pi t \quad I_{rms} = 10 \text{ A}$$

$$V_R = I_m R \sin \omega t = 60\sqrt{2} \sin 100\pi t \text{ volts.}$$

$$V_{R_{rms}} = 60 \text{ volts.}$$

$$V_L = I_m X_L \sin(\omega t + \pi/2) = 100\sqrt{2} \sin(100\pi t + \pi/2)$$

$$V_{L_{rms}} = 100 \text{ volts.}$$

$$V_c = I_m X_c \sin(\omega t - \pi/2) = 180\sqrt{2} \sin(100\pi t - \pi/2)$$

$$V_{c\text{rms}} = 180 \text{ V}$$

$$Z = \sqrt{R^2 + (X_L - X_C)^2} = 10 \Omega$$

$$\theta = \tan^{-1} \frac{X_L - X_C}{R} = -53.2$$

$$V_s = I_m Z \sin(\omega t + \theta) = 100\sqrt{2} \sin(100\pi t - 53.2^\circ)$$

$$V_{s\text{rms}} = 100 \text{ volts}$$

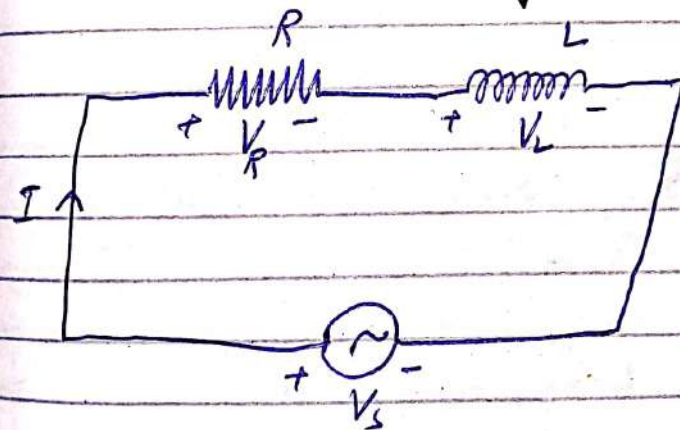
$$P_R = 1200 \sin^2 100\pi t$$

$$P_C = -1800 \sin 2\omega t$$

$$P_L = 1000 \sin 2\omega t$$

Write

### ③ Phasor Analysis of R-L Series



$$i = I_m \sin(\omega t + \theta)$$

$$\text{Max Value} = I_m$$

Rms values

$$I_{\text{rms}} = \frac{I_m}{\sqrt{2}}$$

$$\text{Phasor value } I = I_{\text{rms}} \angle 0^\circ = I_{\text{rms}}$$

$$V_R = I_m R \sin(\omega t + \theta)$$

$$\text{Max value} = V_{\text{Rm}} = I_m R$$

Rms value,  $V_{rms} = \frac{I_m R}{\sqrt{2}}$

$$V_{rms} = I_{rms} R$$

Phasor value  $V_R = I_{rms} R \angle 0^\circ = I_{rms} R$

As  $\theta = 0^\circ$   $\sin 0 = 0$

$$V_L = I_m X_L \sin(\omega t + 90^\circ)$$

Max value =  $I_m X_L$

Rms value,  $V_{rms} = \frac{I_m X_L}{\sqrt{2}} = I_{rms} X_L$

Phasor value  $V_L = I_{rms} X_L \angle 90^\circ = j I_{rms} X_L$

as  $\cos 90^\circ = 0$

max mag  
phasor

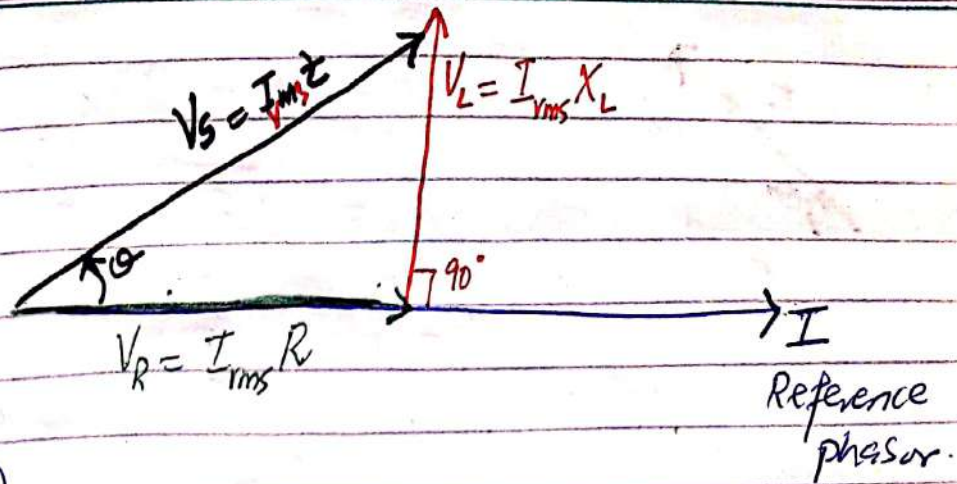
Apply KVL;

The phasor sum of the voltage rises in a loop will be equal to the sum of all the voltage drops.

$$V_s = V_R + V_L$$

We add these two phasors by head to tail rule to get the phasor for the source voltage.

$$V_{s,rms} = \sqrt{V_R^2 + V_L^2}$$



(OR) Consider the instantaneous equation

$$V_s = I_m Z \sin(\omega t + \theta)$$

max Value =  $I_m Z$       Rms  $V_{s,rms} = \frac{I_m Z}{\sqrt{2}}$

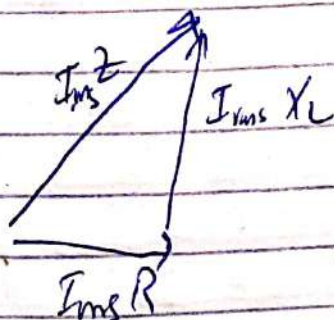
$$V_{s,rms} = I_{rms} Z$$

$$I_{rms} = \frac{V_{s,rms}}{Z}$$

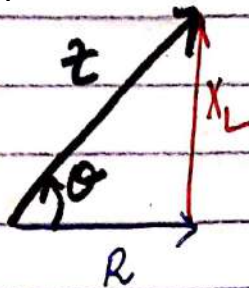
Phasor value  $V_s = I_{rms} Z \angle \theta$



Consider the upper phasor diagram. Let ignoring the current, now we don't call it a phasor diagram rather call it a voltage triangle.



As  $I_{rms}$  is common. Divide all the 3 sides of this  $\Delta$  by  $I_{rms}$  to get a new triangle



$$Z = R + jX_L$$

$$|Z| = \sqrt{R^2 + X_L^2}$$

$$\theta = \tan^{-1} \left( \frac{X_L}{R} \right)$$

$$\cos \theta = \frac{R}{Z}$$

Average values of power

$$P_s = P_R + P_{\cancel{L}}$$

$$P_s = I_{rms}^2 R$$

$$P_s = I_{rms} \times I_{rms} \times R$$

$$\text{As } I_{rms} = \frac{V_{rms}}{Z}$$

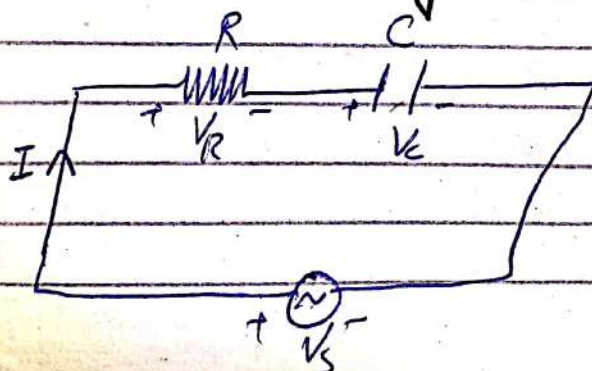
$$P_s = V_{s,rms} \times I_{rms} \times \frac{R}{Z}$$

$$P_s = V_{s,rms} I_{rms} \cos \theta$$

$\cos \theta \rightarrow$  power factor

$\hookrightarrow$  lagging power factor.

## Phasor Analysis of RC Series Circuit



$$i = I_m \sin(\omega t + \theta)$$

$$\text{max value} = I_m$$

$$P_{rms} \text{ value} = I_{rms}$$

$$= \frac{I_m}{\sqrt{2}}$$

Phasor value  $I = I_{rms} \angle 0^\circ = I_{rms}$

$$V_R = I_m R \sin(\omega t + 0^\circ)$$

max value =  $V_{m} = I_m R$

Rms value =  $V_{rms} = \frac{I_m}{\sqrt{2}} R = I_{rms} R$

Phasor value  $V_R = I_{rms} R \angle 0^\circ = I_{rms} R$

$$V_C = I_m X_C \sin(\omega t - 90^\circ)$$

Max value =  $I_m X_C$

Rms value ;  $V_{rms} = \frac{I_m}{\sqrt{2}} X_C = I_{rms} X_C$

Phasor value  $V_C = I_{rms} X_C \angle -90^\circ = -j I_{rms} X_C$

**KVL**  $V_S = V_R + V_C$

$$V_{S rms} = \sqrt{V_R^2 + V_C^2}$$

$$V_S = I_m Z \sin(\omega t - \theta)$$

Max value =  $I_m Z$

$$I_{rms} = \frac{V_{rms}}{Z}$$

$$V_{S rms} = I_{rms} Z$$

$$I_{rms} = \frac{V_{S rms}}{Z}$$

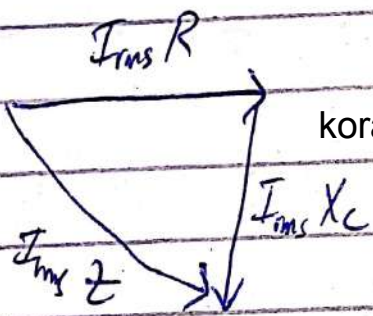
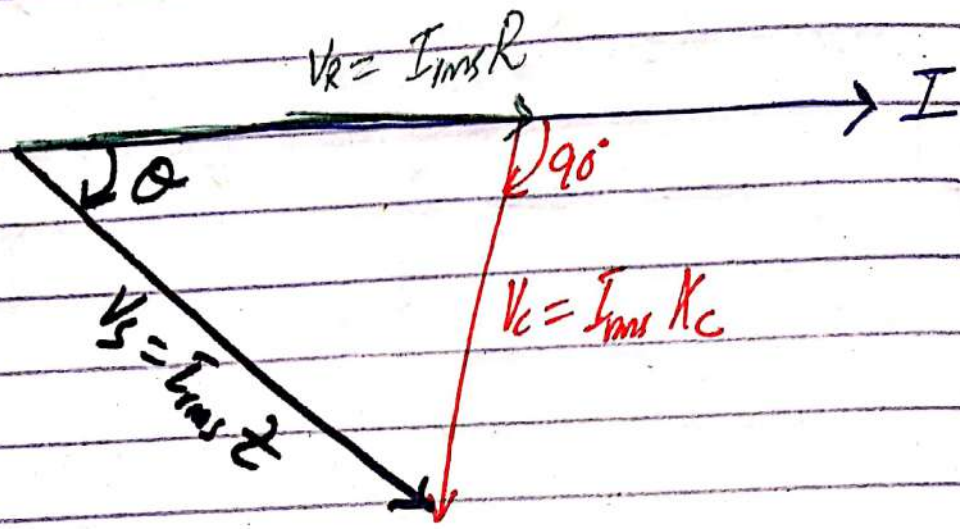
Phasor value  $V_S = I_{rms} Z \angle \theta$

Avg<sup>e</sup>  
power

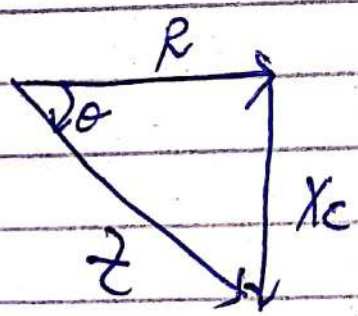
$$P_S = P_R + P_C = I_{rms}^2 R$$

$$P_s = \frac{V_r}{rms} \times \frac{I}{rms} \times \cos \theta$$

$\cos \theta \rightarrow$  power factor  
Here leading power factor



theta angle -



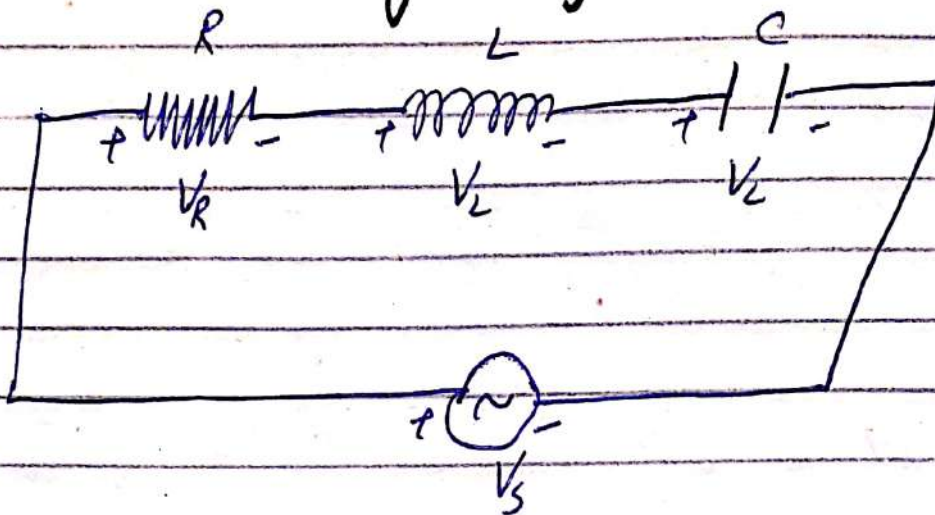
$$Z = R + jX_c$$

$$|Z| = \sqrt{R^2 + X_c^2}$$

$$\theta = \tan^{-1} \left( \frac{X_c}{R} \right)$$

Power factor  $\cos \theta = \frac{R}{Z}$  for this angle.

# Phasor Analysis of RLC Series Circuit



$$i = I_m \sin(\omega t + \theta)$$

Max value =  $I_m$       Rms       $I_{rms} = \frac{I_m}{\sqrt{2}}$

Phasor value       $I = I_{rms} \angle 0^\circ = I_{rms}$

$V_R = I_m R \sin(\omega t + \theta)$

Max value       $V_{Rm} = I_m R$

Rms value       $V_{R_{rms}} = \frac{I_m R}{\sqrt{2}} = I_{rms} R$

Phasor value of  $V_R = I_{rms} R \angle 0^\circ = I_{rms} R$

$$V_L = I_m X_L \sin(\omega t + 90^\circ)$$

Max value =  $I_m X_L$       Rms value       $V_{L_{rms}} = \frac{I_m X_L}{\sqrt{2}}$

$$V_{L_{rms}} = I_{rms} X_L$$

Phasor value of  $V_L = I_{rms} X_L \angle 90^\circ = j I_{rms} X_L$



$$V_c = I_m X_c \sin(\omega t - 90^\circ)$$

$$\text{Max value} = I_m X_c$$

$$\text{Rms value } \frac{V_c}{\text{rms}} = \frac{I_m}{\sqrt{2}} X_c = I_{\text{rms}} X_c$$

$$\text{Phasor value of } V_c = I_{\text{rms}} X_c \angle -90^\circ = -j I_{\text{rms}} X_c$$

Apply KVL

$$V_s = V_R + jV_L - jV_C$$

$$V_s = V_R + j(V_L - V_C)$$

$$V_{s \text{ rms}} = \sqrt{V_R^2 + (V_L - V_C)^2} \rightarrow \textcircled{1}$$

① Resistive Circuit

$$\text{If } X_L = X_C$$

$$I_{\text{rms}} X_L = I_{\text{rms}} X_C$$

$$V_{L \text{ rms}} = V_{C \text{ rms}}$$

$$V_L - V_C = 0$$

$$\textcircled{1} \Rightarrow V_s = V_R$$



## (ii) Inductive Circuit

$$\text{if } X_L > X_C$$

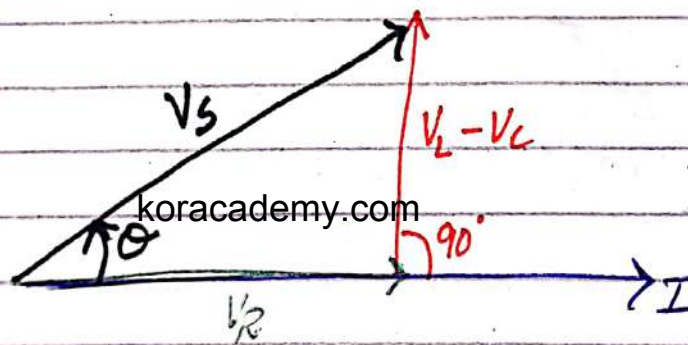
$$I_{\text{rms}} X_L > I_{\text{rms}} X_C$$

$$V_{L_{\text{rms}}} > V_{C_{\text{rms}}}$$

$$V_L - V_C > 0 \Rightarrow j \rightarrow \text{positive}$$

The combination of inductor and capacitor behaves like an inductor.

Voltage across inductor leads current by  $90^\circ$ .



## (iii) Capacitive Circuit

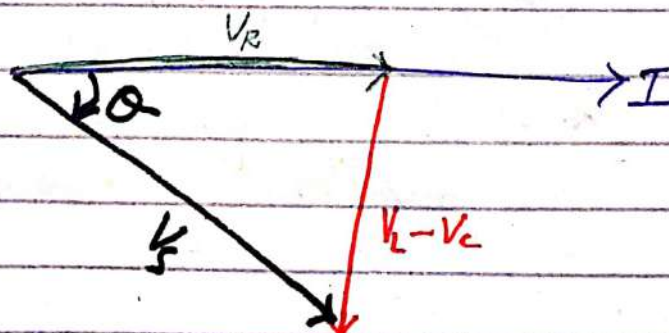
$$\text{if } X_L < X_C$$

$$I_{\text{rms}} X_L < I_{\text{rms}} X_C$$

$$V_{L_{\text{rms}}} < V_{C_{\text{rms}}}$$

$$V_L - V_C = -ve$$

$$j = -ve$$



## Example 1:

R-L Circuit.

$$R = 3\Omega \quad X_L = 4\Omega \quad V_{s\text{rms}} = 100 \text{ volts}$$

$$Z = R + jX_L = 3 + j4 \Omega$$

$$|Z| = \sqrt{9+16} = 5\Omega$$

$$\theta = \tan^{-1}\left(\frac{4}{3}\right) = 51.3^\circ$$

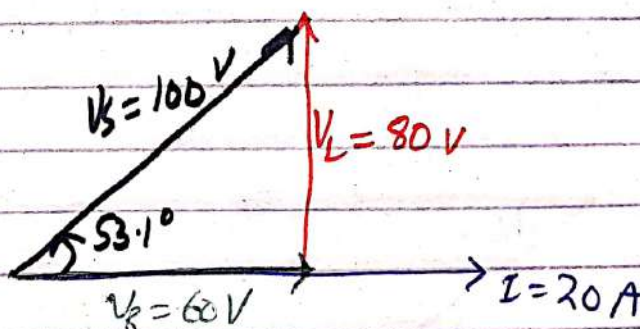
$$I_{\text{rms}} = \frac{V_{s\text{rms}}}{Z} = \frac{100}{5} = 20\text{A}$$

$$V_{R\text{rms}} = I_{\text{rms}} R = 60\text{V}$$

$$V_{L\text{rms}} = I_{\text{rms}} X_L = 80\text{V}$$

$$\text{Power factor} = \cos\theta = \frac{R}{Z} = \frac{3}{5} = 0.6 \text{ lagging power factor}$$

$$P_s = I_{\text{rms}}^2 R = \frac{V_s}{\text{rms}} \times I_{\text{rms}} \times \cos\theta = 100 \times 20 \times 0.6 = 1200 \text{ Watt.}$$



Example 2

R-C circuit

$$R = 3\Omega \quad X_C = 4\Omega \quad V_{s\text{rms}} = 100 \text{ volts}$$

$$Z = R - jX_C = 3 - j4 \Omega$$

$$|Z| = \sqrt{9 + 16} = 5\Omega$$

$$\theta = \tan^{-1}\left(\frac{4}{3}\right) = 53.1^\circ$$

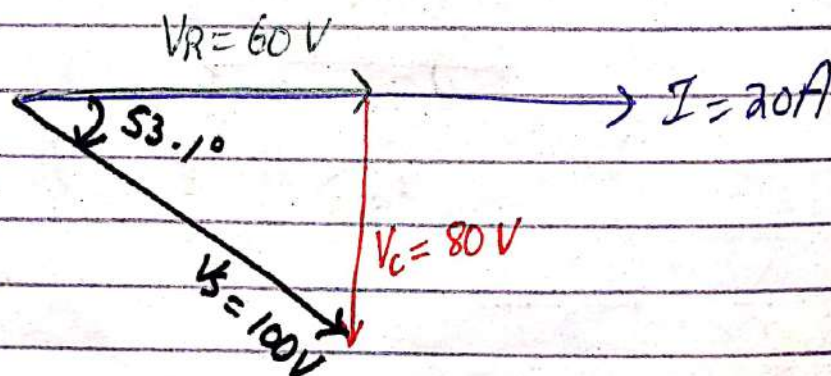
$$I_{\text{rms}} = \frac{V_{s\text{rms}}}{Z} = \frac{100}{5} = 20 \text{ A}$$

$$V_{\text{rms}} = I_{\text{rms}} R = 60 \text{ V}$$

$$V_{\text{rms}} = I_{\text{rms}} X_C = 80 \text{ V}$$

$$\text{Power factor} = \frac{R}{Z} = \frac{3}{5} = 0.6 \rightarrow \text{leading P.f.}$$

$$P_s = I_{\text{rms}}^2 R = \frac{V_s}{\text{rms}} \times \frac{I_s}{\text{rms}} \times \cos\theta = 1200 \text{ W}$$



Example 3:

RLC Series Circuit

$$X_L = 4 \Omega \quad R = 3 \Omega \quad X_C = 8 \Omega \quad V_{s_{rms}} = 100 V$$

$$Z = R + j(X_L - X_C)$$

$$Z = R - jX = 3 - j4 \Omega$$

$$|Z| = \sqrt{9 + 16} = 5 \Omega \quad \theta = \tan^{-1}\left(\frac{4}{3}\right) = 53.1^\circ$$

$$I_{rms} = \frac{V_{s_{rms}}}{Z} = \frac{100}{5} = 20 A$$

$$V_{R_{rms}} = I_{rms} R = 60 V$$

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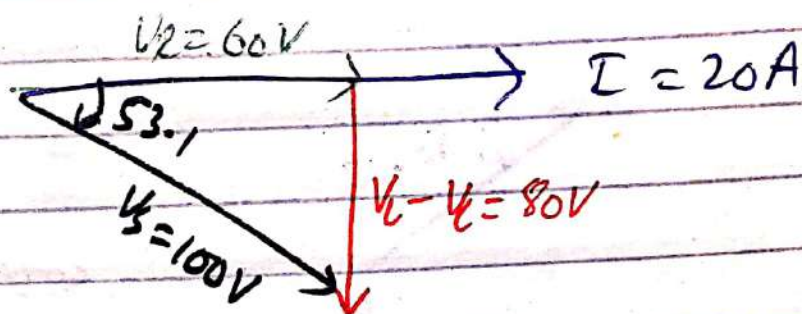
koracademy.com

$$V_{C_{rms}} = I_{rms} X_C = 160 V$$

$$V_{L_{rms}} = I_{rms} X_L = 80 V$$

$$\text{Power factor} = \frac{R}{Z} = 0.6 \rightarrow \text{leading power factor.}$$

$$P_s = I_{rms}^2 R = I_{rms} \times V_s \times \cos \theta = 1200 W$$



Example 4.

## RLC Series Circuit

$$X_C = 4\Omega \quad R = 3\Omega \quad X_L = 8\Omega \quad V_{s_{rms}} = 100\text{ V}$$

$$Z = R + j(X_L - X_C)$$

$$Z = R + jX = 3 + j4 \Omega$$

$$|Z| = \sqrt{9 + 16} = 5\Omega \quad \theta = \tan^{-1}\left(\frac{4}{3}\right) = 53.1^\circ$$

$$I_{rms} = \frac{V_{s_{rms}}}{Z} = \frac{100}{5} = 20\text{ A}$$

$$V_{R_{rms}} = I_{rms} R = 60\text{ V}$$

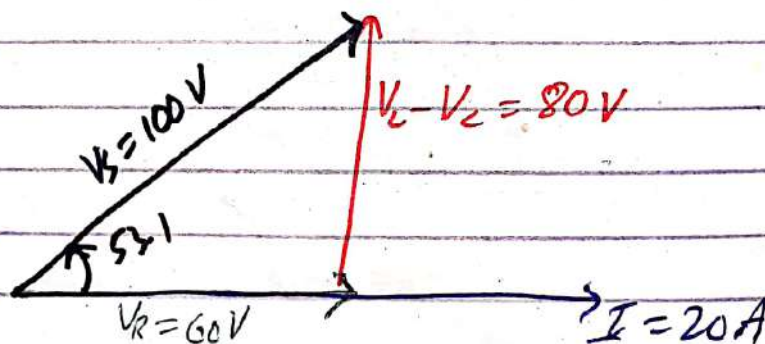
$$V_{L_{rms}} = I_{rms} X_L = 160\text{ V}$$

$$V_{C_{rms}} = I_{rms} X_C = 80\text{ V}$$

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$$\text{Power factor} = \frac{R}{Z} = \frac{3}{5} = 0.6 \rightarrow \text{lagging power factor.}$$

$$P_s = I_{rms}^2 R = I_{rms} \times V_{R_{rms}} = \cos\theta = 1200\text{ Watts.}$$



purely inductive circuit.

Example 5

$$X_C = 8 \Omega \quad R = 5 \Omega \quad X_L = 8 \Omega$$

$$V_{s_{rms}} = 100 \text{ volts}$$

$$Z = R + j(X_L - X_C)$$

$$Z = R = 5 \Omega \quad \text{Or } \tan^{-1}\left(\frac{0}{5}\right) = 0$$

$$I_{rms} = \frac{V_{rms}}{Z} = \frac{100}{5} = 20 \text{ A}$$

$$V_s = V_R = I_{rms} R = 100 \text{ V}$$

$$V_L = I_{rms} X_L = 160 \text{ V}$$

$$V_C = I_{rms} X_C = 160 \text{ V}$$

$$\text{Power factor} = \frac{R}{Z} = \frac{5}{5} = 1 \rightarrow \text{unity power factor}$$

$$P_{s2} = I_{rms}^2 R = V_s \times I_{rms} \times \cos \theta = 2000 \text{ Watt}$$

Circuit behaves purely resistive.



Example 6

$$P_s = 1200 \text{ W} \quad R = 6 \Omega \quad P.f = 0.6 \text{ lagging}$$

→ RL Series Circuit

$$P_s = P_R = \frac{I_{\text{rms}}^2 R}{\text{ms}}$$

$$I_{\text{rms}}^2 = \frac{P_s}{R} = \frac{1200}{6} = 200$$

$$I_{\text{rms}} = \sqrt{200} = 10 \text{ A}$$

$$\text{Power factor} = \frac{R}{Z} = 0.6 \Rightarrow Z = \frac{6}{0.6} = 10 \Omega$$

$$Z^2 = R^2 + X_L^2 \Rightarrow X_L = \sqrt{Z^2 - R^2}$$

$$X_L = \sqrt{100 - 36} = \sqrt{64} \Rightarrow X_L = 8 \Omega$$

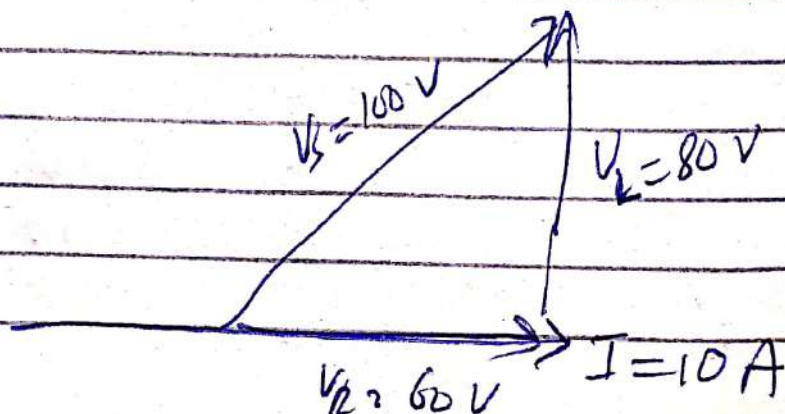
$$\cos \theta = 0.6$$

$$\Rightarrow \theta = \cos^{-1}(0.6) = 53.1^\circ$$

$$V_{\text{rms}} = I_{\text{rms}} Z = 10 \times 10 = 100 \text{ V}$$

$$V_R = I_{\text{rms}} R = 60 \text{ V}$$

$$V_L = I_{\text{rms}} X_L = 80 \text{ V}$$

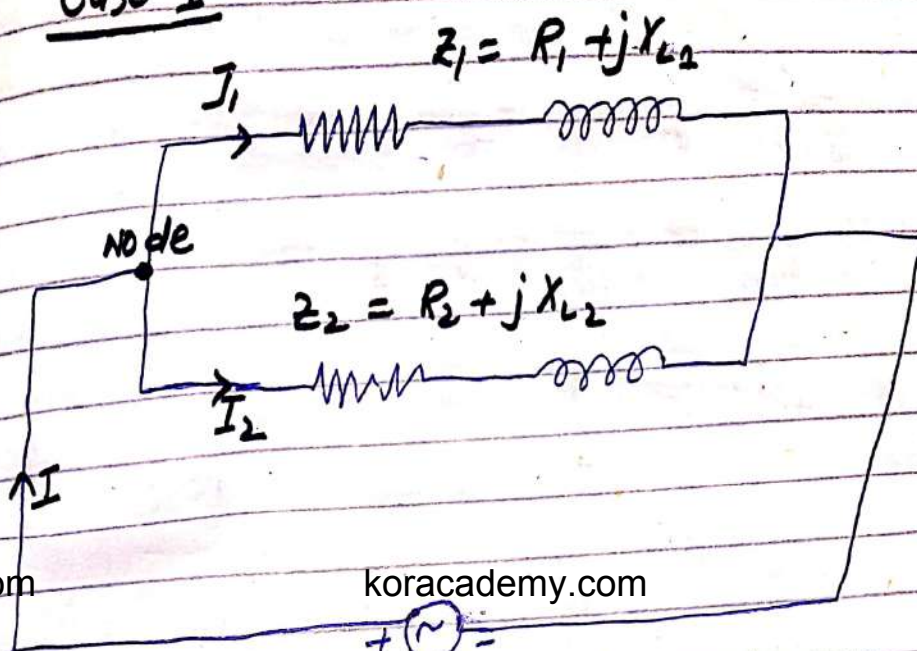




# Parallel Circuits.

## Impedance method

### Case 1



$$+ \text{ } \ominus =$$

$$V_s = V_{rms} \angle 0^\circ \rightarrow \text{polar}$$

$$V_s = V_{rms} \rightarrow \text{rectangular}$$

$$Z_1 = |Z_1| \angle \theta_1 \quad Z_2 = |Z_2| \angle \theta_2$$

$$I_1 = \frac{V_s}{Z_1} = \frac{V_{rms} \angle 0^\circ}{|Z_1| \angle \theta_1}$$

$$I_1 = \frac{V_{rms}}{|Z_1|} \angle -\theta_1 = I_{rms} \angle -\theta_1$$

In rectangular  $I_1 = I_{a1} - j I_{b1}$

Similarly  $I_2 = \frac{V}{Z_2} = \frac{V_{rms} \angle 0^\circ}{|Z_2| \angle \theta_2}$

$$I_2 = \frac{V_s}{|Z_2|} \angle -\theta_2 = \frac{I_2}{\text{rms}} \angle -\theta_2$$

In rectangular form

$$I_2 = \frac{I}{a_2} - jI_{b2}$$

Apply KCL to the only node of the circuit.

$$I = I_1 + I_2$$

$$I = I_{a1} - jI_{b1} + I_{a2} - jI_{b2}$$

$$I = (I_{a1} + I_{a2}) - j(I_{b1} + I_{b2})$$

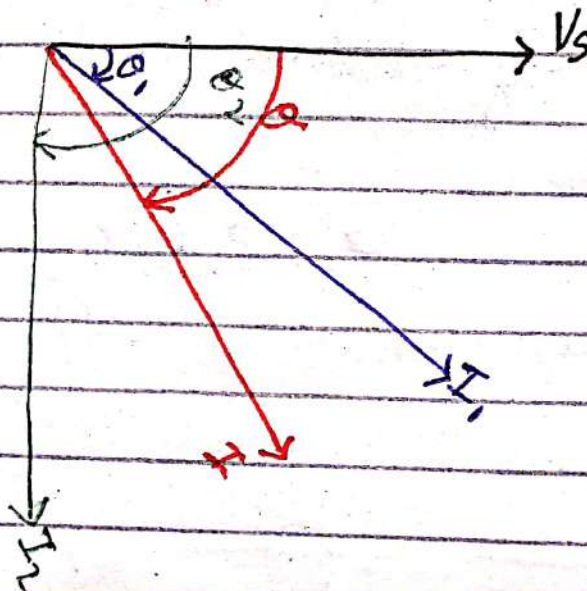
$$I = I_{\text{rms}} \angle -\theta$$

$$P.f_1 = \frac{R_1}{|Z_1|} = \cos \theta_1$$

Lagging power factor.

$$P.f_2 = \frac{R_2}{|Z_2|} = \cos \theta_2$$

$$P.f = \cos \theta$$



The power taken by the two branches:

$$P_1 = \frac{I_1^2}{\text{rms}} R_1 = \frac{V_s}{\text{rms}} \times \frac{I_1}{\text{rms}} \times \cos \theta_1$$

$$P_2 = \frac{I_2^2}{\text{rms}} R_2 = \frac{V_s}{\text{rms}} \times \frac{I_2}{\text{rms}} \times \cos \theta_2$$

$$P_s = P_t = P_1 + P_2$$

Total power delivered by source

$$P_s = \frac{V_s}{\text{rms}} \times I_{\text{rms}} \times \cos \theta$$

### Instantaneous Expressions

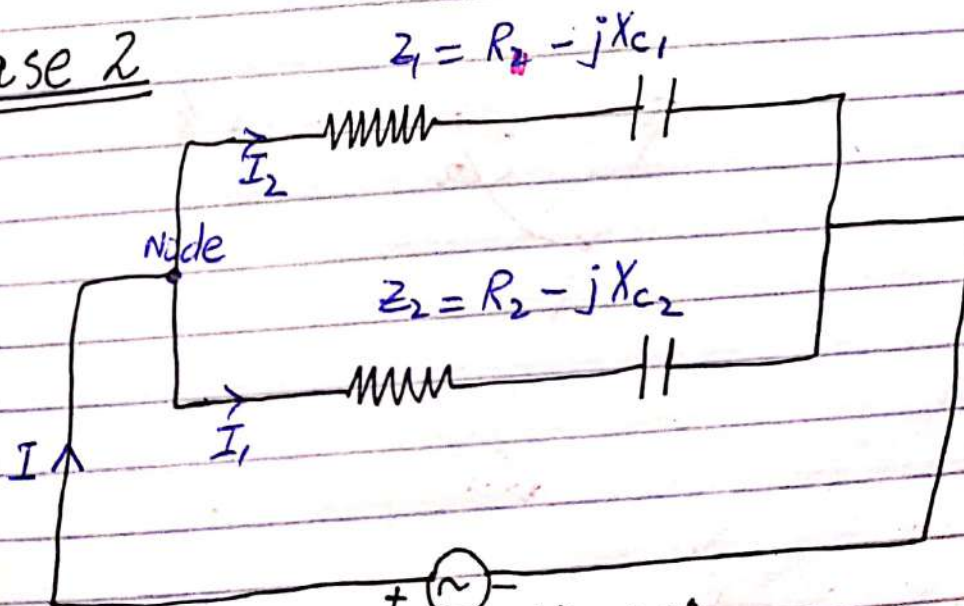
$$v_s = \sqrt{2} V_{\text{rms}} \sin(\omega t + 0^\circ)$$

$$i_1 = \sqrt{2} \frac{I_1}{\text{rms}} \sin(\omega t - \theta_1)$$

$$i_2 = \sqrt{2} \frac{I_2}{\text{rms}} \sin(\omega t - \theta_2)$$

$$i = \sqrt{2} \frac{I}{\text{rms}} \sin(\omega t - \theta)$$

### Case 2



Polar  $V_s = V_s \angle 0^\circ$   
 Rectangular  $V_s = V_s \text{ rms}$

$$Z_1 = |Z_1| \angle -\theta_1, \quad Z_2 = |Z_2| \angle -\theta_2$$

$$I_1 = \frac{V_s}{Z_1} = \frac{V_{s,rms} \angle 0^\circ}{|Z_1| \angle -\theta_1}$$

$$I_1 = \frac{V_{s,rms}}{|Z_1|} \angle +\theta_1 = \frac{I_1}{rms} \angle +\theta_1$$

In rectangular form  $I_1 = I_{a1} + j I_{b1}$

Similarly  $I_2 = \frac{V_s}{Z_2} = \frac{V_{s,rms} \angle 0^\circ}{|Z_2| \angle -\theta_2}$

$$I_2 = \frac{V_{s,rms}}{|Z_2|} \angle +\theta_2 = \frac{I_2}{rms} \angle +\theta_2$$

$$I_2 = I_{a2} + j I_{b2}$$

Apply KCL to the only node.

$$I = I_1 + I_2$$

$$I = I_{a1} + j I_{b1} + I_{a2} + j I_{b2} = (I_{a1} + I_{a2}) + j (I_{b1} + I_{b2})$$

$$I = I_{rms} \angle +\theta$$

$$P.f_1 = \frac{R_1}{|Z_1|} = \cos \theta_1$$

$$P.f_2 = \frac{R_2}{|Z_2|} = \cos \theta_2$$

$$P.f = \cos \theta \quad \text{leading.}$$

$$P_1 = I_{rms}^2 R_1 = V_s \times I_1 \times \cos \theta_1$$

$$P_2 = I_{rms}^2 R_2 = V_s \times I_2 \times \cos \theta_2$$

$$P_s = P_t = P_1 + P_2$$

$$P_s = V_s \times I_{rms} \times \cos \theta$$

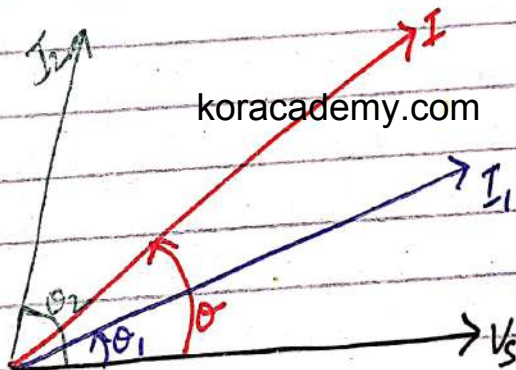
Instantaneous expressions

$$v_s = \sqrt{2} V_s \sin(\omega t + 0^\circ)$$

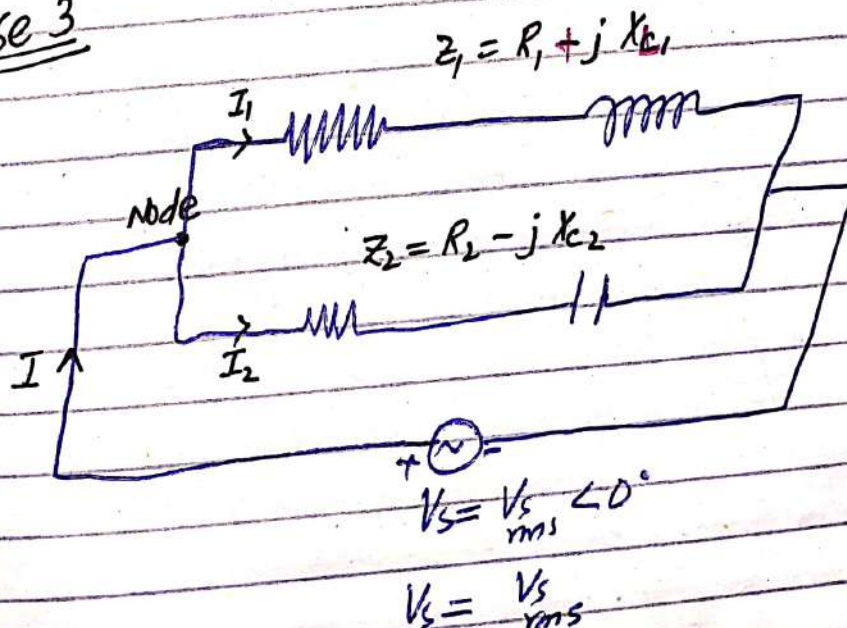
$$i_1 = \sqrt{2} I_{1rms} \sin(\omega t + \theta_1)$$

$$i_2 = \sqrt{2} I_{2rms} \sin(\omega t + \theta_2)$$

$$i = \sqrt{2} I_{rms} \sin(\omega t + \theta)$$



Case 3



$$Z_1 = |Z_1| \angle \theta_1 \quad Z_2 = |Z_2| \angle \theta_2$$

$$I_1 = \frac{V_s}{Z_1} = \frac{V_{s_{rms}} \angle 0^\circ}{|Z_1| \angle \theta_1} = \frac{V_{s_{rms}}}{|Z_1|} \angle -\theta_1 = \frac{I_1}{rms} \angle -\theta_1$$

Rectangular  $I_1 = I_{a1} - jI_{b1}$

$$I_2 = \frac{V_s}{Z_2} = \frac{V_{s_{rms}} \angle 0^\circ}{|Z_2| \angle \theta_2} = \frac{V_{s_{rms}}}{|Z_2|} \angle +\theta_2 = \frac{I_2}{rms} \angle +\theta_2$$

Rectangular  $I_2 = I_{a2} + jI_{b2}$

Apply KCL  $I = I_1 + I_2$

$$I = I_{a1} - jI_{b1} + I_{a2} + jI_{b2}$$

$$\vec{I} = (I_{a1} + I_{a2}) - j(I_{b1} - I_{b2})$$

$$I = I_{rms} \angle \pm \theta$$

$$P.f_1 = \frac{R_1}{|Z_1|} \cos \theta_1 \quad P.f_2 = \frac{R_2}{|Z_2|} \cos \theta_2$$

$$P.f = \cos \theta$$

$$P_1 = I_{rms}^2 R_1 = \frac{V_s}{rms} \times \frac{I_1}{rms} \times \cos \theta_1$$

$$P_2 = I_{rms}^2 R_2 = \frac{V_s}{rms} \times \frac{I_2}{rms} \times \cos \theta_2$$

$$P_s = P_t = P_1 + P_2$$

$$P_s = \frac{V_s}{rms} \times I_{rms} \times \cos \theta$$

## Instantaneous Expressions.

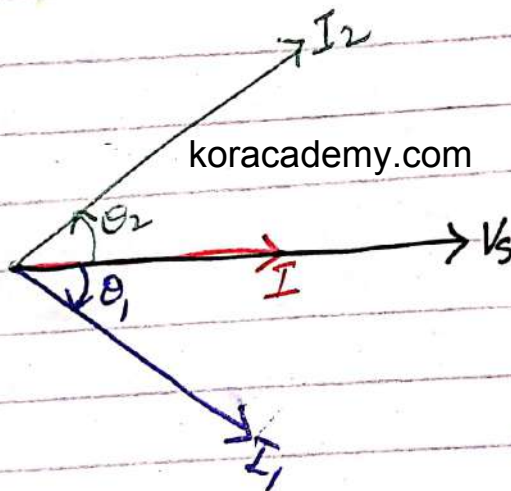
$$v_s = \sqrt{2} \frac{V_s}{\text{rms}} \sin(\omega t + 0^\circ)$$

$$i_1 = \sqrt{2} \frac{I_1}{\text{rms}} \sin(\omega t - \theta_1)$$

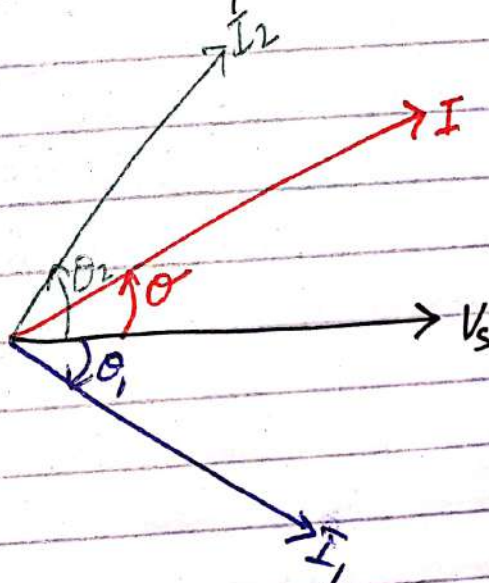
$$i_2 = \sqrt{2} \frac{I_2}{\text{rms}} \sin(\omega t + \theta_2)$$

$$\phi \quad i = \sqrt{2} I_{\text{rms}} \sin(\omega t \pm \theta)$$

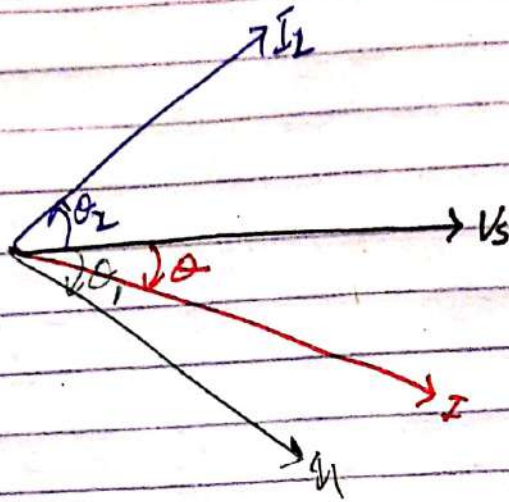
if  $\theta = 0^\circ \Rightarrow$  Resistive circuit.



if  $\theta = +ve \Rightarrow$  capacitive circuit



if  $\theta = -ve \Rightarrow$  inductive circuit

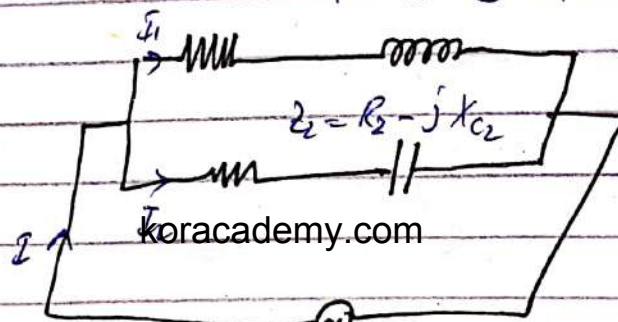


### Parallel Circuits By Admittance method.

(Case 1+2  $\rightarrow$  H.W)

$$Z_1 = R_1 + jX_{L1}$$

Case 3



$$V_s = \frac{V_s}{ms} \angle 0^\circ$$

$$V_s = V_{ms}$$

$$I_1 = \frac{V_s}{Z_1} = V_s \times \frac{1}{Z_1}$$

$$\frac{1}{Z_1} \rightarrow \text{Admittance} = Y \quad I_1 = V_s Y_1 \rightarrow \textcircled{1}$$

$$\frac{1}{Z_1} = \frac{1}{R + jX_L} \times \frac{R - jX_C}{R - jX_C}$$

$$Y_1 = \frac{R_1}{R_1^2 + X_{L1}^2} - j \frac{X_{L1}}{R_1^2 + X_{L1}^2}$$

$$Y_1 = G_1 - jB_1$$



$$\textcircled{1} \Rightarrow I_1 = V_{s, \text{rms}} (G_1 - jB_1)$$

$$I_1 = I_{G1} - jI_{B1}$$

$$I_1 = I_1 \angle -\theta_1$$

$$I_2 = V_2 \times \frac{1}{Z_2} = V_2 Y_2 \rightarrow \textcircled{2}$$

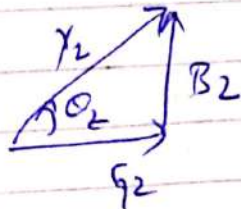
$$V_2 = \frac{1}{Z_2} = \frac{1}{R_2 - jX_{C2}} \times \frac{R_2 + jX_{C2}}{R_2 + jX_{C2}}$$

$$Y_2 = \frac{R_2}{R_2^2 + X_{C2}^2} + j \frac{X_{C2}}{R_2^2 + X_{C2}^2}$$

$$Y_2 = G_2 + jB_2$$

conductance

susceptance



$$\textcircled{2} \Rightarrow I_2 = V_{s, \text{rms}} (G_2 + jB_2)$$

$$I_2 = I_{G2} + jI_{B2}$$

$$I_2 = I_2 \angle \theta_2$$

Apply KCL

$$I_2 = I_1 + I_2$$

$$I_2 = I_{G1} - jI_{B1} + I_{G2} + jI_{B2}$$

$$I_2 = (I_{G1} + I_{G2}) + j(-I_{B1} + I_{B2})$$

$$I = I_{rms} \angle \theta$$

$$P.f_1 = \frac{Z_1}{|R_1|} \cos \theta_1 \quad P.f_2 = \frac{Z_2}{|R_2|} \cos \theta_2$$

$$P.f = \cos \theta$$

$$P_1 = I_{rms}^2 R_1 = \frac{V_s}{rms} \times I_{rms} \times \cos \theta_1$$

$$P_2 = I_{rms}^2 R_2 = \frac{V_s}{rms} \times I_{rms} \times \cos \theta_2$$

$$P_s = P_t = P_1 + P_2$$

$$P_s = \frac{V_s}{rms} \times I_{rms} \times \cos \theta$$

Instantaneous expressions

$$V_s = \sqrt{2} \frac{V_s}{rms} \sin(\omega t + 0^\circ)$$

$$i_1 = \sqrt{2} \frac{I_1}{rms} \sin(\omega t - \theta_1)$$

$$i_2 = \sqrt{2} \frac{I_2}{rms} \sin(\omega t + \theta_2)$$

$$I = \sqrt{2} \frac{I}{rms} \sin(\omega t + \theta)$$

Example ex 3.

$$Z_1 = 3 + j4 \Omega$$

$$Z_2 = 3 - j4 \Omega$$

$$V_s = 100 \angle 0^\circ \quad V_s = 100 V$$

$$Z_1 = 5 \angle 53.1^\circ \Omega$$

$$Z_2 = 5 \angle -53.1^\circ \Omega$$

$$I_1 = \frac{V_s}{Z_1} = \frac{100}{5 \angle 53.1^\circ}$$

$$I_1 = 20 \angle -53.1^\circ$$

$$I_1 = 12 - j16 \text{ A}$$

$$I_2 = \frac{V_s}{Z_2} = \frac{100}{5 \angle -53.1^\circ} = 20 \angle 53.1^\circ \text{ A}$$

$$I_2 = 12 + j16 \text{ A}$$

By Admittance

$$Y_1 = \frac{3}{25} - j \frac{4}{25} \text{ S}$$

$$I_1 = V_s Y_1 = 100 \left( \frac{3}{25} - j \frac{4}{25} \right)$$

$$I_1 = 12 - j16 \text{ A} = 20 \angle -53.1^\circ \text{ A}$$

$$Y_2 = \frac{3}{25} + j \frac{4}{25} \text{ S}$$

$$I_2 = V_s Y_2 = 100 \left( \frac{3}{25} + j \frac{4}{25} \right)$$

$$I_2 = 12 + j16 \text{ A} = 20 \angle 53.1^\circ \text{ A}$$

$$I_2 \text{ or } I_1 + I_2 = 24 \text{ A}$$

$$P.f_1 = \cos \theta_1 = \frac{3}{5} = 0.6$$

$$P.f_2 = \cos \theta_2 = \frac{3}{5} = 0.6$$

$$P.f = \cos 0^\circ = 1$$

$$P_1 = \frac{I_1^2}{w} R_1 = 1200 \text{ W}$$

$$P_2 = I_2^2 R_2 = 1200 \text{ W}$$

$$P_3 = P_1 + P_2 = 2400 \text{ W}$$

Instantaneous expressions

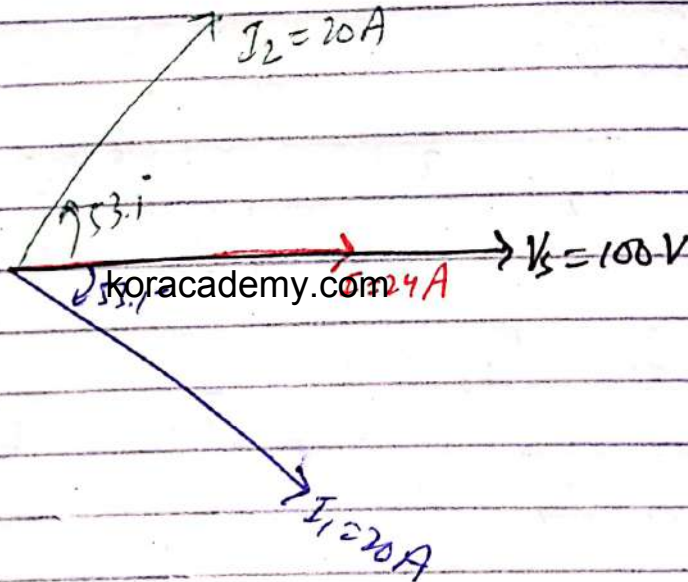
Let  $f = 50 \text{ Hz}$

$$V_s = 100\sqrt{2} \sin(100\pi t)$$

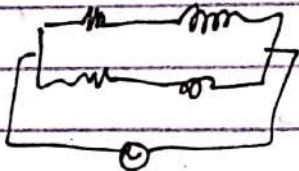
$$i_1 = 20\sqrt{2} \sin(100\pi t - 53.1^\circ)$$

$$i_2 = 20\sqrt{2} \sin(100\pi t + 53.1^\circ)$$

$$i = 24\sqrt{2} \sin(100\pi t + 0)$$



Example 2. Case a



$$Z_1 = 3 + j4 \Omega \quad Z_2 = 3 + j4 \Omega$$

$$V_s = 100 \angle 0^\circ = 100 \text{ V}$$

$$Z_1 = 5 \angle 53.1^\circ \Omega \quad Z_2 = 5 \angle 53.1^\circ \Omega$$

$$I_1 = \frac{V_s}{Z_1} = \frac{100}{5 \angle 53.1^\circ} = 20 \angle -53.1^\circ \text{ A}$$

$$I_2 = 12 - j16 \text{ A}$$

$$I_2 = \frac{V_s}{Z_2} = \frac{100}{5 \angle 53.1^\circ} = 20 \angle -53.1^\circ \text{ A}$$

$$I_1 = 12 - j16 \text{ A}$$

Admittance method

$$Y_1 = \frac{3}{25} - j \frac{4}{25} \text{ S}$$

$$I_1 = V_s Y_1 = 100 \left( \frac{3}{25} - j \frac{4}{25} \right)$$

$$I_1 = 12 - j16 \text{ A} = 20 \angle -53.1^\circ \text{ A}$$

$$Y_2 = \frac{3}{25} - j \frac{4}{25} \text{ S}$$

$$I_2 = V_s Y_2 = 100 \left( \frac{3}{25} - j \frac{4}{25} \right)$$

$$I_2 = 12 - j16 \text{ A} = 20 \angle -53.1^\circ \text{ A}$$

$$I_2 \text{ or } I_1 + I_2 = 24 - j32 \text{ A} = 40 \angle -53.1^\circ \text{ A}$$

$$P.f. = \cos \theta = \frac{3}{5} = 0.6 \text{ A} = \cos \theta_L = P.f.L$$

$$P.f. = \cos 53.1^\circ = 0.6$$

$$P_1 = I_1^2 R_1 = 1200 \text{ W}$$

$$P_2 = I_2^2 R_2 = 1200 \text{ W}$$

$$P_s = P_1 + P_2 = 2400 \text{ W}$$

Instantaneous expressions

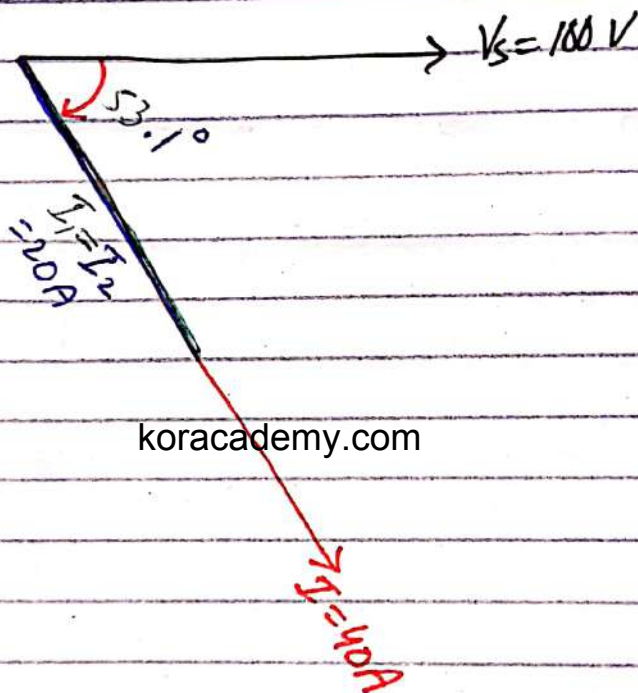
$$\text{if } f = 80 \text{ Hz}$$

$$v_s = 100\sqrt{2} \sin(100\pi t)$$

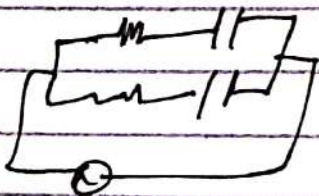
$$i_f = 20\sqrt{2} \sin(\omega t - 53.1^\circ)$$

$$i_v = 20\sqrt{2} \sin(100\pi t - 53.1^\circ)$$

$$i = 24\sqrt{2} \sin(100\pi t - 53.1^\circ)$$



Example 3 Case 2.



$$z_1 = 3 - j4 \Omega \quad z_2 = 3 - j4 \Omega$$

$$V_s = 100 \angle 0^\circ = 100 \text{ V} \quad f = 50 \text{ Hz}$$

$$Z_1 = 5 \angle -53.1^\circ$$

$$Z_2 = 5 \angle -53.1^\circ$$

$$I_1 = \frac{V_s}{Z_1} = \frac{100}{5 \angle -53.1^\circ} = 20 \angle 53.1^\circ$$

$$I_2 = 12 + 16j \text{ A}$$

$$I_1 = \frac{V_s}{Z_1} = \frac{100}{5 \angle -53.1^\circ} = 20 \angle 53.1^\circ$$

$$I_1 = 12 + j16 \text{ A}$$

Admittance

$$Y_1 = \frac{3}{25} + j \frac{4}{25} \text{ S}$$

$$I_1 = V_s Y_1 = 100 \left( \frac{3}{25} + j \frac{4}{25} \right) = 12 + j16 \text{ A} \\ = 20 \angle 53.1^\circ \text{ A}$$

$$Y_2 = \frac{3}{25} + j \frac{4}{25} \text{ S}$$

$$I_2 = V_s Y_2 = 100 \left( \frac{3}{25} + j \frac{4}{25} \right)$$

$$I_2 = 12 + j16 \text{ A} = 20 \angle 53.1^\circ \text{ A}$$

$$I = I_1 + I_2 = 24 + j32 = 40 \angle 53.1^\circ$$

$$P.f. = \cos \theta_1 = \frac{3}{5} = 0.6$$

$$P.f. = \cos \theta_2 = \frac{3}{5} = 0.6$$

$$P.f. = \cos 53.1^\circ = 0.6$$

$$P_1 = I_1^2 R_1 = 1200 \text{ W}$$

$$P_2 = I_2^2 R_2 = 1200 \text{ W}$$

$$P_s = P_1 + P_2 = 2400 \text{ W}$$

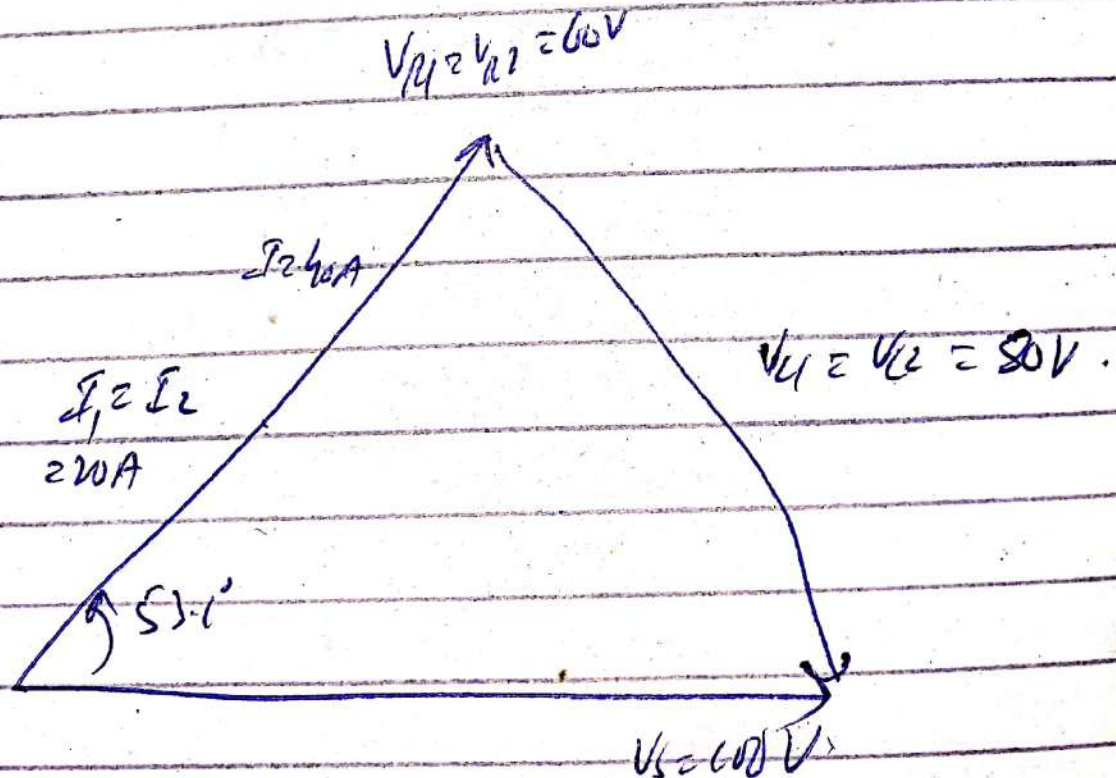
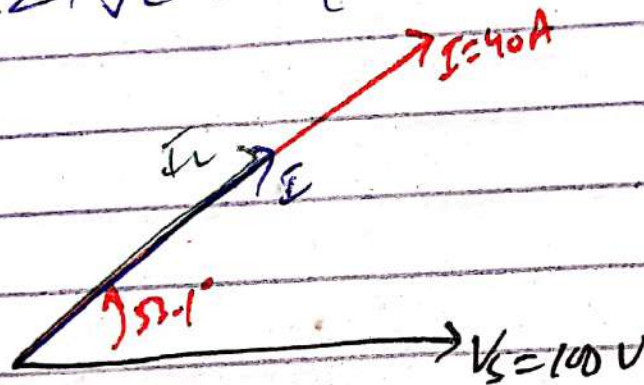
Instans

$$V_s = 100 \sqrt{2} \sin(100\pi t)$$

$$i_1 = 20 \sqrt{2} \sin(100\pi t + 53.1^\circ)$$

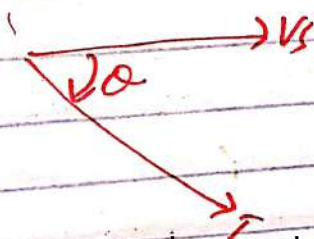
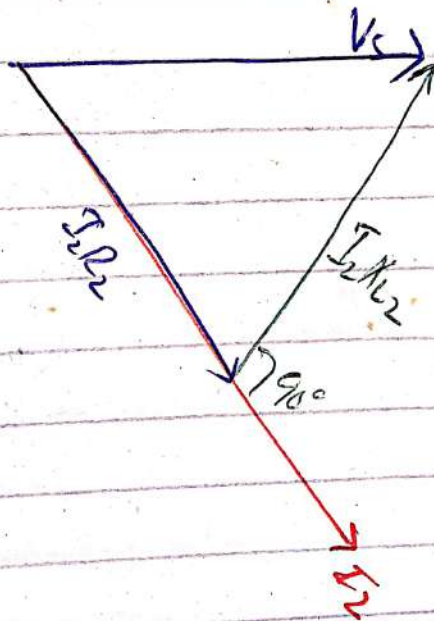
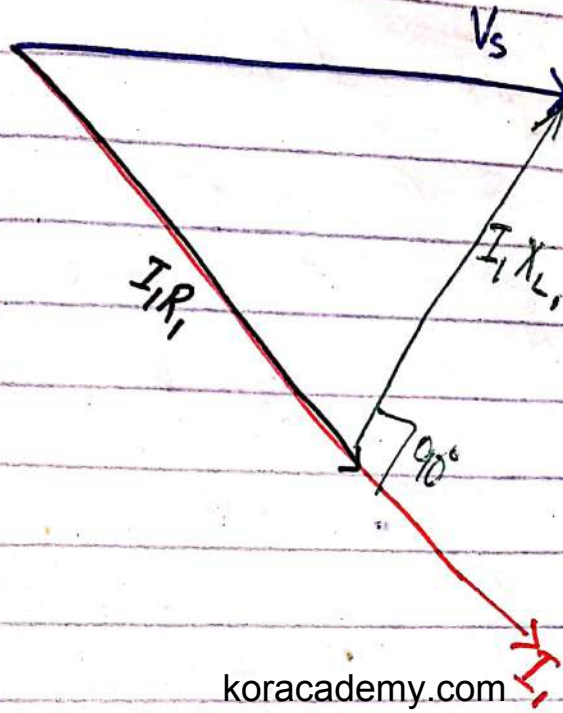
$$i_2 = 20 \sqrt{2} \sin(100\pi t + 53.1^\circ)$$

$$i = 24 \sqrt{2} \sin(100\pi t + 53.1^\circ)$$





for an inductive circuit



## magnetic Circuits

magnetic circuits are analogous to electrical circuits.

magnetic flux  $\phi$  is analogous to current.

Reluctance  $R$  is analogous to resistance  $R$ .

↳ the property of a magnetic material due to which it opposes the flow of magnetic flux.

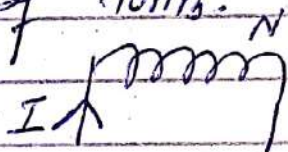
$$R = \frac{l}{\mu A}$$

$$R = \frac{l}{\mu A}$$

Permeability  $\mu$  of a magnetic material is analogous to conductivity.

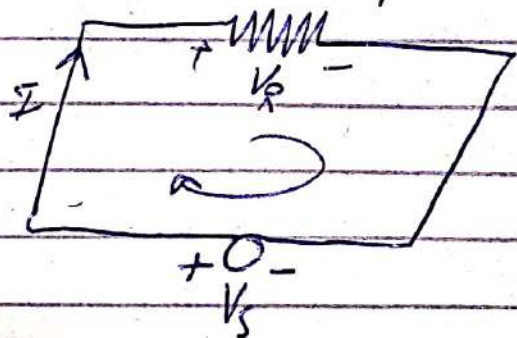
magnetic motive force (MMF) is analogous to voltage  $V$  in electrical circuits.

For generation of magnetic flux, we need to have a current in a loop of  $N$  number of turns.



$$\text{MMF} = F = NI \quad (\text{AT})$$

### Review of An Electrical circuit



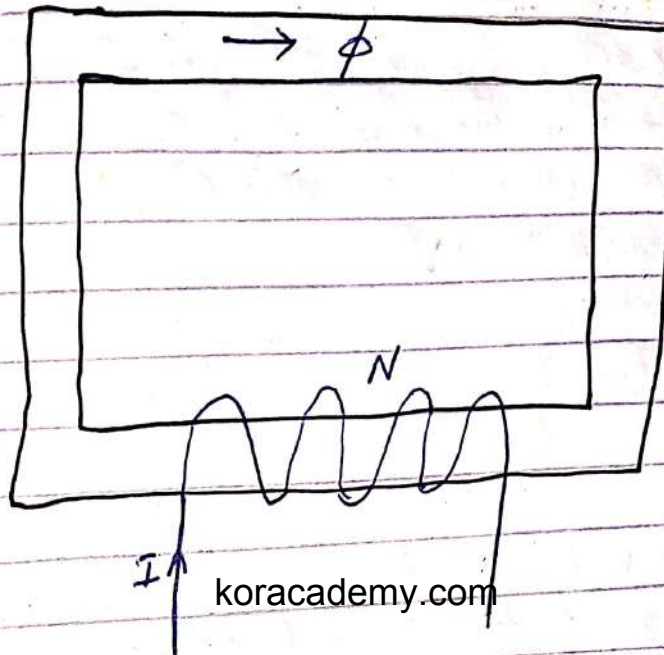
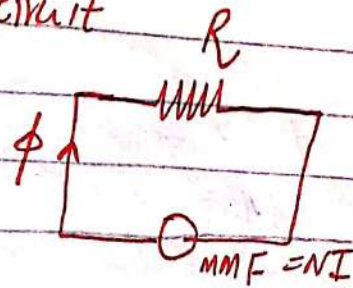
$$V_s = V_R$$

$$V_s = V_R = IR$$

$$I = \frac{V_s}{R}$$

Consider a magnetic material.

→ Equivalent circuit



mean length =  $l$  Area =  $A$  permeability =  $\mu$ .  
Strength of magnetic field  $H$ .

$$H \propto \frac{NI}{l}$$

$$H = \frac{NI}{l}$$

$$\Rightarrow \boxed{\text{MMF} = NI = Hl}$$

magnetic flux density is the flux  
per unit area  $\rightarrow B$

$$B = \mu H$$

$$\boxed{H = \frac{B}{\mu}}$$

$$B = \frac{\phi}{A} \Rightarrow H = \frac{\phi}{\mu A}$$

$$\Rightarrow \text{MMF} = NI = \frac{\phi \times l}{\mu A}$$

Reluctance  $R = \frac{l}{\mu A}$

$$\Rightarrow \text{MMF} = \phi R$$

$$\phi = \frac{\text{MMF}}{R}$$

Example

$$l = 1\text{m} \quad A = 0.5\text{m}^2 \quad \mu_r = 100 \quad N = 1000$$

$$\phi = 2 \times 10^{-3} \text{ Wb} \quad \text{What is } I?$$

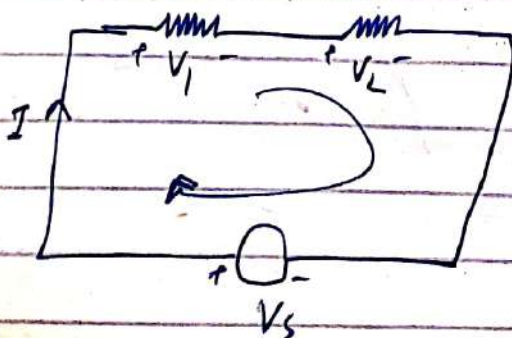
koracademy.com

$$R = \frac{l}{\mu A} = \frac{1}{4 \times 10^{-7} \times 100 \times 0.5}$$

$$\Rightarrow R = 1.591 \times 10^4 \text{ AT/Wb}$$

$$I = \frac{\phi R}{N} = \frac{2 \times 10^{-3} \times 1.591 \times 10^4}{1000} = 31.8 \text{ mA}$$

## Review Series Electrical Circuit



$$\text{KVL } V_s = V_1 + V_2$$

$$V_1 = IR_1 \quad V_2 = IR_2$$

$$V_s = I(R_1 + R_2)$$

where

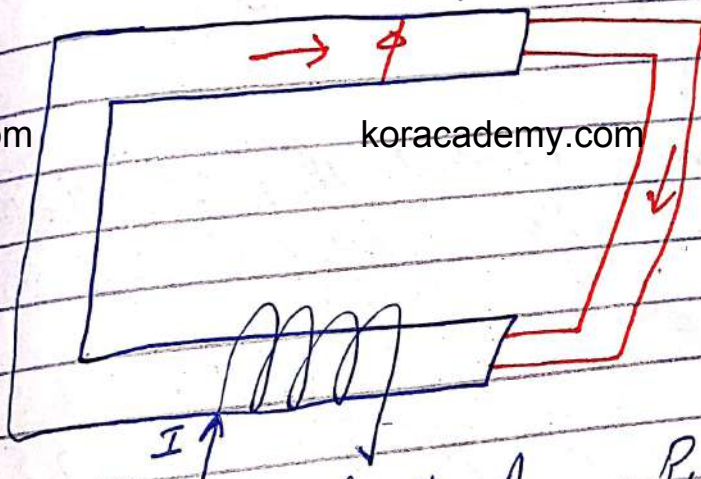
$$R_T = R_1 + R_2$$

# Series Magnetiz Circuit

Consider a circuit of two different materials.  
Put a coil of  $N$  turns on any material.

Similar direction of flux from more than one coil  $\rightarrow$  +ve sign  $\rightarrow$  add.  
If in opposite direction  $\rightarrow$  -ve sign  $\rightarrow$  subtract.

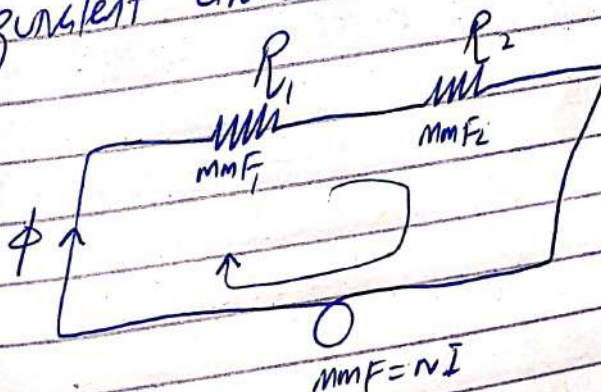
In this case the same flux will flow in both the materials.



Material 1  $\Rightarrow l_1, \mu_1, A_1$       $P_1 = \frac{l_1}{\mu_1 A_1}$

Material 2  $\Rightarrow l_2, \mu_2, A_2$       $P_2 = \frac{l_2}{\mu_2 A_2}$

Equivalent circuit



$$\text{MMF} = \text{MMF}_1 + \text{MMF}_2$$

$$\text{MMF}_1 = \phi R_1$$

$$\text{MMF}_2 = \phi R_2$$

$$\text{MMF} = NI = \phi R_1 + \phi R_2 = \phi (R_1 + R_2)$$

$$\text{where } R_T = R_1 + R_2$$

Example 4

$$l_1 = 1 \text{ m} \quad A_1 = 0.5 \text{ m}^2 \quad \mu_{r1} = 100$$

$$R_1 = 1.591 \times 10^4 \text{ AT/Wb} \quad l_2 = 1 \text{ m} \quad A_2 = 0.5 \text{ m}^2$$

$$\mu_{r2} = 1000$$

$$R_2 = \frac{l_2}{\mu_0 \mu_{r2} A_2} = \frac{1}{4 \times 10^{-7} \times 1000 \times 0.5}$$

$$R_2 = 1.591 \times 10^3 \text{ AT/Wb}$$

$$\phi = 3 \times 10^{-3} \text{ Wb} \quad I = 1 \text{ A} \quad N = ?$$

$$\text{MMF} = NI = \phi (R_1 + R_2)$$

$$\Rightarrow N = \frac{\phi (R_1 + R_2)}{I}$$

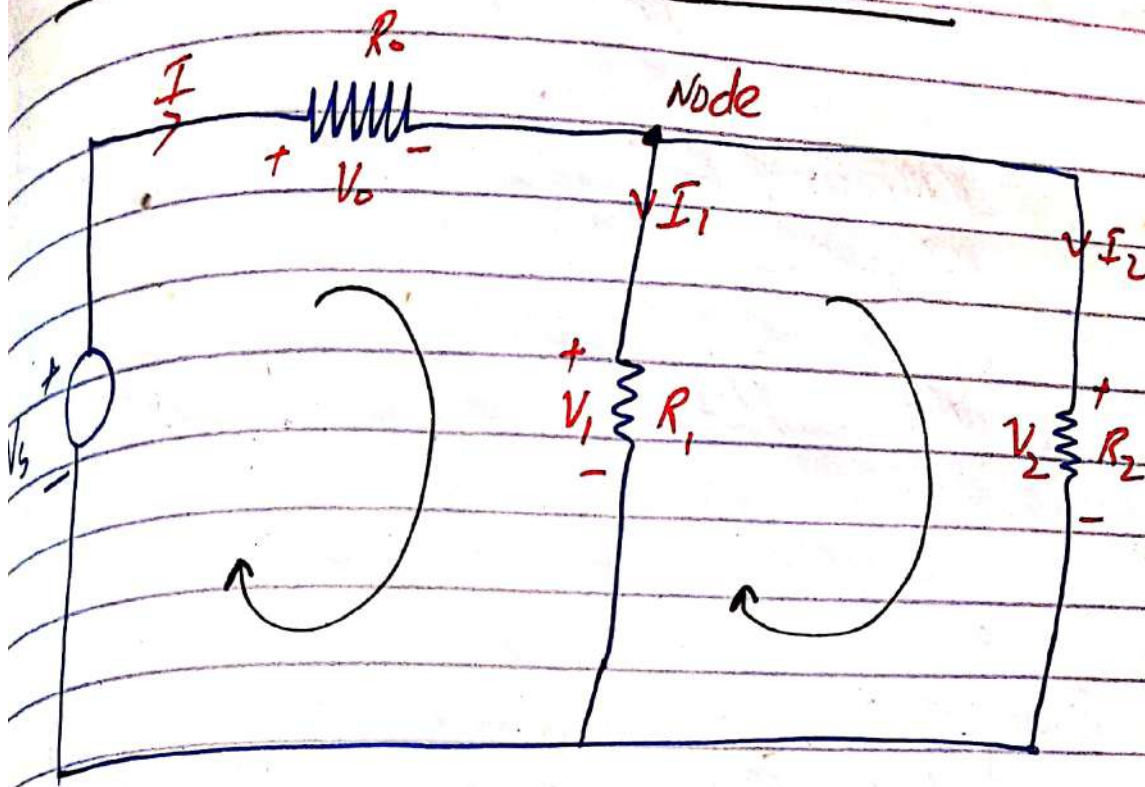
$$N = 3 \times 10^{-3} (1.59 \times 10^4 + 1.59 \times 10^3)$$

$$N = 53 \text{ turns.}$$

We don't have simple parallel magnetic circuits.

To understand series parallel magnetic circuits we first study series parallel electric circuits

# Series Parallel Electric Circuit



Apply KVL to loop 1

$$V_s = V_0 + V_2$$

KVL to loop 2

$$V_1 = V_2$$

$$\Rightarrow V_s = V_0 + V_2$$

$$V_0 = I R_0$$

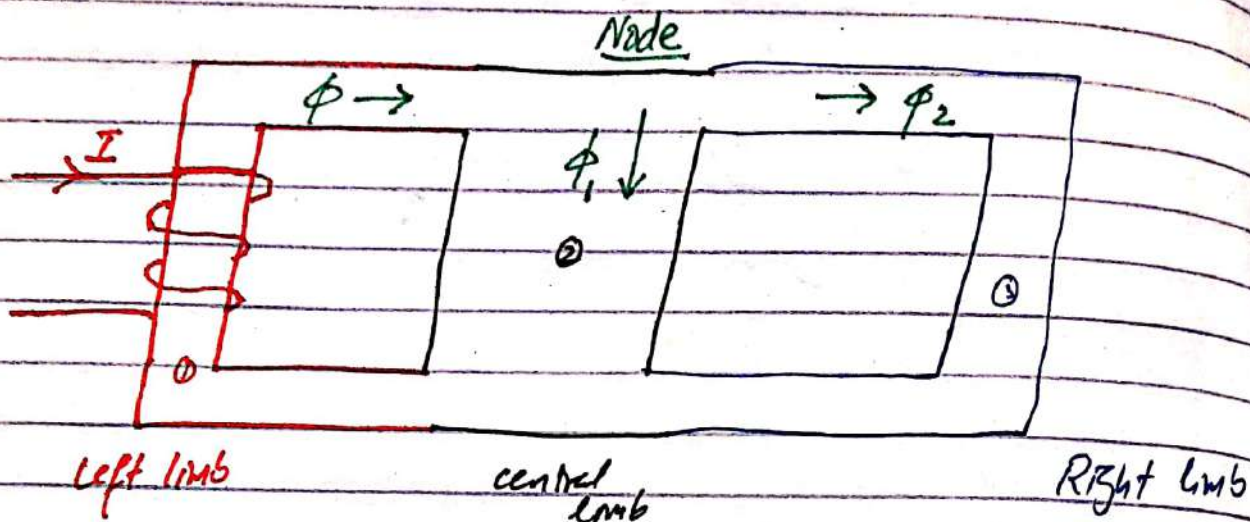
$$V_1 = I_1 R_1$$

$$V_2 = I_2 R_2$$

Apply KCL to the only node

$$I = I_1 + I_2$$

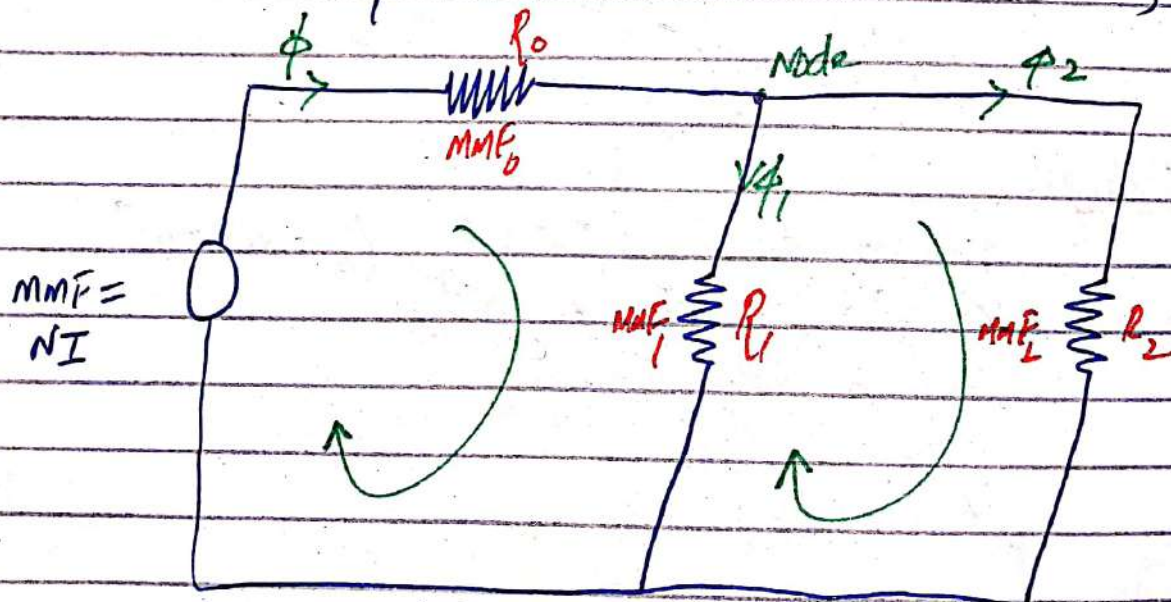
# Series Parallel Magnetic Circuit



we divide this circuit in three sections.

Left limb  $\Rightarrow \mu_0, A_0, l_0, R_0 = l_0 / \mu_0 A_0$   
 Right limb  $\Rightarrow \mu_2, A_2, l_2, R_2 = l_2 / \mu_2 A_2$   
 central limb  $\Rightarrow \mu_1, A_1, l_1, R_1 = l_1 / \mu_1 A_1$

An equivalent circuit to the above is;



Apply a law equivalent to KVL; to loop I.

$$MMF = MMF_0 + MMF_1$$

analogous to  $V_s = V_0 + V_1$



loop 2  $MMF_1 = MMF_2$   
 analogous to  $V_1 = V_2$

$$V_3 = V_0 + V_2$$

$$MMF = MMF_0 + MMF_1$$

$$MMF_0 = \phi R_0$$

analogous to  $V_0 = I R_0$

$$MMF_1 = \phi_1 R_1$$

analogous to  $V_1 = I_1 R_1$

$$MMF_2 = \phi_2 R_2$$

analogous to  $V_2 = I_2 R_2$

Apply KCL to the wire

Apply

$$\phi = \phi_1 + \phi_2$$

analogous to  $I = I_1 + I_2$

Example

$N = 200$  turns  $I = 2A$   $P_0 = 25 \times 10^{-3}$   
 $\phi = 8 \times 10^{-3}$  wb  $\phi_1 = 4 \times 10^{-3}$  wb  
 calculate  $R_1, R_2, \phi_2$

$$\text{As } \phi = \phi_1 + \phi_2 \Rightarrow \phi_2 = \phi - \phi_1$$

$$\Rightarrow \phi_2 = 4 \times 10^{-3} \text{ wb}$$

$$\text{As } MMF = NI = MMF_0 + MMF_1$$

$$NI = \phi R_0 + \phi_1 R_1$$

$$R_1 = \frac{NI - \Phi R_0}{\Phi_1} = \frac{200 \times 2 - 8 \times 10^{-3} \times 25 \times 10^3}{7 \times 10^{-3}}$$

$$R_1 = 50 \times 10^3 \text{ AT/Wb}$$

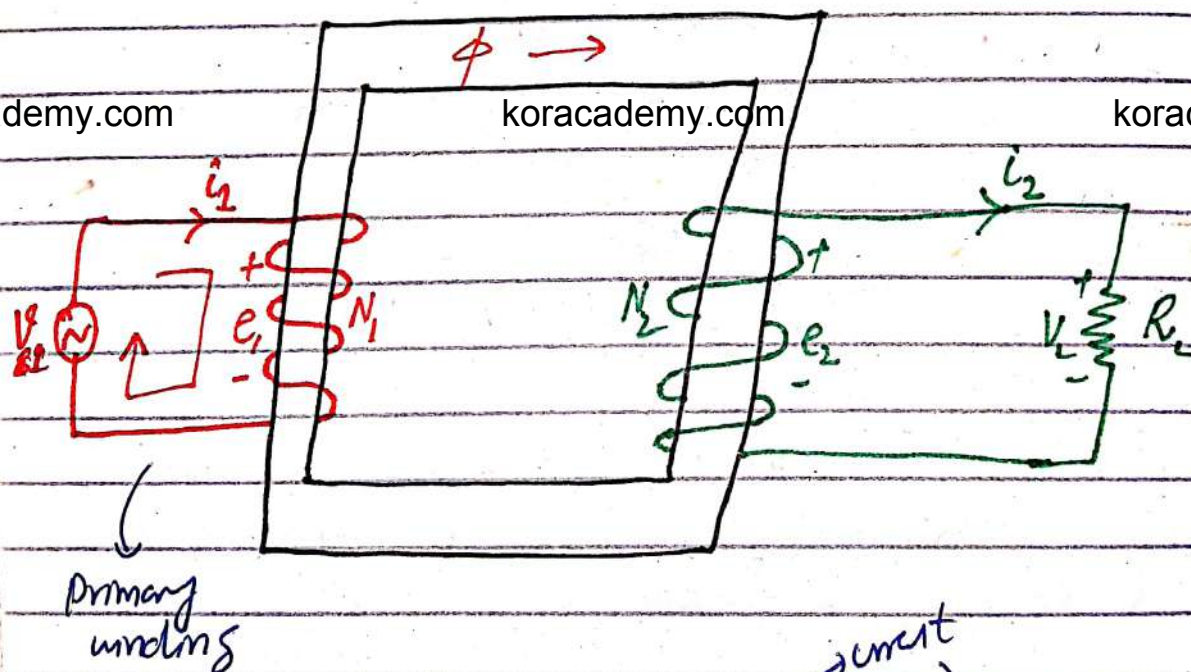
$$\text{As } MMF_2 = MMF_1$$

$$\Phi_2 R_2 = \Phi_1 R_1$$

$$R_2 = R_1 = 50 \times 10^3 \text{ AT/Wb.}$$

## Single Phase AC Transformer

non magnetic material  $\rightarrow$  whose relative permeability  $\mu_r \approx 1$ .  $\rightarrow$  because its reluctance is very high.



AC voltage  $\rightarrow$  primary coil  $\rightarrow$  magnetic flux  $\rightarrow$  are

The nature of flux depends on the nature of the current.

The current in the winding will induce a voltage through the winding.

$$i_1 = I_m \sin \omega t$$

$$\phi = \phi_m \sin \omega t$$

KVL  $v_1 = e_1$

The induced voltage across the winding according to Faraday's law is

$$v_1 = e_1 = \frac{N_1 d\phi}{dt}$$

$$v_1 = \omega N_1 \phi_m \cos \omega t$$

$$v_1 = 2\pi f \phi_m N_1 \sin(\omega t + \pi/2)$$

$$V_{m1} = 2\pi f \phi_m N_1$$

$$V_{rms} = 0.707 \times 2\pi f \times N_1 \times \phi_m$$

$$V_{rms} = 4.44 f N_1 \phi_m$$

Secondary winding is placed in a time varying magnetic flux  $\rightarrow$  so voltage is induced according to Faraday's law.

if load is connected  $\rightarrow v_2, i_2 \dots$

KVL  $v_2 = e_2 = \frac{N_2 d\phi}{dt}$

$$v_2 = \omega N_2 \phi_m \cos \omega t = 2\pi f N_2 \phi_m \sin(\omega t + \pi/2)$$

$$V_{a_{max}} = 2\pi f N_2 \phi_m$$

$$V_{a_{rms}} = 0.707 \times 2\pi f N_2 \phi_m = 4.44 f N_2 \phi_m$$

$$\frac{V_1}{V_2} = \frac{N_1}{N_2} = \text{Turn ratio.}$$

Supposing this to be an ideal transformer.

$$P_1 = P_2$$

$$V_{1_{rms}} \times I_{1_{rms}} \times \cos\phi = V_{2_{rms}} \times I_{2_{rms}} \times \cos\phi$$

$$\frac{V_1}{V_2} = \frac{I_2}{I_1} = \text{Turn ratio.}$$